

# CA3080, CA3080A Types

## Operational Transconductance Amplifiers (OTA's)

### Gateable-Gain Blocks

The RCA-CA3080 and CA3080A are Gateable-Gain Blocks which utilize the unique Operational Transconductance Amplifier (OTA) concept described in Application Note ICAN-6668, "Applications of the CA3080 and CA3080A High-Performance Operational Transconductance Amplifiers".

The CA3080 and CA3080A have differential input and a single-ended, push-pull, class A output. In addition, these types have an amplifier bias input which may be used either for gating or for linear gain control. These types also have a high output impedance and their transconductance ( $g_m$ ) is directly proportional to the amplifier bias current ( $I_{ABC}$ ).

The CA3080 and CA3080A are notable for their excellent slew rate (50 V/ $\mu$ s), which makes them especially useful for multiplex and fast unity-gain voltage followers. These types are especially applicable for multiplex applications because power is consumed only when the devices are in the "ON" channel state.

The CA3080A is rated for operation over the full military-temperature range (-55 to +125°C) and its characteristics are specifically controlled for applications such as sample-and-hold, gain-control, multiplex, etc. Operational transconductance amplifiers are also useful in programmable power-switch applications, e.g., as described in Application Note ICAN-6048, "Some Applications of a Programmable Power Switch/Amplifier" (CA3094, CA3094A, CA3094B).

These types are supplied in the 8-lead TO-5 style package (CA3080, CA3080A), and in the 8-lead TO-5 style package with dual-in-line formed leads ("DIL-CAN", CA3080S, CA3080AS). The CA3080E is also supplied in the 8-lead dual-in-line plastic ("MINI-DIP") package (CA3080E), and in chip form (CA3080H).

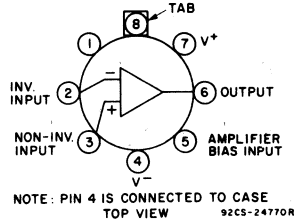
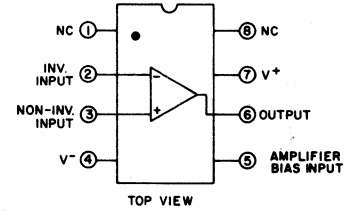


Fig. 1 - Functional diagrams.



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### Plastic Package (CA3080E)

#### Applications:

- Sample and hold
- Multiplex
- Voltage follower
- Multipplier
- Comparator

#### Features:

- Slew rate (unity gain, compensated): 50 V/ $\mu$ s
- Adjustable power consumption: 10  $\mu$ W to 30 mW
- Flexible supply voltage range:  $\pm 2$  V to  $\pm 15$  V
- Fully adjustable gain: 0 to  $g_{mR_L}$  limit
- Tight  $g_m$  spread: CA3080 (2:1), CA3080A (1.6:1)
- Extended  $g_m$  linearity: 3 decades

#### MAXIMUM RATINGS, Absolute-Maximum Values at $T_A = 25^\circ\text{C}$

DC SUPPLY VOLTAGE (between $V^+$ and $V^-$ terminals)	36 V
DIFFERENTIAL INPUT VOLTAGE	$\pm 5$ V
DC INPUT VOLTAGE	$V^+$ to $V^-$
INPUT SIGNAL CURRENT	1 mA
AMPLIFIER BIAS CURRENT	2 mA
OUTPUT SHORT-CIRCUIT DURATION*	Indefinite
DEVICE DISSIPATION	125 mW
TEMPERATURE RANGE:	
Operating	
CA3080	0 to +70 $^\circ\text{C}$
CA3080A	-55 to +125 $^\circ\text{C}$
Storage	-65 to +150 $^\circ\text{C}$
LEAD TEMPERATURE (During Soldering):	
At distance $1/16 \pm 1/32$ in. (1.59 $\pm$ 0.79 mm)	+265 $^\circ\text{C}$
from case for 10 s max.	

\* Short circuit may be applied to ground or to either supply.

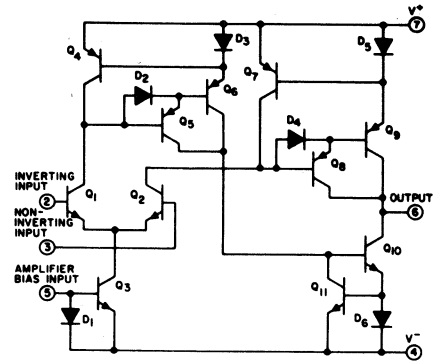


Fig. 2 - Schematic diagram for CA3080 and CA3080A.

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<sup>A</sup> Type CA3080 can be operated over the temperature of -55 to +125  $^\circ\text{C}$ , although the published limits for certain electrical specifications apply only over the temperature range of 0 to 70  $^\circ\text{C}$ .

### TYPICAL CHARACTERISTICS CURVES AND TEST CIRCUITS FOR THE CA3080 AND CA3080A

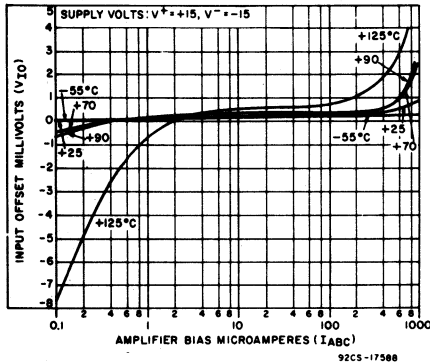


Fig. 3 - Input offset voltage vs. amplifier bias current.

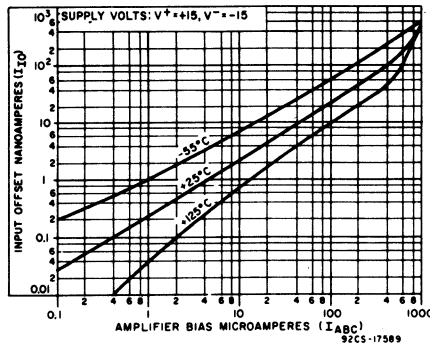


Fig. 4 - Input offset current vs. amplifier bias current.

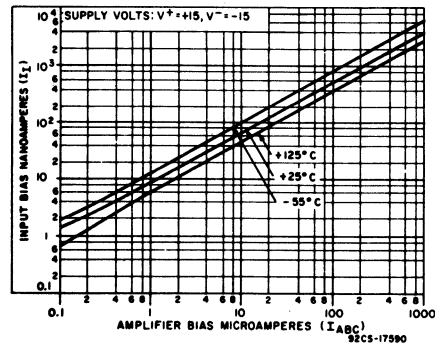


Fig. 5 - Input bias current vs. amplifier bias current.

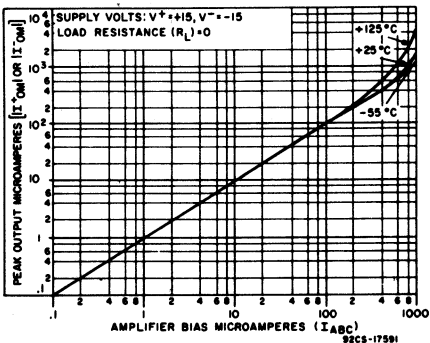


Fig. 6 - Peak output current vs. amplifier bias current.

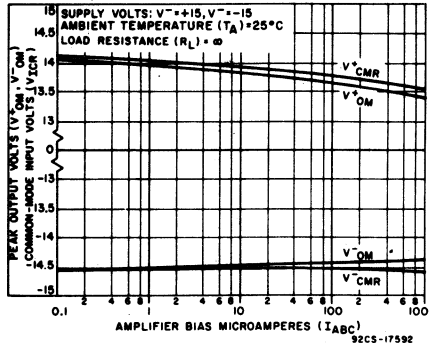


Fig. 7 - Peak output voltage vs. amplifier bias current.

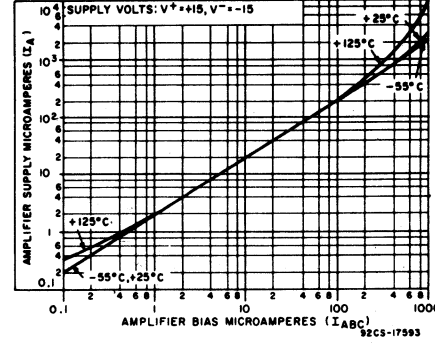


Fig. 8 - Amplifier supply current vs. amplifier bias current.

# CA3080, CA3080A Types

**ELECTRICAL CHARACTERISTICS**  
For Equipment Design

CA3080

CHARACTERISTICS	SYMBOLS	Circuit Fig.	TEST CONDITIONS		LIMITS			UNITS
			$V^+ = +15\text{ V}, V^- = -15\text{ V}$ $I_{ABC} = 500\ \mu\text{A}$ $T_A = 25^\circ\text{C}$ (unless indicated otherwise)	Typical Characteristics Curves Fig.	Min.	Typ.	Max.	
Input Offset Voltage	$V_{IO}$	-	$T_A = 0\text{ to }70^\circ\text{C}$	3	-	0.4	5	mV
Input Offset Current	$I_{IO}$	-		4	-	0.12	0.6	$\mu\text{A}$
Input Bias Current	$I_I$	-	$T_A = 0\text{ to }70^\circ\text{C}$	5	-	2	7	$\mu\text{A}$
Forward Transconductance (large signal)	$g_m$	-	$T_A = 0\text{ to }70^\circ\text{C}$	14	6700	9600	13000	$\mu\text{mho}$
Peak Output Current	$ I_{OM} $	-	$R_L = 0$ $R_L = 0, T_A = 0\text{ to }70^\circ\text{C}$	6	350	500	650	$\mu\text{A}$
Peak Output Voltage: Positive	$V^+_{OM}$	-	$R_L = \infty$	7	12	13.5	-	V
Negative	$V^-_{OM}$				-12	-14.4	-	
Amplifier Supply Current	$I_A$	-		8	0.8	1	1.2	mA
Device Dissipation	$P_D$	-		9	24	30	36	mW
Input Offset Voltage Sensitivity: Positive	$\Delta V_{IO}/\Delta V^+$	-		-	-	-	150	$\mu\text{V/V}$
Negative	$\Delta V_{IO}/\Delta V^-$	-		-	-	-	150	
Common-Mode Rejection Ratio	CMRR	-		-	80	110	-	dB
Common-Mode Input-Voltage Range	$V_{ICR}$	-		7	12 to -12	13.6 to -14.6	-	V
Input Resistance	$R_I$	-		15	10	26	-	$\text{k}\Omega$

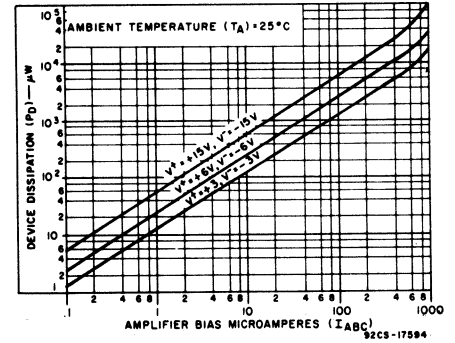


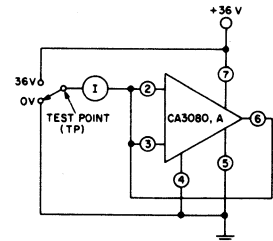
Fig. 9 - Total power dissipation vs. amplifier bias current.

**ELECTRICAL CHARACTERISTICS**

Typical Values Intended Only for Design Guidance

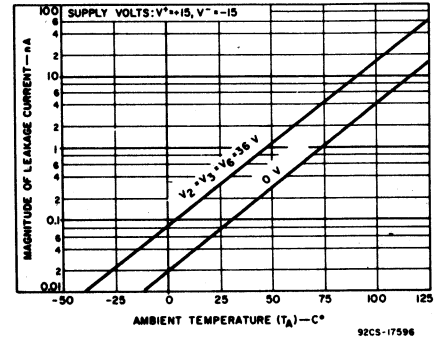
CA3080

Input Offset Voltage	$V_{IO}$	-	$I_{ABC} = 5\ \mu\text{A}$	3	0.3	mV
Input Offset Voltage Change	$ \Delta V_{IO} $	-	Change in $V_{IO}$ between $I_{ABC} = 500\ \mu\text{A}$ and $I_{ABC} = 5\ \mu\text{A}$	-	0.2	mV
Peak Output Current	$I_{OM}$	-	$I_{ABC} = 5\ \mu\text{A}$	6	5	$\mu\text{A}$
Peak Output Voltage: Positive	$V^+_{OM}$	-	$I_{ABC} = 5\ \mu\text{A}$	7	13.8	V
Negative	$V^-_{OM}$				-14.5	
Magnitude of Leakage Current		10	$I_{ABC} = 0, V_{TP} = 0$ $I_{ABC} = 0, V_{TP} = 36\ \text{V}$	11	0.08 0.3	nA
Differential Input Current		12	$I_{ABC} = 0, V_{DIFF} = 4\ \text{V}$	13	0.008	nA
Amplifier Bias Voltage	$V_{ABC}$	-		16	0.71	V
Slew Rate: Maximum (uncompensated)	SR	23			75	$\text{V}/\mu\text{s}$
Unity Gain (compensated)					50	
Open-Loop Bandwidth	BWOL	-		-	2	MHz
Input Capacitance	$C_I$	-	$f = 1\ \text{MHz}$	17	3.6	pF
Output Capacitance	$C_O$	-	$f = 1\ \text{MHz}$	17	5.6	pF
Output Resistance	$R_O$	-		18	15	$\text{M}\Omega$
Input-to-Output Capacitance	$C_{I-O}$	19	$f = 1\ \text{MHz}$	20	0.024	pF



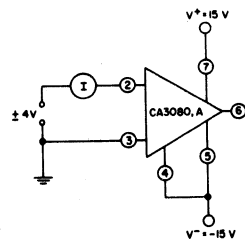
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Fig. 10 - Leakage current test circuit.



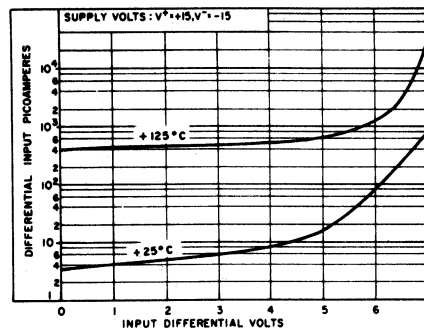
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Fig. 11 - Leakage current vs. temperature.



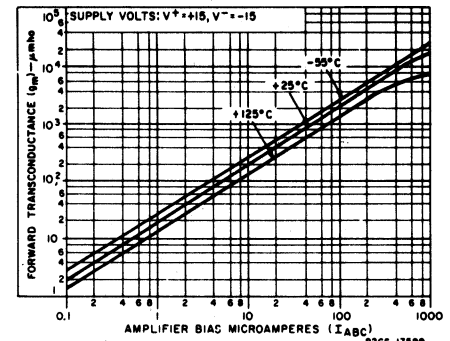
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Fig. 12 - Differential input current test circuit.



92CS-17598

Fig. 13 - Input current vs. input differential voltage.



92CS-17599

Fig. 14 - Transconductance vs. amplifier bias current.

# CA3080, CA3080A Types

## ELECTRICAL CHARACTERISTICS For Equipment Design

CA3080A

CHARACTERISTICS	SYMBOLS	Circuit Fig.	TEST CONDITIONS $V^+ = 15\text{ V}, V^- = -15\text{ V}$ $I_{ABC} = 500\ \mu\text{A}$ $T_A = 25^\circ\text{C}$ (unless indicated otherwise)	Typical Characteristics Curves Fig.	LIMITS			UNITS
					Min.	Typ.	Max.	
Input Offset Voltage	$V_{IO}$	—	$I_{ABC} = 5\ \mu\text{A}$	3	—	0.3	2	mV
		—	$T_A = -55\text{ to }+125^\circ\text{C}$		—	0.4	2	
		—			—	—	5	
Input Offset Voltage Change	$ \Delta V_{IO} $	—	Change in $V_{IO}$ between $I_{ABC} = 500\ \mu\text{A}$ and $I_{ABC} = 5\ \mu\text{A}$	3	—	0.1	3	mV
Input Offset Current	$I_{IO}$	—		4	—	0.12	0.6	$\mu\text{A}$
Input Bias Current	$I_I$	—	$T_A = -55\text{ to }+125$	5	—	2	5	$\mu\text{A}$
		—			—	—	8	
Forward Transconductance (large signal)	$g_m$	—	$T_A = -55\text{ to }+125^\circ\text{C}$	14	7700	9600	12000	$\mu\text{mho}$
		—			4000	—	—	
Peak Output Current	$ I_{OM} $	—	$I_{ABC} = 5\ \mu\text{A}, R_L = 0$	6	3	5	7	$\mu\text{A}$
		—	$R_L = 0$		350	500	650	
		—	$R_L = 0, T_A = -55\text{ to }+125^\circ\text{C}$		300	—	—	
Peak Output Voltage: Positive	$V^{+OM}$	—	$I_{ABC} = 5\ \mu\text{A}$	7	12	13.8	—	V
Negative	$V^{-OM}$	—	$R_L = \infty$		-12	-14.5	—	
Positive	$V^{+OM}$	—	$R_L = \infty$		-12	13.5	—	
Negative	$V^{-OM}$	—			-12	-14.4	—	
Amplifier Supply Current	$I_A$	—		8	0.8	1	1.2	mA
Device Dissipation	$P_D$	—		9	24	30	36	mW
Input Offset Voltage Sensitivity: Positive	$\Delta V_{IO}/\Delta V^+$	—		—	—	—	150	$\mu\text{V/V}$
Negative	$\Delta V_{IO}/\Delta V^-$	—			—	—	150	
Magnitude of Leakage Current		10	$I_{ABC} = 0, V_{TP} = 0$	11	—	0.08	5	nA
			$I_{ABC} = 0, V_{TP} = 36\text{ V}$		—	0.3	5	
Differential Input Current		12	$I_{ABC} = 0, V_{DIFF} = 4\text{ V}$	13	—	0.008	5	nA
Common-Mode Rejection Ratio	CMRR	—		—	80	110	—	dB
Common-Mode Input-Voltage Range	$V_{ICR}$	—		7	12 to -12	13.6 to -14.6	—	V
Input Resistance	$R_I$	—		15	10	26	—	k $\Omega$

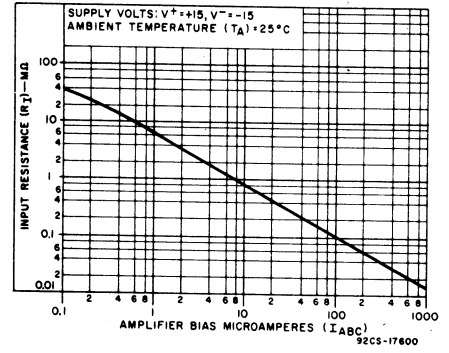


Fig. 15 - Input resistance vs. amplifier bias current.

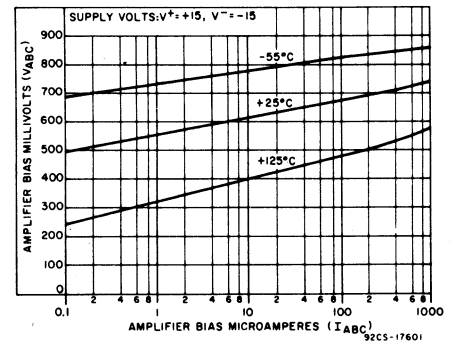


Fig. 16 - Amplifier bias voltage vs. amplifier bias current.

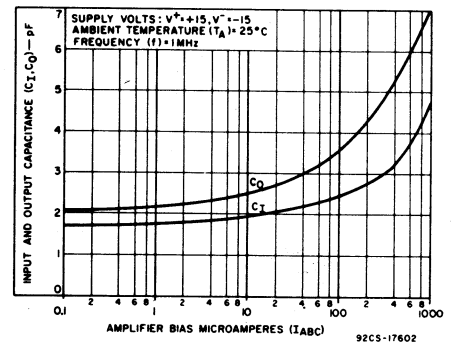


Fig. 17 - Input and output capacitance vs. amplifier bias current.

## ELECTRICAL CHARACTERISTICS Typical Values Intended Only For Design Guidance

CA3080A

Amplifier Bias Voltage	$V_{ABC}$	—		16	0.71	V
Slew Rate: Maximum (uncompensated)					75	$\text{V}/\mu\text{s}$
Unity Gain (uncompensated)	SR	23			50	
Open-Loop Bandwidth	BWOL	—			2	MHz
Input Capacitance	$C_I$	—	$f = 1\text{ MHz}$	17	3.6	pF
Output Capacitance	$C_O$	—	$f = 1\text{ MHz}$	17	5.6	pF
Output Resistance	$R_O$	—		18	15	M $\Omega$
Input-to-Output Capacitance	$C_{I-O}$	19	$f = 1\text{ MHz}$	20	0.024	pF

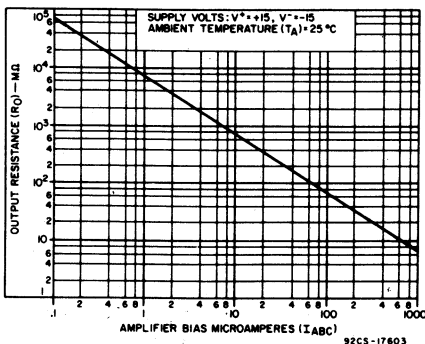


Fig. 18 - Output resistance vs. amplifier bias current.

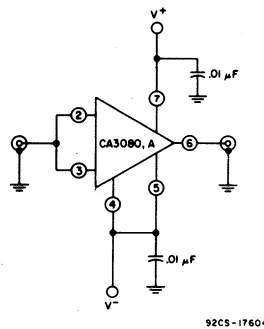


Fig. 19 - Input-to-output capacitance test circuit.

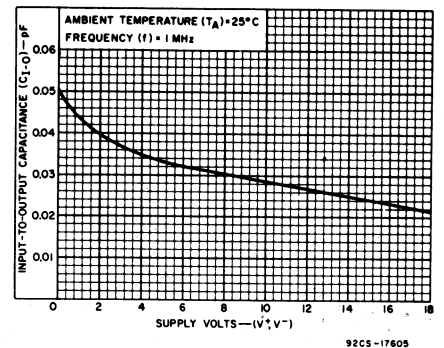


Fig. 20 - Input-to-output capacitance vs. supply voltage.

# CA3081, CA3082 Types

## General-Purpose High-Current N-P-N Transistor Arrays

CA3081—Common-Emitter Array    CA3082—Common-Collector Array

Directly Drive 7-Segment Incandescent Displays and Light-Emitting-Diode (LED) Displays

### Features

- 7 transistors permit a wide range of applications in either a common-emitter (CA3081) or common-collector (CA3082) configuration
- High  $I_C$ : 100 mA max.    Low  $V_{CE\ sat}$  (at 50 mA): 0.4 V typ.

### Applications

- Drivers for:
  - Incandescent display devices (e.g. RCA NUMITRON DR2000 Series and lamps)
  - LED (e.g. RCA-SG1002 GaAs High-Efficiency Emitting Diode)
  - Relay control    Thyristor firing

RCA-CA3081\* and CA3082\* consist of seven high-current (to 100 mA) silicon n-p-n transistors on a common monolithic substrate. The CA3081 is connected in a common-emitter configuration and the CA3082 is connected in a common-collector configuration.

The CA3081 and CA3082 are capable of directly driving seven-segment displays, such as the RCA NUMITRON devices (DR2000 and DR2010), and light-emitting diode

(LED) displays. These types are also well-suited for a variety of other driver applications, including relay control and thyristor firing.

The CA3081 and CA3082 are supplied in a 16-lead dual-in-line plastic package, and the CA3081F and CA3082F in a 16-lead dual-in-line frit-seal ceramic package, which includes a separate substrate connection for maximum flexibility in circuit design.

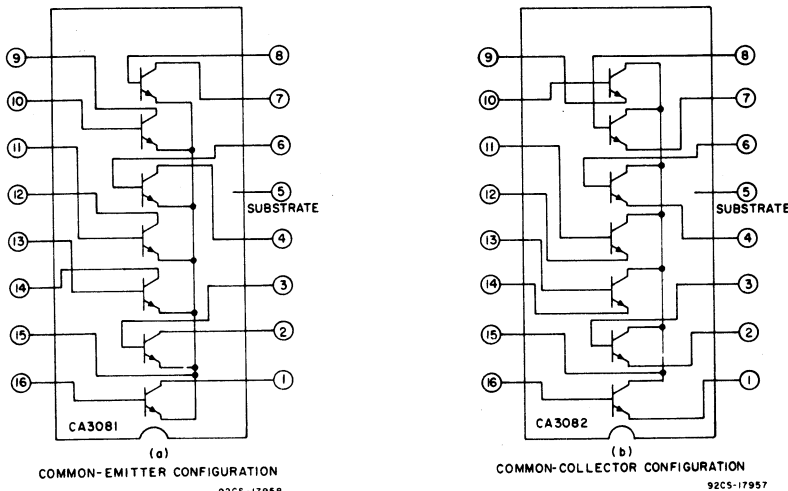


Fig. 1—Functional diagrams of types CA3081 and CA3082.

### TYPICAL STATIC CHARACTERISTICS FOR EACH TRANSISTOR OF TYPES CA3081 AND CA3082

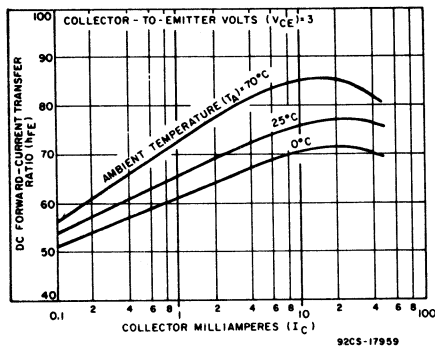


Fig. 2— $h_{FE}$  vs.  $I_C$

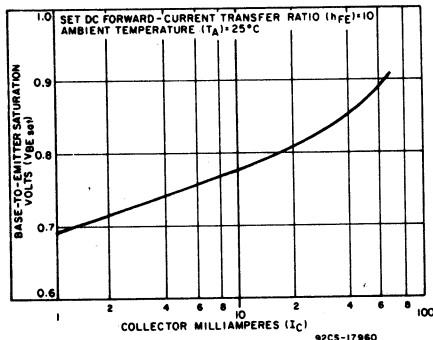


Fig. 3— $V_{BE\ sat}$  vs.  $I_C$

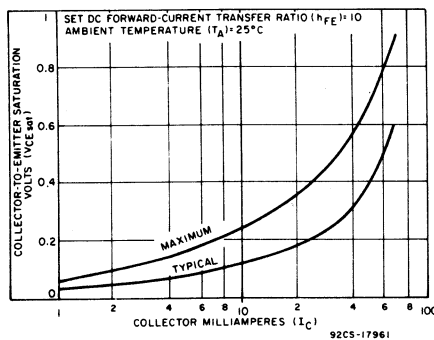


Fig. 4— $V_{CE\ sat}$  vs.  $I_C$  at  $T_A = 25^\circ C$ .

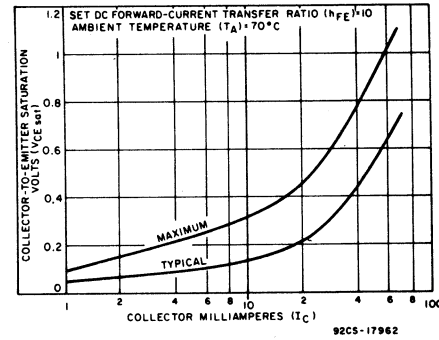


Fig. 5— $V_{CE\ sat}$  vs.  $I_C$  at  $T_A = 70^\circ C$ .

### MAXIMUM RATINGS, Absolute-Maximum Values at $T_A = 25^\circ C$

#### Power Dissipation:

Any one transistor	500	mW
Total package	750	mW
Above $55^\circ C$	Derate linearly 6.67	$mW/^\circ C$

#### Ambient Temperature Range:

Operating	-55 to +125	$^\circ C$
Storage	-65 to +150	$^\circ C$

#### Lead Temperature (During Soldering):

At distance $1/16'' \pm 1/32''$ (1.59 mm $\pm 0.79$ mm) from case for 10 seconds max.	265	$^\circ C$
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The following ratings apply for each transistor in the device:

Collector-to-Emitter Voltage ( $V_{CE0}$ )	16	V
Collector-to-Base Voltage ( $V_{CBO}$ )	20	V
Collector-to-Substrate Voltage ( $V_{C10}$ ) <sup>■</sup>	20	V
Emitter-to-Base Voltage ( $V_{EBO}$ )	5	V
Collector Current ( $I_C$ )	100	mA
Base Current ( $I_B$ )	20	mA

■ The collector of each transistor of the CA3081 and CA3082 is isolated from the substrate by an integral diode. The substrate must be connected to a voltage which is more negative than any collector voltage in order to maintain isolation between transistors and

provide normal transistor action. To avoid undesired coupling between transistors, the substrate terminal (5) should be maintained at either DC or signal (AC) ground. A suitable bypass capacitor can be used to establish a signal ground.

# CA3081, CA3082 Types

ELECTRICAL CHARACTERISTICS at  $T_A = 25^\circ\text{C}$   
For Equipment Design

CHARACTERISTIC	SYMBOL	TEST CONDITIONS	Typ. Char. Curve Fig. No.	LIMITS			UNITS
				Min.	Typ.	Max.	
Collector-to-Base Breakdown Voltage	$V_{(BR)CES}$	$I_C = 500 \mu\text{A}, I_E = 0$	—	20	60	—	V
Collector-to-Substrate Breakdown Voltage	$V_{(BR)CIO}$	$I_C = 500 \mu\text{A}, I_E = 0, I_B = 0$	—	20	60	—	V
Collector-to-Emitter Breakdown Voltage	$V_{(BR)CEO}$	$I_C = 1 \text{ mA}, I_B = 0$	—	16	24	—	V
Emitter-to-Base Breakdown Voltage	$V_{(BR)EBO}$	$I_C = 500 \mu\text{A}$	—	5	6.9	—	V
DC Forward-Current Transfer Ratio	hFE	$V_{CE} = 0.5 \text{ V}, I_C = 30 \text{ mA}$	—	30	68	—	
		$V_{CE} = 0.8 \text{ V}, I_C = 50 \text{ mA}$	—	40	70	—	
Base-to-Emitter Saturation Voltage	$V_{BE \text{ sat}}$	$I_C = 30 \text{ mA}, I_B = 1 \text{ mA}$	3	—	0.87	1.0	V
Collector-to-Emitter Saturation Voltage: CA3081, CA3082 CA3081 CA3082	$V_{CE \text{ sat}}$	$I_C = 30 \text{ mA}, I_B = 1 \text{ mA}$	—	—	0.27	0.5	V
		$I_C = 50 \text{ mA}, I_B = 5 \text{ mA}$	4	—	0.4	0.7	
		$I_C = 50 \text{ mA}, I_B = 5 \text{ mA}$	4	—	0.4	0.8	
Collector-Cutoff Current	$I_{CEO}$	$V_{CE} = 10 \text{ V}, I_B = 0$	—	—	—	10	$\mu\text{A}$
Collector-Cutoff Current	$I_{CBO}$	$V_{CB} = 10 \text{ V}, I_E = 0$	—	—	—	1	$\mu\text{A}$

### TYPICAL READ-OUT DRIVER APPLICATIONS

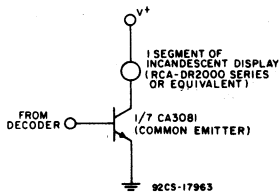


Fig. 6—Schematic diagram showing one transistor of the CA3081 driving one segment of an incandescent display.

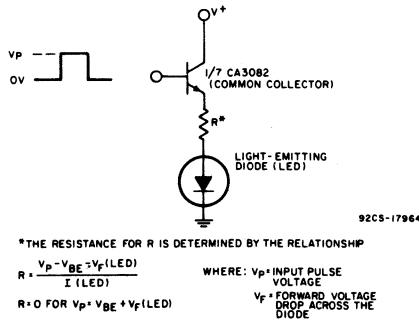


Fig. 7—Schematic diagram showing one transistor of the CA3082 driving a light-emitting diode (LED).

# CA3083

## General-Purpose High-Current N-P-N Transistor Array

RCA-CA3083 is a versatile array of five high-current (to 100mA) n-p-n transistors on a common monolithic substrate. In addition, two of these transistors (Q1 and Q2) are matched at low currents (i.e. 1mA) for applications in which offset parameters are of special importance.

Independent connections for each transistor plus a separate terminal for the substrate permit maximum flexibility in circuit design. The CA3083 is supplied in a 16-lead dual-in-line plastic package, and the CA3083F in a 16-lead dual-in-line frit-seal ceramic package.

### Features

- High  $I_C$ : 100mA max.
- Low  $V_{CEsat}$  (at 50mA): 0.7V max.
- Matched pair (Q1 and Q2) –  $V_{IO}$  ( $V_{BE}$  matched):  $\pm 5$  mV max.  $I_{IO}$  (at 1 mA): 2.5  $\mu$ A max.
- 5 independent transistors plus separate substrate connection
- The CA3083 is available in a sealed-junction Beam-Lead version (CA3083L). For further information see File No. 515, "Beam-Lead Devices for Hybrid Circuit Applications".

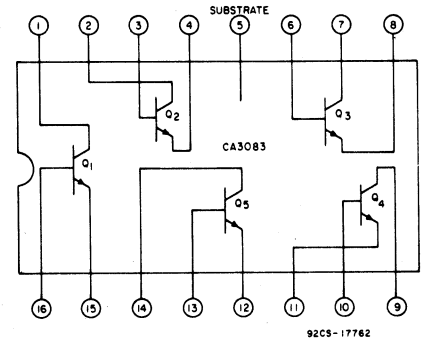


Fig.1—Functional diagram of the CA3083.

### Applications

- Signal processing and switching systems operating from DC to VHF
- Lamp and relay driver
- Differential amplifier
- Temperature-compensated amplifier
- Thyristor firing
- See RCA Application Note, ICAN-5296 "Application of the RCA-CA3018 Circuit Transistor Array" for suggested applications

### MAXIMUM RATINGS, Absolute-Maximum Values at $T_A = 25^\circ\text{C}$

Power Dissipation:			
Any one transistor	500	mW	
Total package	750	mW	
Above $55^\circ\text{C}$	Derate linearly 6.67		mW/ $^\circ\text{C}$
Ambient Temperature Range:			
Operating	-55 to +125	$^\circ\text{C}$	
Storage	-65 to +150	$^\circ\text{C}$	
Lead Temperature (During Soldering):			
At distance 1/16" $\pm$ 1/32" (1.59 mm $\pm$ 0.79 mm) from case for 10 seconds max.	265	$^\circ\text{C}$	
The following ratings apply for each transistor in the device:			
Collector-to-Emitter Voltage ( $V_{CEO}$ )	15	V	
Collector-to-Base Voltage ( $V_{CBO}$ )	20	V	
Collector-to-Substrate Voltage ( $V_{CIO}$ )	20	V	
Emitter-to-Base Voltage ( $V_{EBO}$ )	5	V	
Collector Current ( $I_C$ )	100	mA	
Base Current ( $I_B$ )	20	mA	

The collector of each transistor of the CA3083 is isolated from the substrate by an integral diode. The substrate must be connected to a voltage which is more negative than any collector voltage in order to maintain isolation between transistors and provide normal transistor action. To avoid undesired coupling between transistors, the substrate terminal (5) should be maintained at either DC or signal (AC) ground. A suitable bypass capacitor can be used to establish a signal ground.

### ELECTRICAL CHARACTERISTICS at $T_A = 25^\circ\text{C}$ For Equipment Design

CHARACTERISTICS	SYMBOL	TEST CONDITIONS	LIMITS			UNITS
			Min.	Typ.	Max.	
For Each Transistor:						
Collector-to-Base Breakdown Voltage	$V_{(BR)CBO}$	$I_C = 100\mu\text{A}, I_E = 0$	20	60	—	V
Collector-to-Emitter Breakdown Voltage	$V_{(BR)CEO}$	$I_C = 1\text{mA}, I_B = 0$	15	24	—	V
Collector-to-Substrate Breakdown Voltage	$V_{(BR)CIO}$	$I_{C1} = 100\mu\text{A}, I_B = 0, I_E = 0$	20	60	—	V
Emitter-to-Base Breakdown Voltage	$V_{(BR)EBO}$	$I_E = 500\mu\text{A}, I_C = 0$	5	6.9	—	V
Collector-Cutoff-Current	$I_{CEO}$	$V_{CE} = 10\text{V}, I_B = 0$	—	—	10	$\mu\text{A}$
Collector-Cutoff-Current	$I_{CBO}$	$V_{CB} = 10\text{V}, I_E = 0$	—	—	1	$\mu\text{A}$
DC Forward-Current Transfer Ratio	$h_{FE}$	$V_{CE} = 3\text{V}, I_C = 10\text{mA}, I_B = 50\text{mA}$	2	40	76	—
Base-to-Emitter Voltage	$V_{BE}$	$V_{CE} = 3\text{V}, I_C = 10\text{mA}$	0.65	0.74	0.85	V
Collector-to-Emitter Saturation Voltage	$V_{CEsat}$	$I_C = 50\text{mA}, I_B = 5\text{mA}$	—	0.40	0.70	V
For Transistors Q1 and Q2 (As a Differential Amplifier):						
Absolute Input Offset Voltage	$ V_{IO} $	$V_{CE} = 3\text{V}, I_C = 1\text{mA}$	7	—	1.2	5 mV
Absolute Input Offset Current	$ I_{IO} $		8	—	0.7	2.5 $\mu\text{A}$

### TYPICAL STATIC CHARACTERISTICS FOR EACH TRANSISTOR

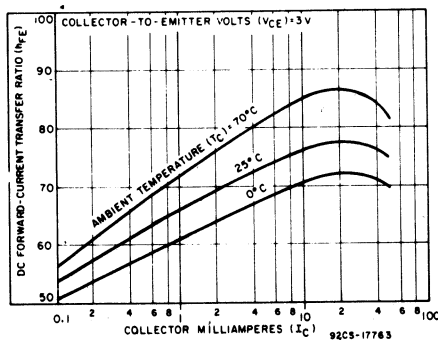


Fig.2— $h_{FE}$  vs  $I_C$

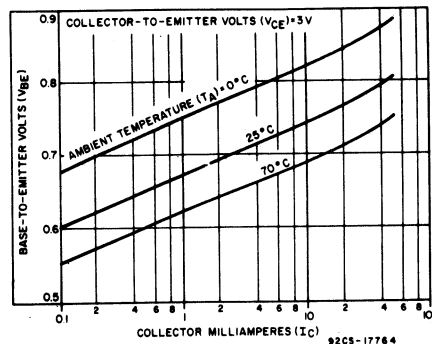


Fig.3— $V_{BE}$  vs  $I_C$

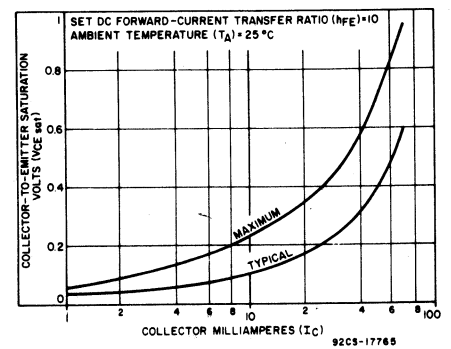


Fig.4— $V_{CEsat}$  vs  $I_C$  at  $25^\circ\text{C}$

CA3083

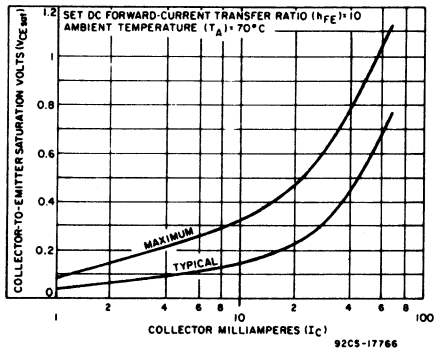


Fig.5 -  $V_{CEsat}$  vs  $I_C$  at 70°C

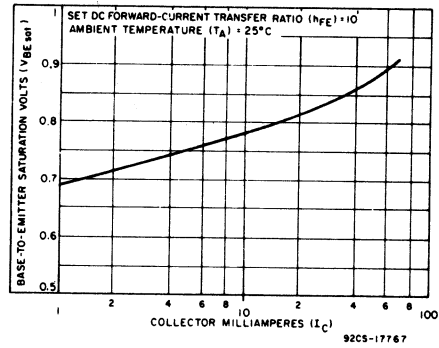


Fig.6 -  $V_{BEsat}$  vs  $I_C$

TYPICAL STATIC CHARACTERISTICS FOR DIFFERENTIAL AMPLIFIER

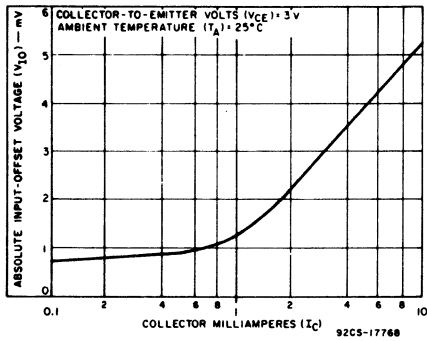


Fig.7 -  $V_{IO}$  vs  $I_C$  (transistors Q1 and Q2 as a differential amplifier).

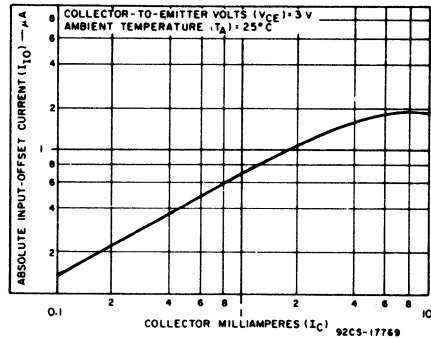


Fig.8 -  $I_{IO}$  vs  $I_C$  (transistors Q1 and Q2 as a differential amplifier).

# CA3084

## General-Purpose P-N-P Transistor Array

RCA-CA3084 is a general-purpose silicon p-n-p transistor array incorporating two independent transistors, a Darlington circuit, and a current-mirror pair with a shared diode.

The two independent transistors in the array may be used in a variety of circuit applications. The Darlington pair may be employed as the equivalent of a single high-beta transistor. The current-mirror pair is well suited for constant-current applications and can also be used as the active loads in a differential amplifier which uses n-p-n transistors.

The total array is especially useful for a wide range of applications in systems having low-power and low-frequency requirements. Although the transistors may be used as discrete units in conventional circuits, they offer the advantages inherent in integrated-circuit construction, that is, to provide close electrical and thermal matching.

The CA3084 utilizes the 14-lead dual-in-line plastic package.

### FEATURES

- Matched transistor pair (Q1 and Q2)
  - $V_{IO}$  ( $V_{BE}$  matched):  $\pm 6mV$  max.
  - $I_{IO}$  (at  $100 \mu A$ ):  $\pm 0.6 \mu A$
- Wide operating current range
- Low noise figure -- 3.2 dB typ. at 1 kHz
- The CA3084 is available in a sealed-junction Beam-Lead version (CA3084L). For further information see File No. 515, "Beam-Lead Devices for Hybrid Circuit Applications".

### APPLICATIONS

- General use in signal processing systems having low-power and low-frequency requirements
- Differential amplifiers
- Temperature compensated amplifiers
- Active loads for differential amplifiers using n-p-n transistors
- Complementary uses with RCA n-p-n transistor arrays

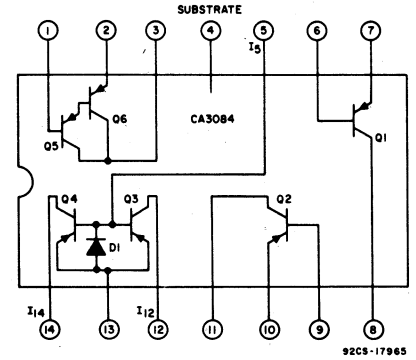


Fig. 1 - Functional diagram of the CA3084.

### STATIC CHARACTERISTICS FOR EACH TRANSISTOR

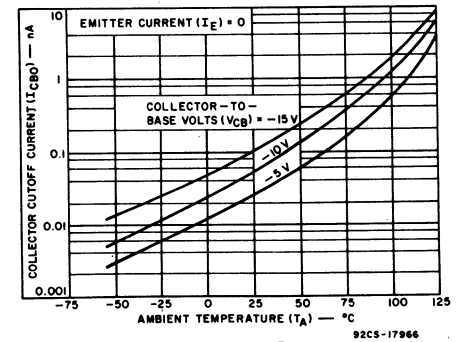


Fig. 2 -  $I_{CBO}$  vs  $T_A$ .

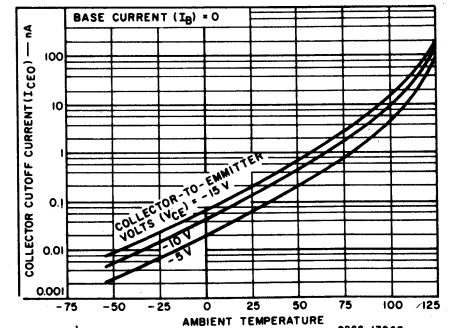


Fig. 3 -  $I_{CEO}$  vs  $T_A$ .

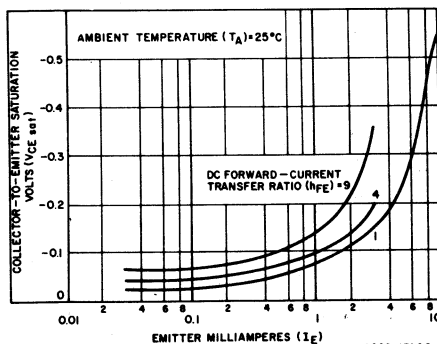


Fig. 4 -  $V_{CEsat}$  vs  $I_E$ .

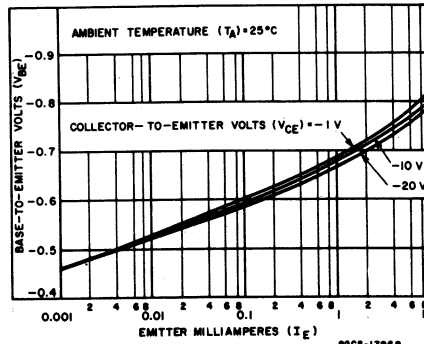


Fig. 5 -  $V_{BE}$  vs  $I_E$ .

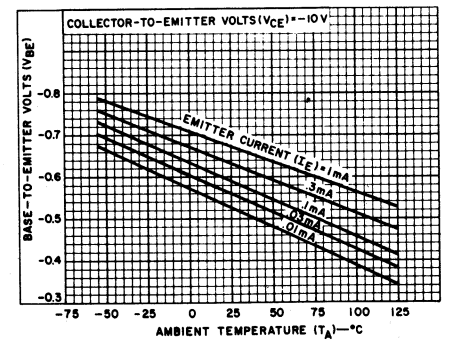


Fig. 6 -  $V_{BE}$  vs  $T_A$ .



# CA3084

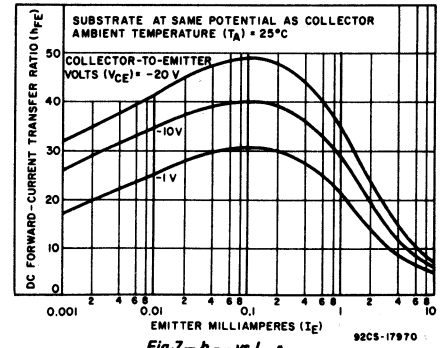
ELECTRICAL CHARACTERISTICS at  $T_A = 25^\circ\text{C}$   
For Equipment Design

CHARACTERISTICS	SYMBOL	TEST CONDITIONS	Typ. Characteristics Curve Fig. No.	LIMITS			UNITS
				Min.	Typ.	Max.	
<b>For Each Transistor:</b>							
Collector-Cutoff Current	$I_{CBO}$	$V_{CB} = -10\text{V}, I_E = 0$	2	—	-0.055	-100	nA
Collector-Cutoff Current	$I_{CEO}$	$V_{CE} = -10\text{V}, I_B = 0$	3	—	-0.12	-100	nA
Collector-to-Emitter Breakdown Voltage	$V_{(BR)CEO}$	$I_{CE} = -100\mu\text{A}, I_B = 0$	—	-40	-70	—	V
Collector-to-Base Breakdown Voltage	$V_{(BR)CBO}$	$I_{CB} = -100\mu\text{A}, I_E = 0$	—	-40	-80	—	V
Emitter-to-Base Breakdown Voltage	$V_{(BR)EBO}$	$I_{EB} = -100\mu\text{A}, I_C = 0$	—	-40	-100	—	V
Emitter-to-Substrate Breakdown Voltage	$V_{(BR)EIO}$	$I_{EI} = 100\mu\text{A}$	—	40	100	—	V
Collector-to-Emitter Saturation Voltage	$V_{CEsat}$	$I_E = 1\text{mA}, I_B = 100\mu\text{A}$	4	—	-0.125	-0.25	V
Base-to-Emitter Voltage	$V_{BE}$	$I_E = 100\mu\text{A}, V_{CE} = -10\text{V}$	5	-0.50	-0.59	-0.68	V
DC Forward-Current Transfer Ratio	$h_{FE}$	$I_E = 100\mu\text{A}, V_{CE} = -10\text{V}$	7	15	40	—	
<b>For Transistors Q1 and Q2 (As a Differential Amplifier):</b>							
Magnitude of Input Offset Voltage	$ V_{IO} $	$I_E = 100\mu\text{A}, V_{CE} = -10\text{V}$	8	—	0.422	6	mV
Input Offset Current	$I_{IO}$		—	-0.6	0	0.6	$\mu\text{A}$
<b>For Transistors Q3 and Q4 (Current-Mirror Configuration):</b>							
Collector Current (Normalized)	$I_C/I_5$	$V_{CE} = -5\text{V}, V_{CIO} = -5\text{V}$	10	0.85	1.00	1.15	
Magnitude of Collector Current Ratio	$ I_C(Q3)/I_C(Q4) $	Term. 13 = Gnd. $I_5 = -100\mu\text{A}$	11	0.90	1.00	1.10	
<b>For Transistors Q5 and Q6 (Darlington Configuration):</b>							
Collector-Cutoff Current	$I_{CEO}$	$V_{CE} = -10\text{V}, I_B = 0$	—	—	—	-1.0	$\mu\text{A}$
Base-to-Emitter Voltage	$V_{BE}$	$I_E = 100\mu\text{A}, V_{CE} = -10\text{V}$	13	0.92	1.07	1.20	V
DC Forward-Current Transfer Ratio	$h_{FE}$		15	100	1230	—	

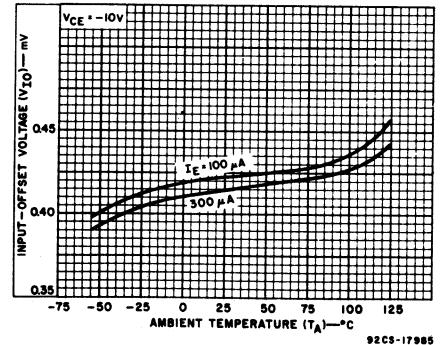
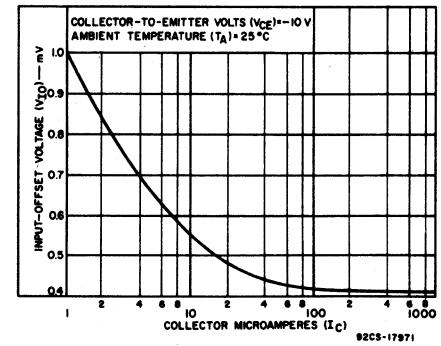
ELECTRICAL CHARACTERISTICS at  $T_A = 25^\circ\text{C}$   
Typical Values Intended Only For Design Guidance

<b>Magnitude of Temperature Coefficient:</b>							
$V_{BE}$ (for each transistor)	$ \Delta V_{BE}/\Delta T $	$I_E = 100\mu\text{A}$	6	—	-1.78	—	$\text{mV}/^\circ\text{C}$
$V_{IO}$ (as a differential amplifier)	$ \Delta V_{IO}/\Delta T $	$V_{CE} = -10\text{V}$	9	—	0.54	—	$\mu\text{V}/^\circ\text{C}$
$V_{BE}$ (Darlington configuration)	$ \Delta V_{BE}/\Delta T $		14	—	-3.7	—	$\text{mV}/^\circ\text{C}$
<b>For Each Transistor:</b>							
Input Resistance	$R_i$	$f = 1\text{kHz}, V_{CE} = -10\text{V}$	19	—	9	—	$\text{k}\Omega$
Output Resistance	$R_o$	$I_C = -100\mu\text{A}$	20	—	600	—	$\text{k}\Omega$
Forward Transconductance	$g_m$		22	—	3	—	$\text{mmho}$
Collector-to-Base Capacitance	$C_{CBO}$	$I_{CB} = 0$	23	—	3.3	—	pF
Collector-to-Emitter Capacitance	$C_{CEO}$	$I_{CE} = 0$	23	—	2.5	—	pF
Base-to-Substrate Capacitance	$C_{BIO}$	$I_{CIO} = 0$	23	—	4.5	—	pF

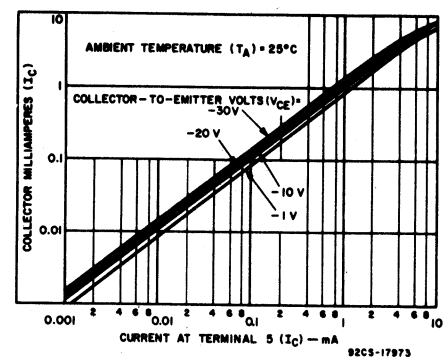
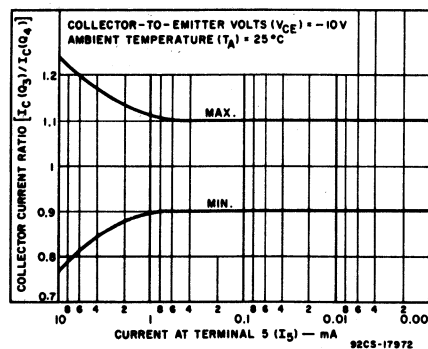
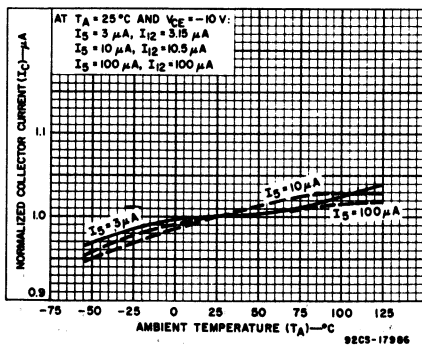
## STATIC CHARACTERISTICS FOR EACH TRANSISTOR



## STATIC CHARACTERISTICS FOR DIFFERENTIAL AMPLIFIER



## STATIC CHARACTERISTICS FOR CURRENT-MIRROR CONFIGURATION



# CA3084

## STATIC CHARACTERISTICS FOR DARLINGTON CONFIGURATION

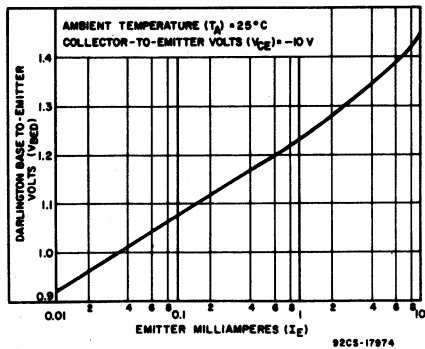


Fig. 13 - V<sub>BE</sub> vs I<sub>E</sub> (transistors Q5 and Q6 in a darlington configuration).

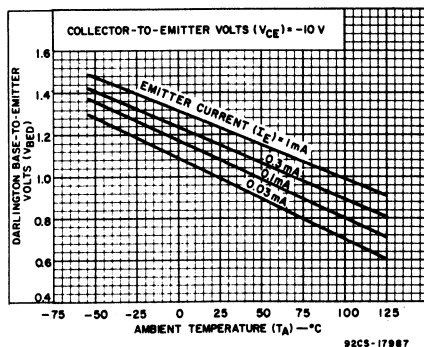


Fig. 14 - V<sub>BE</sub> vs T<sub>A</sub> (transistors Q5 and Q6 in a darlington configuration).

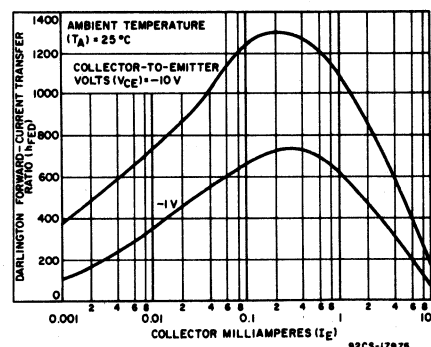


Fig. 15 - h<sub>FE</sub> vs I<sub>E</sub> (transistors Q5 and Q6 in a darlington configuration).

## DYNAMIC CHARACTERISTICS FOR EACH TRANSISTOR

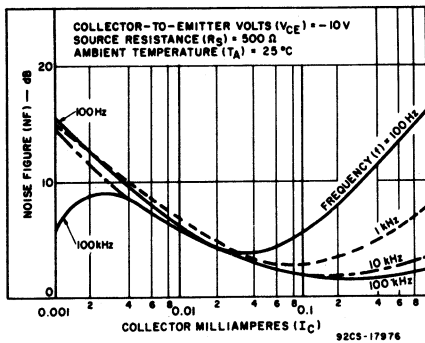


Fig. 16 - NF vs I<sub>C</sub> at R<sub>S</sub> = 500 Ω

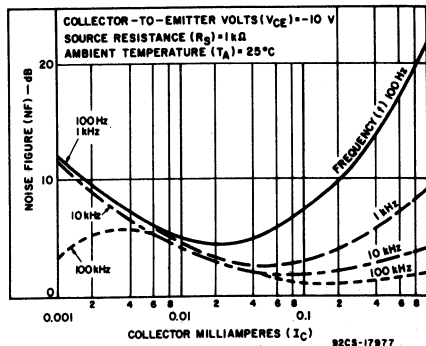


Fig. 17 - NF vs I<sub>C</sub> at R<sub>S</sub> = 1 kΩ

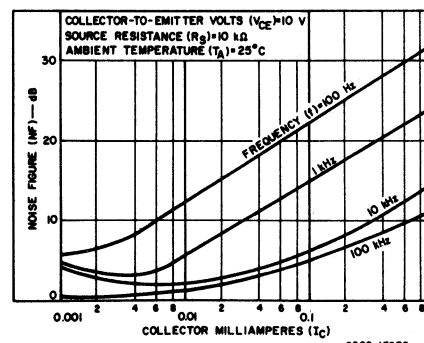


Fig. 18 - NF vs I<sub>C</sub> at R<sub>S</sub> = 10kΩ

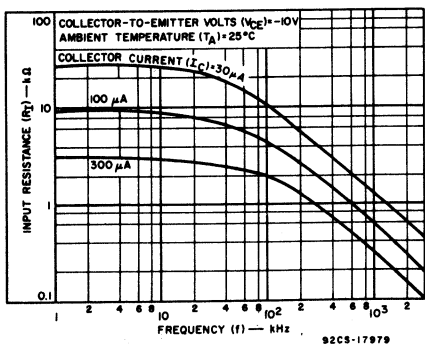


Fig. 19 - R<sub>i</sub> vs f

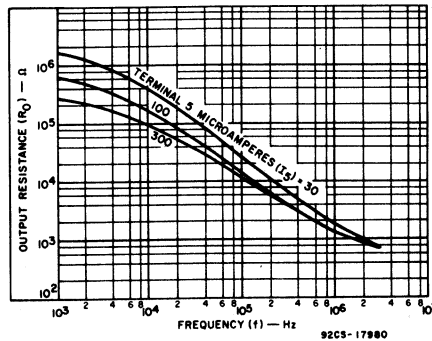


Fig. 20 - R<sub>o</sub> vs f

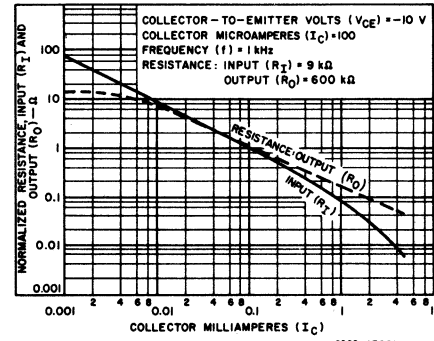


Fig. 21 - Normalized R<sub>i</sub> and R<sub>o</sub> vs I<sub>C</sub>

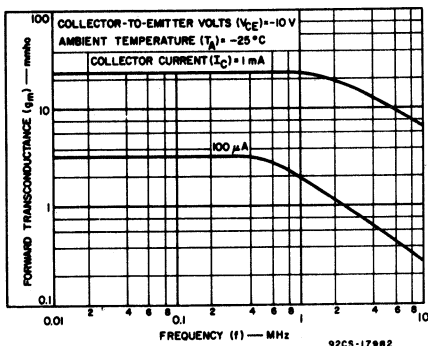


Fig. 22 - g<sub>m</sub> vs f

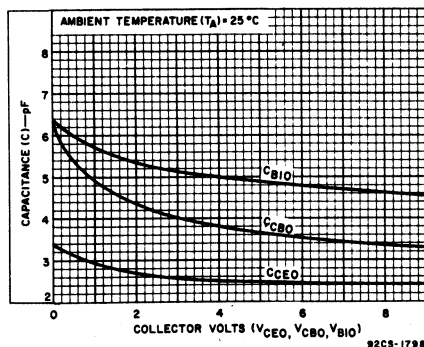


Fig. 23 - Transistor capacitances vs collector voltages (V<sub>CEO</sub>, V<sub>CBO</sub>, V<sub>B1O</sub>)

# CA3085, CA3085A, CA3085B Types

## Positive Voltage Regulators

For Regulated Voltages from 1.7V to 46V at Currents up to 100 mA

RCA-CA3085, CA3085A, and CA3085B are silicon monolithic integrated circuits designed specifically for service as voltage regulators at output voltages ranging from 1.7 to 46 volts at currents up to 100 milliamperes.

A block diagram of the CA3085 Series is shown in Fig. 1. The diagram shows the connecting terminals that provide access to the regulator circuit components. The voltage regulators provide important features such as: frequency compensation, short-circuit protection, temperature-compensated reference voltage, current limiting, and booster input. These devices are useful in a wide range of applications for regulating high-current, switching, shunt, and positive and negative voltages. They are also applicable for current and dual-tracking regulation.

The CA3085A and CA3085B have output current capabilities up to 100 mA and the CA3085 up to 12 mA without the use of external pass transistors. However, all the devices can provide voltage regulation at load currents greater than 100 mA with the use of suitable external pass transistors. The CA3085 Series has an unregulated input voltage ranging from 7.5 to 30 V (CA3085), 7.5 to 40 V (CA3085A), and 7.5 to 50 V (CA3085B) and a minimum regulated output voltage of 26 V (CA3085), 36 V (CA3085A), and 46 V (CA3085B).

The CA3085A is unilaterally interchangeable with the CA3055.

The CA3085 is available in a sealed-junction Beam-Lead version (CA3085L). For further information see File No. 515, "Beam-Lead Devices for Hybrid Circuit Applications".

Type	V <sub>IN</sub> Range V	V <sub>OUT</sub> Range V	Max I <sub>OUT</sub> mA	Max. Load Regulation % V <sub>OUT</sub>
CA3085	7.5 to 30	1.8 to 26	12*	0.1
CA3085A	7.5 to 40	1.7 to 36	100	0.15
CA3085B	7.5 to 50	1.7 to 46	100	0.15

\* This value may be extended to 100 mA; however, regulation is not specified beyond 12 mA.

These types are supplied in the 8-lead TO-5 style package (CA3085, CA3085A, CA3085B, and the 8-lead TO-5 with dual-in-line formed leads ("DIL-CAN", CA3085S, CA3085AS, CA3085BS). The CA3085 is also supplied in the 8-lead dual-in-line plastic package ("MINI-DIP", CA3085E), and in chip form (CA3085H).

### Features

- Up to 100 mA output current
- Input and output short-circuit protection
- Load and line regulation: 0.025%
- Pin compatible with LM100 Series
- Adjustable output voltage
- Low noise

### Applications

- Shunt voltage regulator
- Current regulator
- Switching voltage regulator
- High-current voltage regulator
- Combination positive and negative voltage regulator
- Dual tracking regulator
- See Application Note ICAN-6157 "Applications of the CA3085-Series Monolithic IC Voltage Regulators".

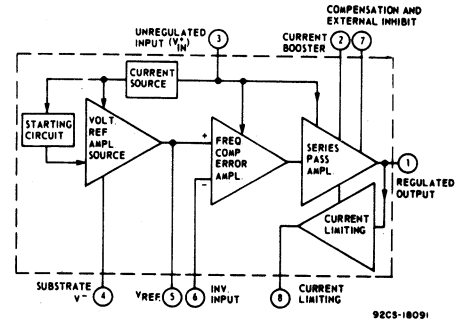


Fig. 1—Block diagram of CA3085 Series.

### MAXIMUM RATINGS, ABSOLUTE-MAXIMUM VALUES at T<sub>A</sub> = 25°C

POWER DISSIPATION: WITHOUT HEAT SINK	WITH HEAT SINK (TO-5 ONLY)
up to T <sub>A</sub> = 55°C ..... 630 mW	up to T <sub>C</sub> = 55°C ..... 1.6 W
above T <sub>A</sub> = 55°C ..... derate linearly @ 6.67 mW/°C	above T <sub>C</sub> = 55°C ..... derate linearly at 16.7 mW/°C

### TEMPERATURE RANGE:

Operating	..... -55 to +125°C
Storage	..... -65 to +150°C

### UNREGULATED INPUT VOLTAGE:

CA3085	..... 30 V
CA3085A	..... 40 V
CA3085B	..... 50 V

### LEAD TEMPERATURE (DURING SOLDERING):

At distance 1/16 ± 1/32 inch (1.59 ± 0.79mm) from case for 10 seconds max. .... +265°C

### Maximum Voltage Ratings

The following chart gives the range of voltages which can be applied to the terminals listed vertically with respect to the terminals listed horizontally. For example, the voltage range between vertical Terminal No. 7 and horizontal Terminal No. 1 is +3 to -10 volts.

### MAXIMUM VOLTAGE RATINGS

TERMINAL No.	5	6	7	8	1	2	3	4	
5	-	+5 -5	-	-	-	-	-	+10 0	* Voltages are not normally applied between these terminals; however, voltages appearing between these terminals are safe, if the specified voltage limits between all other terminals are not exceeded. ±30 V for CA3085 40 V for CA3085A 50 V for CA3085B
6	-	-	-	-	-	-	-	-	
7	-	-	-	+3 -10	+3 -10	-	-	+1 0	
8	-	-	-	-	+5 -1	-	-	-	
1	-	-	-	-	-	+10 -1	0 -1	+1 0	
2	-	-	-	-	-	-	0 -1	+1 0	
3	-	-	-	-	-	-	-	+1 0	
4	-	-	-	-	-	-	-	Substrate & Case	

### MAXIMUM CURRENT RATINGS

TERMINAL No.	I <sub>IN</sub> mA	I <sub>OUT</sub> mA
5	10	1.0
6	1.0	-0.1
7	1.0	-1.0
8	0.1	10
1	20	150
2	150	60
3	150	60
4	-	-

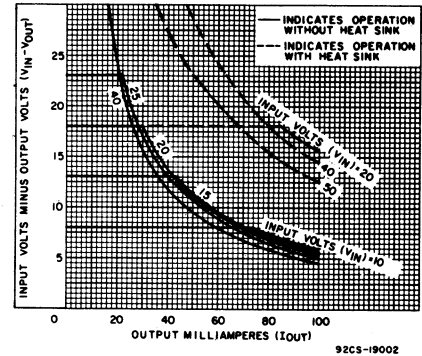


Fig. 3—Dissipation limitation (V<sub>IN</sub>-V<sub>OUT</sub> vs. I<sub>OUT</sub>).

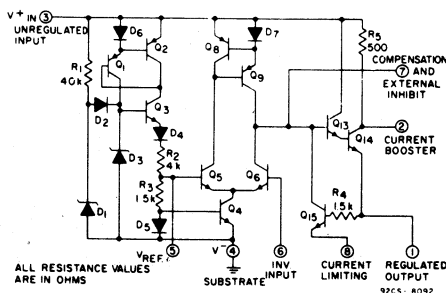


Fig. 2—Schematic diagram of CA3085 Series.

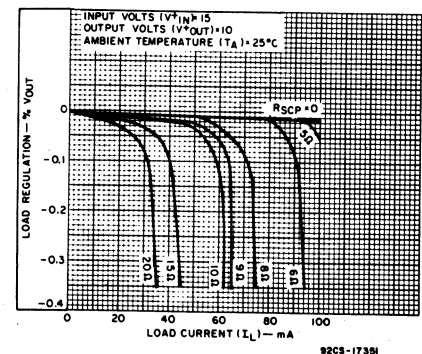


Fig. 4—Load regulation characteristics.

# CA3085, CA3085A, CA3085B Types

## ELECTRICAL CHARACTERISTICS

CHARACTERISTICS	SYMBOL	Test Circuit Fig. No.	TEST CONDITIONS			LIMITS									UNITS
			T <sub>A</sub> = 25°C (Unless indicated otherwise)	CA3085			CA3085A			CA3085B					
				MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.			
Reference Voltage	V <sub>REF</sub>	15	V <sup>+</sup> <sub>IN</sub> = 15V	1.4	1.6	1.8	1.5	1.6	1.7	1.5	1.6	1.7	V		
Quiescent Regulator Current	I <sub>quiescent</sub>	15	V <sup>+</sup> <sub>IN</sub> = 30V	-	3.3	4.5	-	-	-	-	-	-	mA		
			V <sup>+</sup> <sub>IN</sub> = 40V	-	-	-	-	3.65	5	-	-	-			
			V <sup>+</sup> <sub>IN</sub> = 50V	-	-	-	-	-	-	-	4.05	7		-	
				-	-	-	-	-	-	-	-	-		-	
Input Voltage Range	V <sub>IN(range)</sub>	-	-	7.5	-	30	7.5	-	40	7.5	-	.50	V		
Maximum Output Voltage	V <sub>O(max.)</sub>	15	V <sup>+</sup> <sub>IN</sub> = 30, 40, 50V <sup>+</sup> ; R <sub>L</sub> = 365 Ω; Term. No. 6 to Gnd.	26	27	-	36	37	-	46	47	-	V		
Minimum Output Voltage	V <sub>O(min.)</sub>	15	V <sup>+</sup> <sub>IN</sub> = 30V	-	1.6	1.8	-	1.6	1.7	-	1.6	1.7	V		
Input-Output Voltage Differential	V <sub>IN-VOUT</sub>	-	-	4	-	28	4	-	38	3.5	-	48	V		
Limiting Current	I <sub>LIM</sub>	16	V <sup>+</sup> <sub>IN</sub> = 16V, V <sup>+</sup> <sub>OUT</sub> = 10V R <sub>SCP</sub> = 6 Ω	-	96	120	-	96	120	-	96	120	mA		
Load Regulation <sup>⊙</sup>	-	-	I <sub>L</sub> = 1 to 100mA, R <sub>SCP</sub> = 0	-	-	-	-	0.025	0.15	-	0.025	0.15	%V <sub>OUT</sub>		
			I <sub>L</sub> = 1 to 100mA, R <sub>SCP</sub> = 0 T <sub>A</sub> = 0°C to +70°C	-	-	-	-	0.035	0.6	-	0.035	0.6			
			I <sub>L</sub> = 1 to 12mA, R <sub>SCP</sub> = 0	-	0.003	0.1	-	-	-	-	-	-			
Line Regulation <sup>⊙</sup>	-	-	I <sub>L</sub> = 1mA, R <sub>SCP</sub> = 0	-	0.025	0.1	-	0.025	0.075	-	0.025	0.04	%V		
			T <sub>A</sub> = 0°C to +70°C	-	0.04	0.15	-	0.04	0.1	-	0.04	0.08			
Equivalent Noise Output Voltage <sup>⊠</sup>	V <sub>NOISE</sub>	12	V <sup>+</sup> <sub>IN</sub> = 25V	C <sub>REF</sub> = 0	-	0.5	-	0.5	-	0.5	-	-	mV p p		
			C <sub>REF</sub> = 0.22μF	-	0.3	-	0.3	-	0.3	-	0.3	-			
Ripple Rejection	-	13	V <sup>+</sup> <sub>IN</sub> = 25V	C <sub>REF</sub> = 0	-	50	-	50	-	45	50	-	dB		
			f = 1kHz	C <sub>REF</sub> = 2μF	-	56	-	56	-	50	56	-			
Output Resistance	r <sub>o</sub>	13	V <sup>+</sup> <sub>IN</sub> = 25V, f = 1kHz	-	0.075	1.1	-	0.075	0.3	-	0.075	0.3	Ω		
Temperature Coefficient of Reference and Output Voltages	ΔV <sub>REF</sub> , ΔV <sub>O</sub>	-	I <sub>L</sub> = 0, V <sub>REF</sub> = 1.6V	-	0.0035	-	-	0.0035	-	-	0.0035	-	%/°C		
Load Transient Recovery Time: Turn On	t <sub>ON</sub>	14	V <sup>+</sup> <sub>IN</sub> = 25V, +50mA Step	-	1	-	-	1	-	-	1	-	μs		
			V <sup>+</sup> <sub>IN</sub> = 25V, -50mA Step	-	3	-	-	3	-	-	3	-			
Line Transient Recovery Time: Turn On	t <sub>ON</sub>	-	V <sup>+</sup> <sub>IN</sub> = 25V, f = 1kHz, 2V Step	-	0.8	-	-	0.8	-	-	0.8	-	μs		
				Turn Off	t <sub>OFF</sub>	-	0.4	-	-	0.4	-	-		0.4	-

# 30V (CA3085), 40V (CA3085A), 50V (CA3085B)

\* RSCP: Short-circuit protection resistance

⊠ Bandwidth DC to 10 MHz.

⊙ Load Regulation =  $\frac{\Delta V_{OUT}}{V_{OUT(initial)}} \times 100\%$

⊙ Line Regulation =  $\frac{\Delta V_{OUT}}{|V_{OUT(initial)}| |\Delta V_{IN}|} \times 100\%$

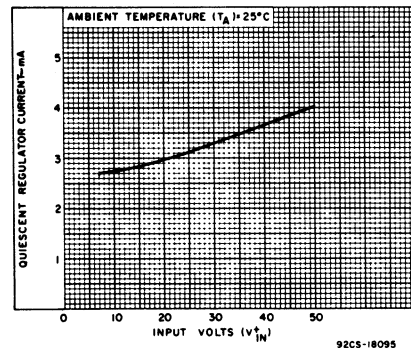


Fig.5—*I*<sub>quiescent</sub> vs. V<sub>IN</sub><sup>+</sup>

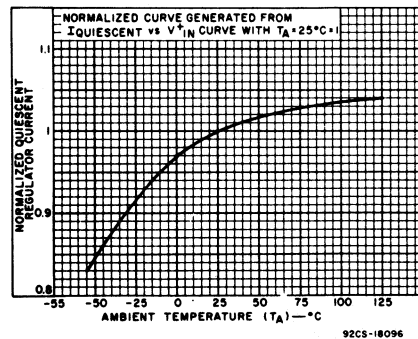


Fig.6—Normalized *I*<sub>quiescent</sub> vs. T<sub>A</sub>

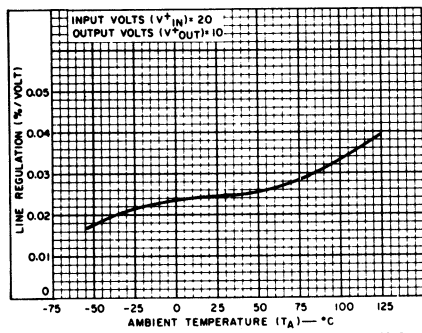


Fig.7—Line regulation temperature characteristics.

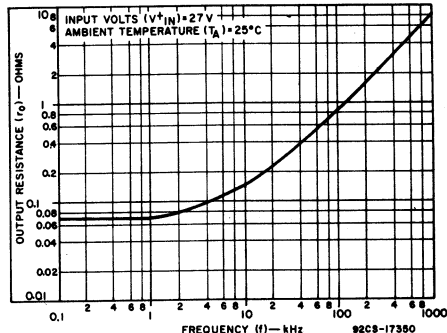


Fig.8—*r*<sub>o</sub> vs. *f*.

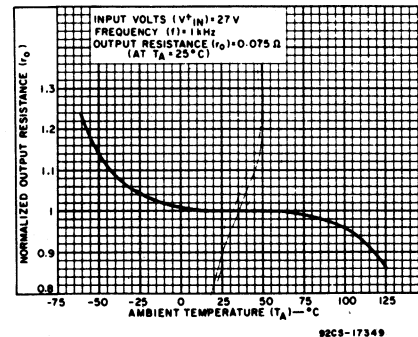


Fig.9—Normalized *r*<sub>o</sub> vs. T<sub>A</sub>

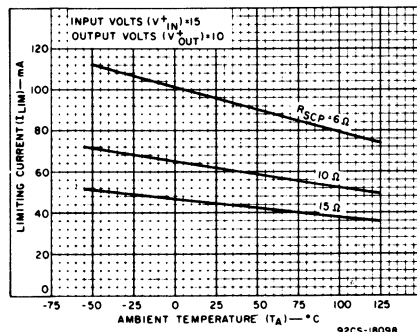


Fig.10—*I*<sub>LIM</sub> vs. T<sub>A</sub>

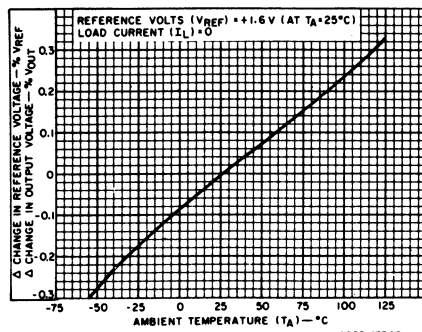


Fig.11—Temperature coefficient of V<sub>REF</sub> and V<sub>OUT</sub>.

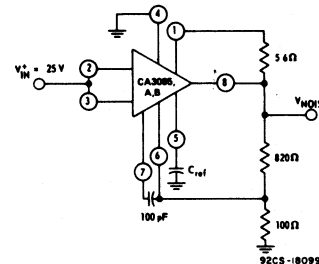


Fig.12—Test circuit for noise voltage.

# CA3085, CA3085A, CA3085B Types

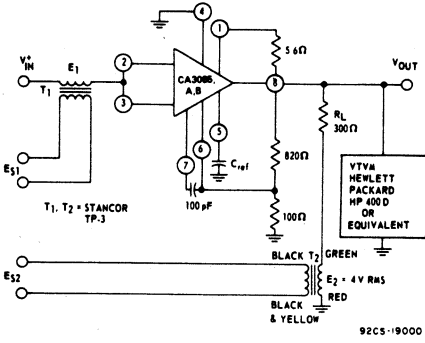


Fig. 13—Test circuit for ripple rejection and output resistance.

### TEST PROCEDURES FOR TEST CIRCUIT FOR RIPPLE REJECTION AND OUTPUT RESISTANCE

#### Output Resistance

Conditions:

1.  $V_{IN} = +25V$ ,  $C_{REF} = 0$ , Short  $E_1$
2. Set  $E_2$  at 1kHz so that  $E_2 = 4V$  rms
3. Read  $V_{OUT}$  on a VTVM, such as a Hewlett-Packard, HP400D or equivalent
4. Calculate  $R_{OUT}$  from  $R_{OUT} = V_{OUT} (I_L / E_2)$

#### Ripple Rejection - I

Conditions:

1.  $V_{IN} = +25V$ ,  $C_{REF} = 0$ , Short  $E_2$
2. Set  $E_1$  at 1kHz so that  $E_1 = 3V$  rms
3. Read  $V_{OUT}$  on a VTVM, such as a Hewlett-Packard, HP400D or equivalent
4. Calculate Ripple Rejection from  $20 \log (E_1 / V_{OUT})$

#### Ripple Rejection - II

Conditions:

1. Repeat Ripple Rejection I with  $C_{REF} = 2 \mu F$

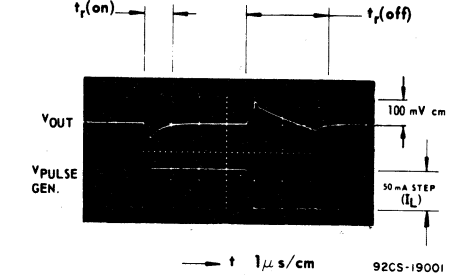
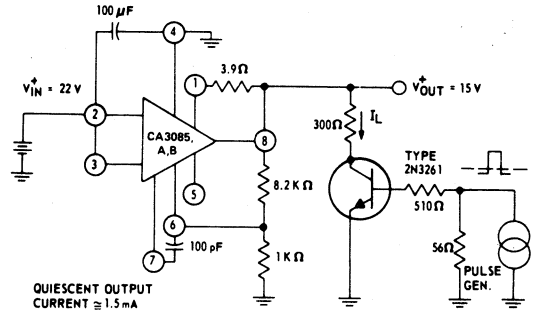
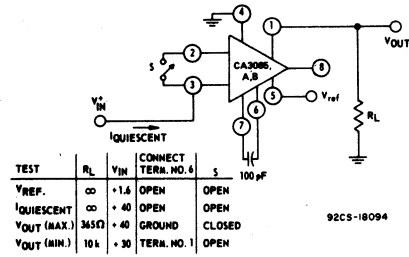
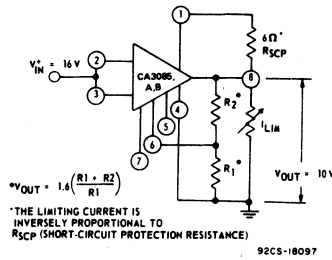


Fig. 14—Turn-on and turn-off recovery time test circuit with associated waveforms.



TEST	$R_L$	$V_{IN}$	CONNECT TERM. NO. 4	$S$
$V_{REF}$	$\infty$	+1.6	OPEN	OPEN
Iquiescent	$\infty$	+40	OPEN	OPEN
$V_{OUT} (MAX.)$	365Ω	+40	CLOSED	CLOSED
$V_{OUT} (MIN.)$	10k	+30	TERM. NO. 1	OPEN



$V_{OUT} = 1.6 \left( \frac{R_1 + R_2}{R_1} \right)$

\*THE LIMITING CURRENT IS INVERSELY PROPORTIONAL TO  $R_{SCP}$  (SHORT-CIRCUIT PROTECTION RESISTANCE)

Fig. 16—Test circuit for limiting current

Fig. 15—Test circuit for  $V_{REF}$ ,  $I_{quiescent}$ ,  $V_{OUT}(max.)$ ,  $V_{OUT}(min.)$ .

## TYPICAL REGULATOR CIRCUITS USING THE CA3085 SERIES

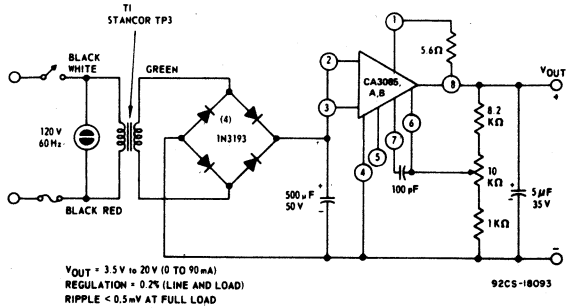


Fig. 17—Application of the CA3085 Series in a typical power supply.

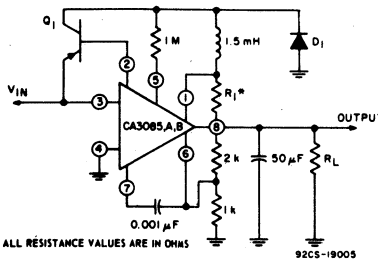


Fig. 18—Typical switching regulator circuit.

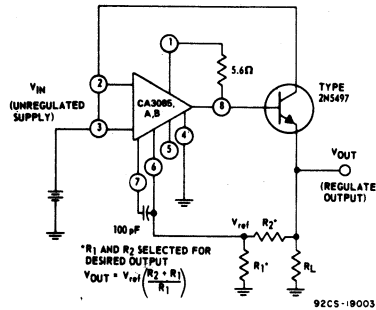


Fig. 19—Typical high-current voltage regulator circuit.

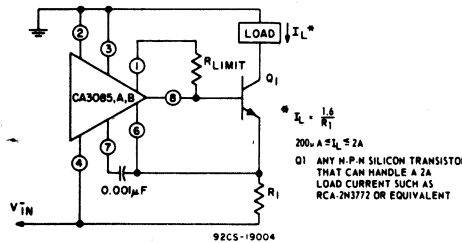


Fig. 20—Typical current regulator circuit.

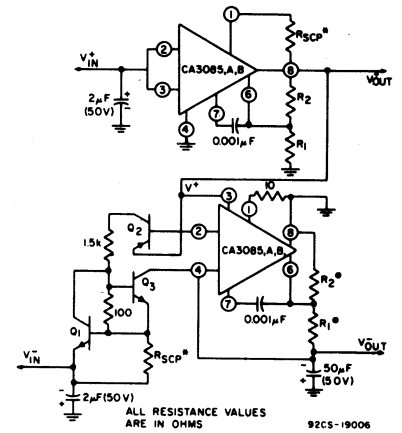


Fig. 21—Combination positive and negative voltage regulator circuit.

# CA3086 General-Purpose N-P-N Transistor Array

Three Isolated Transistors and One Differentially- Connected Transistor Pair

For Low-Power Applications from DC to 120MHz

RCA-CA3086 consists of five general-purpose silicon n-p-n transistors on a common monolithic substrate. Two of the transistors are internally connected to form a differentially-connected pair.

The transistors of the CA3086 are well suited to a wide variety of applications in low-power systems at frequencies from DC to 120 MHz. They may be used as discrete

transistors in conventional circuits. However, they also provide the very significant inherent advantages unique to integrated circuits, such as compactness, ease of physical handling and thermal matching.

The CA3086 is supplied in a 14-lead dual-in line plastic package. The CA3086F is supplied in a 14-lead dual-in-line hermetic (frit-seal) ceramic package.

**Applications**

- General-purpose use in signal processing systems operating in the DC to 120-MHz range
- Temperature compensated amplifiers
- See RCA Application Note, ICAN-5296 "Application of the RCA-CA3018 Integrated-Circuit Transistor Array" for suggested applications.

**MAXIMUM RATINGS, Absolute—Maximum Values at  $T_A = 25^\circ\text{C}$**

**DISSIPATION:**

Any one transistor	300	mW
Total package up to $T_A = 55^\circ\text{C}$	750	mW
Above $T_A = 55^\circ\text{C}$	derate linearly 6.67	mW/ $^\circ\text{C}$

**AMBIENT TEMPERATURE RANGE:**

Operating	-55 to +125	$^\circ\text{C}$
Storage	-65 to +150	$^\circ\text{C}$

**LEAD TEMPERATURE (During soldering):**

At distance 1/16 ± 1/32 inch (1.59 ± 0.79mm)		$^\circ\text{C}$
From case for 10 seconds max.	+265	$^\circ\text{C}$

The following ratings apply for each transistor in the device:

COLLECTOR-TO-EMITTER VOLTAGE, $V_{CE0}$	15	V
COLLECTOR-TO-BASE VOLTAGE, $V_{CBO}$	20	V
COLLECTOR-TO-SUBSTRATE VOLTAGE, $V_{C10}^*$	20	V
EMITTER-TO-BASE VOLTAGE, $V_{EBO}$	5	V
COLLECTOR CURRENT, $I_C$	50	mA

\* The collector of each transistor in the CA3086 is isolated from the substrate by an integral diode. The substrate (terminal 13) must be connected to the most negative point in the external circuit to maintain isolation between transistors and to provide for normal transistor action. To avoid undesirable coupling between transistors, the substrate (terminal 13) should be maintained at either DC or signal (AC) ground. A suitable bypass capacitor can be used to establish a signal ground.

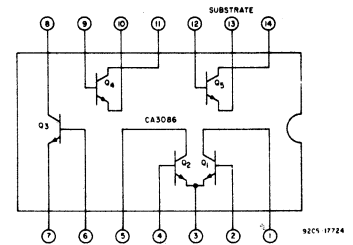


Fig.1 - Functional diagram of the CA3086.

**TYPICAL STATIC CHARACTERISTICS FOR EACH TRANSISTOR**

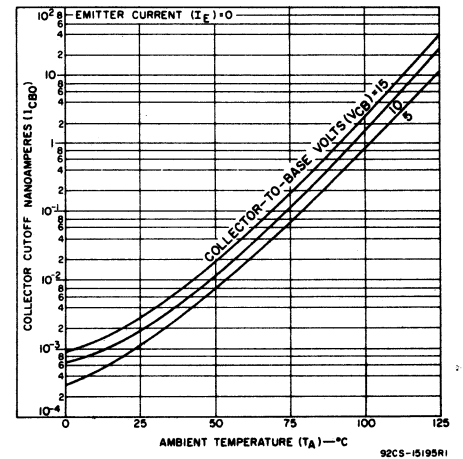


Fig.2 -  $I_{CBO}$  vs  $T_A$ .

**ELECTRICAL CHARACTERISTICS at  $T_A = 25^\circ\text{C}$   
For Equipment Design**

CHARACTERISTICS	SYMBOLS	TEST CONDITIONS	Typ. Characteristic Curves Fig. No.	LIMITS			UNITS
				Min.	Typ.	Max.	
Collector-to-Base Breakdown Voltage	$V_{(BR)CBO}$	$I_C = 10 \mu\text{A}, I_E = 0$	-	20	60	-	V
Collector-to-Emitter Breakdown Voltage	$V_{(BR)CEO}$	$I_C = 1 \text{mA}, I_B = 0$	-	15	24	-	V
Collector-to-Substrate Breakdown Voltage	$V_{(BR)C10}$	$I_C = 10 \mu\text{A}, I_{C1} = 0$	-	20	60	-	V
Emitter-to-Base Breakdown Voltage	$V_{(BR)EBO}$	$I_E = 10 \mu\text{A}, I_C = 0$	-	5	7	-	V
Collector-Cutoff Current	$I_{CBO}$	$V_{CB} = 10 \text{V}, I_E = 0$	2	-	0.002	100	nA
Collector-Cutoff Current	$I_{CEO}$	$V_{CE} = 10 \text{V}, I_B = 0$	3	-	See Curve	5	$\mu\text{A}$
DC Forward-Current Transfer Ratio	$h_{FE}$	$V_{CE} = 3 \text{V}, I_C = 1 \text{mA}$	4	40	100	-	

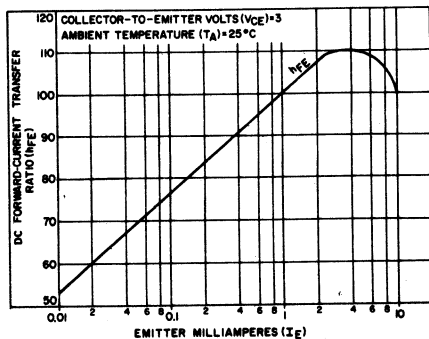


Fig.3 -  $h_{FE}$  vs  $I_E$

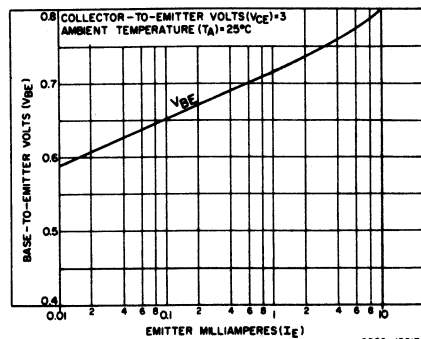


Fig.4 -  $V_{BE}$  vs  $I_E$

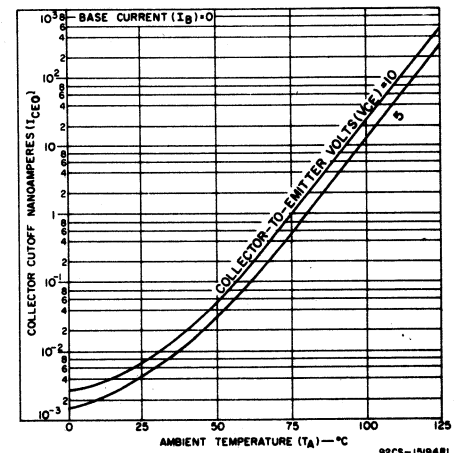


Fig.5 -  $I_{CEO}$  vs  $T_A$ .

# CA3086

ELECTRICAL CHARACTERISTICS at  $T_A = 25^\circ\text{C}$  Typical Values Intended Only for Design Guidance

CHARACTERISTICS	SYMBOL	TEST CONDITIONS		Typ. Characteristics Curves Fig. No.	TYPICAL VALUES	UNITS
		$V_{CE} = 3\text{V}$	$I_C = 10\text{mA}$			
DC Forward-Current Transfer Ratio	$h_{FE}$	$V_{CE} = 3\text{V}$	$I_C = 10\text{mA}$	4	100	
Base-to-Emitter Voltage	$V_{BE}$	$V_{CE} = 3\text{V}$	$I_E = 1\text{mA}$	5	0.715	V
			$I_E = 10\text{mA}$	5	0.800	V
$V_{BE}$ Temperature Coefficient	$\Delta V_{BE}/\Delta T$	$V_{CE} = 3\text{V}, I_C = 1\text{mA}$		6	-1.9	mV/ $^\circ\text{C}$
Collector-to-Emitter Saturation Voltage	$V_{CEsat}$	$I_B = 1\text{mA}, I_C = 10\text{mA}$		-	0.23	V
Noise Figure (low frequency)	NF	$f = 1\text{kHz}, V_{CE} = 3\text{V}, I_C = 100\mu\text{A}, R_S = 1\text{k}\Omega$		-	3.25	dB
Low-Frequency, Small-Signal Equivalent-Circuit Characteristics:						
Forward Current-Transfer Ratio	$h_{fe}$	$f = 1\text{kHz}, V_{CE} = 3\text{V}, I_C = 1\text{mA}$		7	100	-
Short-Circuit Input Impedance	$h_{ie}$			7	3.5	k $\Omega$
Open-Circuit Output Impedance	$h_{oe}$			7	15.6	$\mu\text{mho}$
Open-Circuit Reverse-Voltage Transfer Ratio	$h_{re}$			7	$1.8 \times 10^{-4}$	-
Admittance Characteristics:						
Forward Transfer Admittance	$Y_{fe}$	$f = 1\text{MHz}, V_{CE} = 3\text{V}, I_C = 1\text{mA}$		8	$31 - j1.5$	mmho
Input Admittance	$Y_{ie}$			9	$0.3 + j0.04$	mmho
Output Admittance	$Y_{oe}$			10	$0.001 + j0.03$	mmho
Reverse Transfer Admittance	$Y_{re}$			11	See Curve	-
Gain-Bandwidth Product	$f_T$	$V_{CE} = 3\text{V}, I_C = 3\text{mA}$		12	550	MHz
Emitter-to-Base Capacitance	$C_{EBO}$	$V_{EB} = 3\text{V}, I_E = 0$		-	0.6	pF
Collector-to-Base Capacitance	$C_{CBO}$	$V_{CB} = 3\text{V}, I_C = 0$		-	0.58	pF
Collector-to-Substrate Capacitance	$C_{C10}$	$V_{C1} = 3\text{V}, I_C = 0$		-	2.8	pF

TYPICAL STATIC CHARACTERISTICS FOR EACH TRANSISTOR

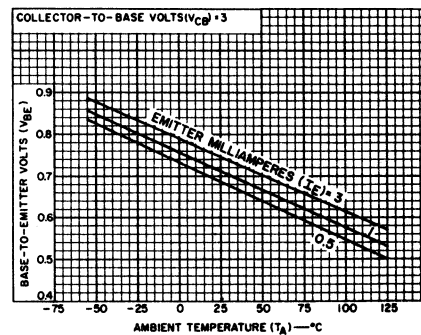


Fig. 6 -  $V_{BE}$  vs  $T_A$ .

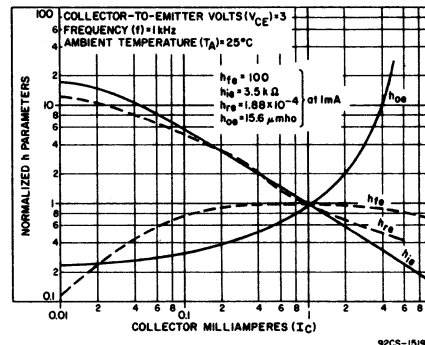


Fig. 7 - Normalized  $h_{fe}$ ,  $h_{ie}$ ,  $h_{oe}$ ,  $h_{re}$  vs  $I_C$ .

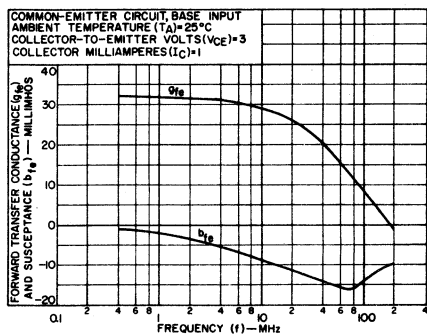


Fig. 8 -  $y_{fe}$  vs  $f$ .

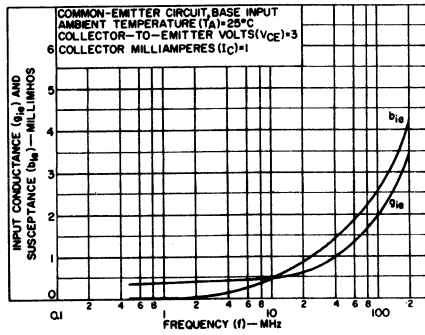


Fig. 9 -  $y_{ie}$  vs  $f$ .

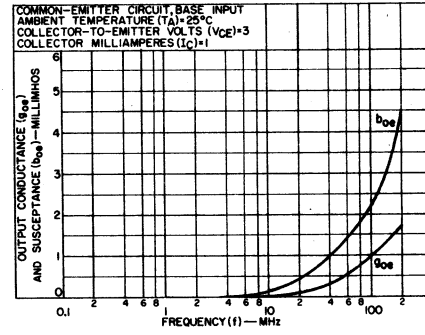


Fig. 10 -  $y_{oe}$  vs  $f$ .

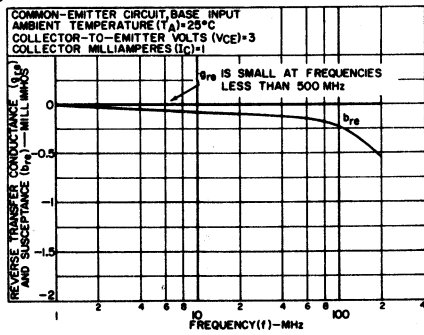


Fig. 11 -  $y_{re}$  vs  $f$ .

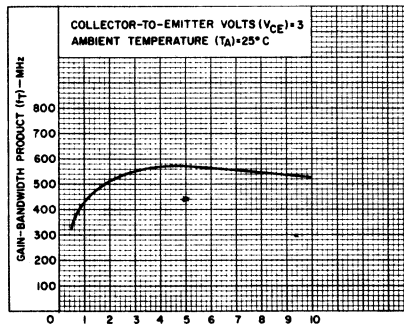


Fig. 12 -  $f_T$  vs  $I_C$ .