

## FEATURES

- Single Supply Operation
- Input Voltage Range Extends to Ground
- Output Swings to Ground while Sinking Current
- Pin Compatible to 1458 and 324 with Precision Specs
- **Guaranteed** Offset Voltage  $150\mu V$  Max.
- **Guaranteed** Low Drift  $2\mu V/\text{ }^{\circ}\text{C}$  Max.
- **Guaranteed** Offset Current  $0.8\text{nA}$  Max.
- **Guaranteed** High Gain
- 5mA Load Current 1.5 Million Min.
- 17mA Load Current 0.8 Million Min.
- **Guaranteed** Low Supply Current  $500\mu A$  Max.
- Low Voltage Noise, 0.1Hz to 10Hz  $0.55\mu V_{\text{p-p}}$
- Low Current Noise—Better than OP-07,  $0.07\text{ pA}/\sqrt{\text{Hz}}$

## APPLICATIONS

- Battery-Powered Precision Instrumentation
- Strain Gauge Signal Conditioners
- Thermocouple Amplifiers
- Instrumentation Amplifiers
- 4mA-20mA Current Loop Transmitters
- Multiple Limit Threshold Detection
- Active Filters
- Multiple Gain Blocks

## Quad Precision Op Amp (LT1014) Dual Precision Op Amp (LT1013)

### DESCRIPTION

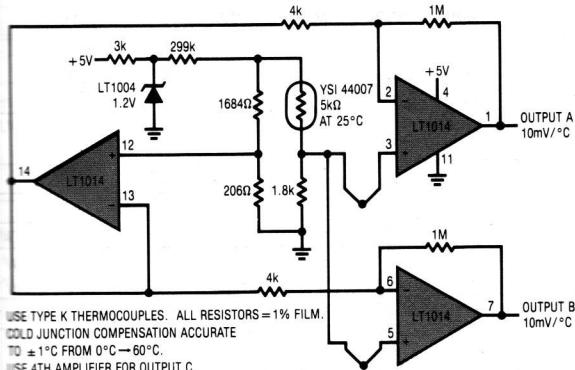
The LT1014 is the first precision quad operational amplifier which directly upgrades designs in the industry standard 14-pin DIP LM324/LM348/OP-11/4156 pin configuration. It is no longer necessary to compromise specifications, while saving board space and cost, as compared to single operational amplifiers.

The LT1014's low offset voltage of  $50\mu V$ , drift of  $0.3\mu V/\text{ }^{\circ}\text{C}$ , offset current of  $0.15\text{nA}$ , gain of 8 million, common-mode rejection of 117dB, and power supply rejection of 120dB qualify it as four truly precision operational amplifiers. Particularly important is the low offset voltage, since no offset null terminals are provided in the quad configuration. Although supply current is only  $350\mu A$  per amplifier, a new output stage design sources and sinks in excess of  $20\text{mA}$  of load current, while retaining high voltage gain.

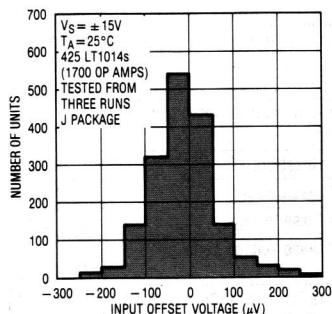
Similarly, the LT1013 is the first precision dual op amp in the 8-pin industry standard configuration, upgrading the performance of such popular devices as the MC1458/1558, LM158 and OP-221. The LT1013's specifications are similar to (even somewhat better than) the LT1014's.

Both the LT1013 and LT1014 can be operated off a single 5V power supply: input common-mode range includes ground; the output can also swing to within a few millivolts of ground. Crossover distortion, so apparent on previous single-supply designs, is eliminated. A full set of specifications is provided with  $\pm 15\text{V}$  and single 5V supplies.

### 3 Channel Thermocouple Thermometer



### LT1014 Distribution of Offset Voltage



**ABSOLUTE MAXIMUM RATINGS**

Supply Voltage	$\pm 22V$
Differential Input Voltage	$\pm 30V$
Input Voltage	Equal to Positive Supply Voltage 5V Below Negative Supply Voltage
Output Short Circuit Duration	Indefinite
Operating Temperature Range	
LT1013AM / LT1013M	
LT1014AM / LT1014M	-55°C to 125°C
LT1013AC / 1013C / 1013D	
LT1014AC / 1014C / 1014D	0°C to 70°C
Storage Temperature Range	
All Grades	-65°C to 150°C
Lead Temperature (Soldering, 10 sec.)	300°C

**PACKAGE/ORDER INFORMATION**

ORDER PART NUMBER
LT1013AMH
LT1013MH
LT1013ACH
LT1013CH
LT1013AMJ8
LT1013MJ8
LT1013ACJ8
LT1013CJ8
LT1013CN8
LT1013DN8
LT1014AMJ
LT1014MJ
LT1014ACJ
LT1014CJ
LT1014CN
LT1014DN

**ELECTRICAL CHARACTERISTICS** $V_S = \pm 15V$ ,  $V_{CM} = 0V$ ,  $T_A = 25^\circ C$  unless otherwise noted

SYMBOL	PARAMETER	CONDITIONS	LT1013AM			LT1013M / LT1013C			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{OS}$	Input Offset Voltage	LT1013 LT1014 LT1013DN8 / LT1014DN	—	40	150	—	60	300	$\mu V$
			—	50	180	—	60	300	$\mu V$
			—	—	—	—	200	800	$\mu V$
	Long Term Input Offset Voltage Stability		—	0.4	—	—	0.5	—	$\mu V / Mo.$
$I_{OS}$	Input Offset Current		—	0.15	0.8	—	0.2	1.5	nA
$I_B$	Input Bias Current		—	12	20	—	15	30	nA
$e_n$	Input Noise Voltage	0.1Hz to 10Hz	—	0.55	—	—	0.55	—	$\mu V_{p-p}$
$e_n$	Input Noise Voltage Density	$f_0 = 10Hz$	—	24	—	—	24	—	$nV/\sqrt{Hz}$
		$f_0 = 1000Hz$	—	22	—	—	22	—	$nV/\sqrt{Hz}$
$i_n$	Input Noise Current Density	$f_0 = 10Hz$	—	0.07	—	—	0.07	—	$pA/\sqrt{Hz}$
		(Note 1)	100	400	—	70	300	—	MΩ
	Input Resistance—Differential Common-Mode		—	5	—	—	4	—	GΩ

**ELECTRICAL CHARACTERISTICS** $V_S = \pm 15V$ ,  $V_{CM} = 0V$ ,  $T_A = 25^\circ C$  unless otherwise noted

SYMBOL	PARAMETER	CONDITIONS	LT1013AM LT1013AC LT1014AM LT1014AC			LT1013M/LT1013C LT1013DN8 LT1014M/LT1014C LT1014DN			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	
$A_{VOL}$	Large Signal Voltage Gain	$V_0 = \pm 10V$ , $R_L = 2k$ $V_0 = \pm 10V$ , $R_L = 600\Omega$	1.5 0.8	8.0 2.5	— —	1.2 0.5	7.0 2.0	— —	$V/\mu V$ $V/\mu V$
$V_{IN}$	Input Voltage Range		+13.5 -15.0	+13.8 -15.3	— —	+13.5 -15.0	+13.8 -15.3	— —	V V
$CMRR$	Common-Mode Rejection Ratio	$V_{CM} = +13.5V$ , $-15.0V$	100	117	—	97	114	—	dB
$PSRR$	Power Supply Rejection Ratio	$V_S = \pm 2V$ to $\pm 18V$	103	120	—	100	117	—	dB
$V_{CS}$	Channel Separation	$V_0 = \pm 10V$ , $R_L = 2k$	123	140	—	120	137	—	dB
$V_{OUT}$	Output Voltage Swing	$R_L = 2k$	$\pm 13$	$\pm 14$	—	$\pm 12.5$	$\pm 14$	—	V
$S$	Slew Rate		0.2	0.4	—	0.2	0.4	—	$V/\mu s$
$I_S$	Supply Current	Per Amplifier	—	0.35	0.50	—	0.35	0.55	mA

Note 1: This parameter is guaranteed by design and is not tested. Typical parameters are defined as the 60% yield of parameter distributions of individual amplifiers; i.e., out of 100 LT1014s (or 100 LT1013s) typically 240 op amps (or 120) will be better than the indicated specification.

**ELECTRICAL CHARACTERISTICS** $V_S^+ = +5V$ ,  $V_S^- = 0V$ ,  $V_{OUT} = 1.4V$ ,  $V_{CM} = 0V$ ,  $T_A = 25^\circ C$  unless otherwise noted

SYMBOL	PARAMETER	CONDITIONS	LT1013AM LT1013AC LT1014AM LT1014AC			LT1013M/LT1013C LT1013DN8 LT1014M/LT1014C LT1014DN			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{OS}$	Input Offset Voltage	LT1013 LT1014 LT1013DN8/LT1014DN	— — —	60 70 —	250 280 —	— — —	90 90 250	450 450 950	$\mu V$ $\mu V$ $\mu V$
$I_{OS}$	Input Offset Current		—	0.2	1.3	—	0.3	2.0	nA
$I_B$	Input Bias Current		—	15	35	—	18	50	nA
$A_{VOL}$	Large Signal Voltage Gain	$V_0 = 5mV$ to $4V$ , $R_L = 500\Omega$	—	1.0	—	—	1.0	—	$V/\mu V$
$V_{IN}$	Input Voltage Range		+3.5 0	+3.8 -0.3	— —	+3.5 0	+3.8 -0.3	— —	V V
$V_{OUT}$	Output Voltage Swing	Output Low, No Load Output Low, $600\Omega$ to Ground Output Low, $I_{SINK} = 1mA$ Output High, No Load Output High, $600\Omega$ to Ground	— — — 4.0 3.4	15 5 220 4.4 4.0	25 10 350 — —	— — — 4.0 3.4	15 5 220 4.4 4.0	25 10 350 — —	mV mV mV V V
$I_S$	Supply Current	Per Amplifier	—	0.31	0.45	—	0.32	0.50	mA

# LT1013/LT1014

## ELECTRICAL CHARACTERISTICS $V_S = \pm 15V, V_{CM} = 0V, -55^\circ C \leq T_A \leq 125^\circ C$ unless otherwise noted

SYMBOL	PARAMETER	CONDITIONS	LT1013AM			LT1014AM			LT1013M/LT1014M			UNITS	
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX		
$V_{OS}$	Input Offset Voltage	$V_S = +5V, 0V; V_0 = +1.4V$ $-55^\circ C \leq T_A \leq 100^\circ C$ $V_{CM} = 0.1V, T_A = 125^\circ C$ $V_S = 0V, T_A = 125^\circ C$	●	—	80	300	—	90	350	—	110	550	$\mu V$
			●	—	80	450	—	90	480	—	100	750	$\mu V$
			●	—	120	450	—	150	480	—	200	750	$\mu V$
			●	—	250	900	—	300	960	—	400	1500	$\mu V$
$I_{OS}$	Input Offset Current	$V_S = +5V, 0V; V_0 = +1.4V$	●	—	0.4	2.0	—	0.4	2.0	—	0.5	2.5	$\mu A/^\circ C$
			●	—	0.3	2.5	—	0.3	2.8	—	0.4	5.0	$nA$
			●	—	0.6	6.0	—	0.7	7.0	—	0.9	10.0	$nA$
$I_B$	Input Bias Current	$V_S = +5V, 0V; V_0 = +1.4V$	●	—	15	30	—	15	30	—	18	45	$nA$
			●	—	20	80	—	25	90	—	28	120	$nA$
$A_{VOL}$	Large Signal Voltage Gain	$V_0 = \pm 10V, R_L = 2k$	●	0.5	2.0	—	0.4	2.0	—	0.25	2.0	—	$V/\mu V$
CMRR	Common-Mode Rejection	$V_{CM} = +13.0V, -14.9V$	●	97	114	—	96	114	—	94	113	—	dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 2V$ to $\pm 18V$	●	100	117	—	100	117	—	97	116	—	dB
$V_{OUT}$	Output Voltage Swing	$R_L = 2k$ $V_S = +5V, 0V;$ $R_L = 600\Omega$ to Ground Output Low Output High	●	$\pm 12$	$\pm 13.8$	—	$\pm 12$	$\pm 13.8$	—	$\pm 11.5$	$\pm 13.8$	—	V
			●	—	6	15	—	6	15	—	6	18	$mV$
			●	3.2	3.8	—	3.2	3.8	—	3.1	3.8	—	V
$I_S$	Supply Current per Amplifier	$V_S = +5V, 0V; V_0 = +1.4V$	●	—	0.38	0.60	—	0.38	0.60	—	0.38	0.7	$mA$
			●	—	0.34	0.55	—	0.34	0.55	—	0.34	0.65	$mA$

## ELECTRICAL CHARACTERISTICS $V_S = \pm 15V, V_{CM} = 0V, 0^\circ C \leq T_A \leq 70^\circ C$ unless otherwise noted

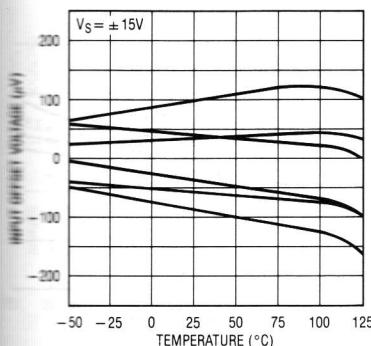
SYMBOL	PARAMETER	CONDITIONS	LT1013AC			LT1014AC			LT1013C/LT1013DN8 LT1014C/LT1014DN			UNITS	
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX		
$V_{OS}$	Input Offset Voltage	$LT1013DN8, LT1014DN$ $V_S = +5V, 0V; V_0 = 1.4V$ $LT1013DN, LT1014DN$	●	—	55	240	—	65	270	—	80	400	$\mu V$
			●	—	—	—	—	—	—	—	230	1000	$\mu V$
			●	—	75	350	—	85	380	—	110	570	$\mu V$
			●	—	—	—	—	—	—	—	280	1200	$\mu V$
$A_{VOL}$	Average Input Offset Voltage Drift	(Note 2) $LT1013DN, LT1014DN$	●	—	0.3	2.0	—	0.3	2.0	—	0.4	2.5	$\mu V/^\circ C$
			●	—	—	—	—	—	—	—	0.7	5.0	$\mu V/^\circ C$
$I_{OS}$	Input Offset Current	$V_S = +5V, 0V; V_0 = 1.4V$	●	—	0.2	1.5	—	0.2	1.7	—	0.3	2.8	$nA$
			●	—	0.4	3.5	—	0.4	4.0	—	0.5	6.0	$nA$
$I_B$	Input Bias Current	$V_S = +5V, 0V; V_0 = 1.4V$	●	—	13	25	—	13	25	—	16	38	$nA$
			●	—	18	55	—	20	60	—	24	90	$nA$
$A_{VOL}$	Large Signal Voltage Gain	$V_0 = \pm 10V, R_L = 2k$	●	1.0	5.0	—	1.0	5.0	—	0.7	4.0	—	$V/\mu V$
CMRR	Common-Mode Rejection Ratio	$V_{CM} = +13.0V, -15.0V$	●	98	116	—	98	116	—	94	113	—	dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 2V$ to $\pm 18V$	●	101	119	—	101	119	—	97	116	—	dB
$V_{OUT}$	Output Voltage Swing	$R_L = 2k$ $V_S = +5V, 0V; R_L = 600\Omega$ Output Low Output High	●	$\pm 12.5$	$\pm 13.9$	—	$\pm 12.5$	$\pm 13.9$	—	$\pm 12.0$	$\pm 13.9$	—	V
			●	—	6	13	—	6	13	—	6	13	$mV$
			●	3.3	3.9	—	3.3	3.9	—	3.2	3.9	—	V
$I_S$	Supply Current per Amplifier	$V_S = +5V, 0V; V_0 = 1.4V$	●	—	0.36	0.55	—	0.36	0.55	—	0.37	0.60	$mA$
			●	—	0.32	0.50	—	0.32	0.50	—	0.34	0.55	$mA$

Note 2: This parameter is not 100% tested.

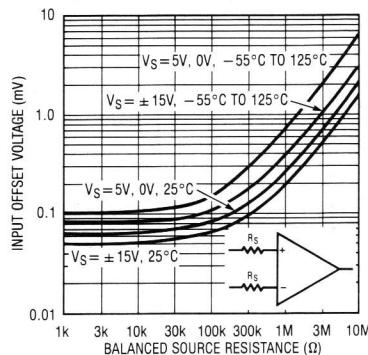
The ● denotes the specifications which apply over the full operating temperature range.

## TYPICAL PERFORMANCE CHARACTERISTICS

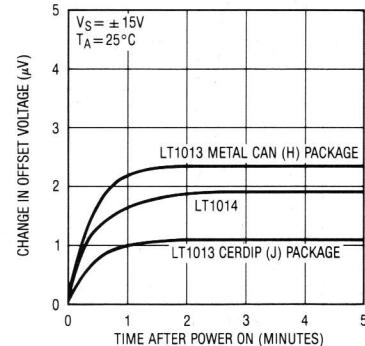
**Offset Voltage Drift with Temperature of Representative Units**



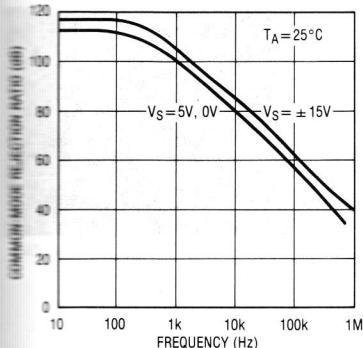
**Offset Voltage vs Balanced Source Resistance**



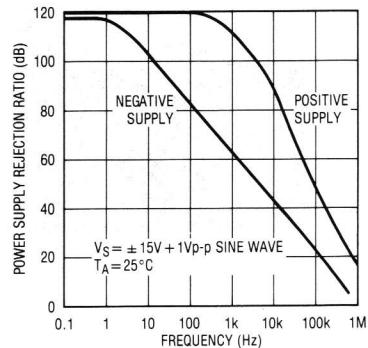
**Warm-Up Drift**



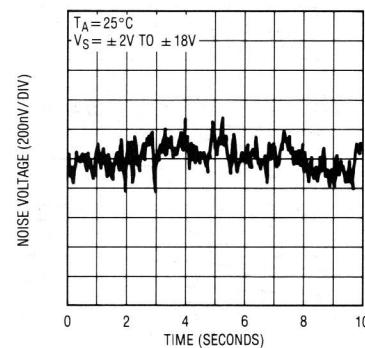
**Common-Mode Rejection Ratio vs Frequency**



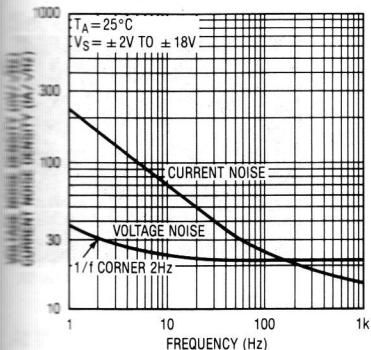
**Power Supply Rejection Ratio vs Frequency**



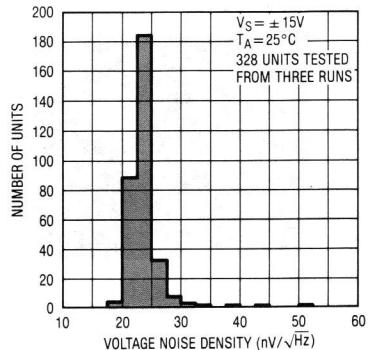
**0.1Hz to 10Hz Noise**



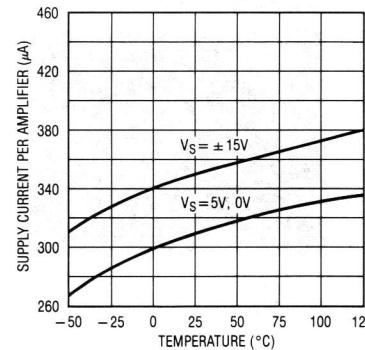
**Noise Spectrum**



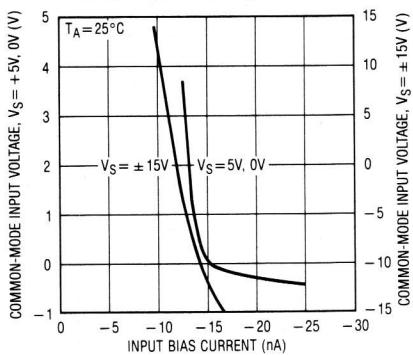
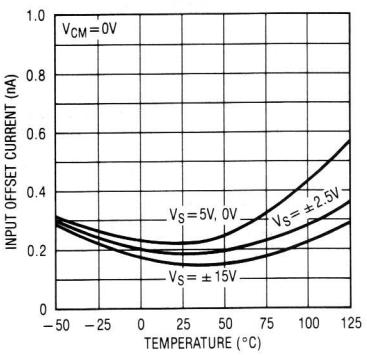
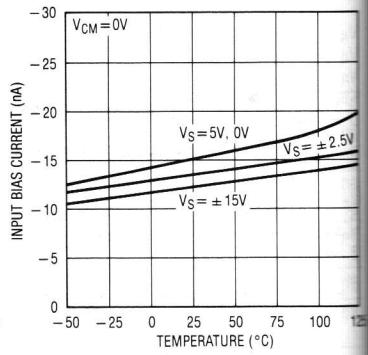
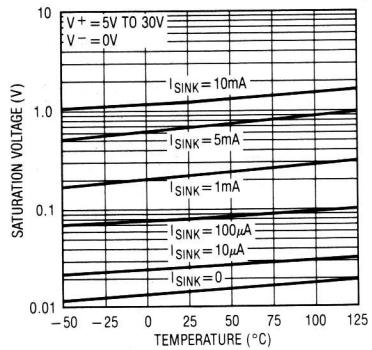
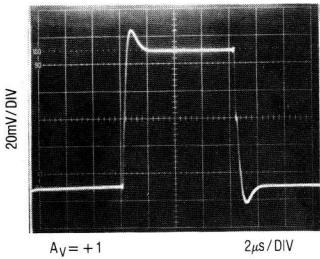
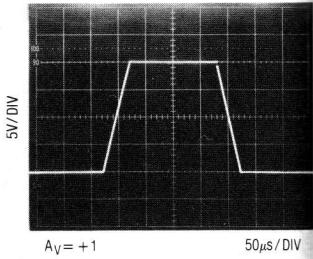
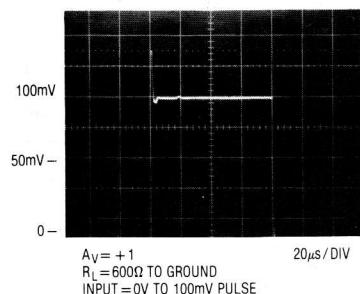
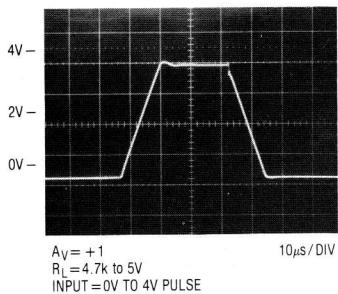
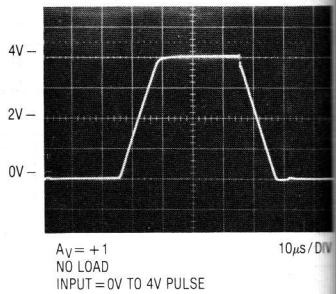
**10Hz Voltage Noise Distribution**



**Supply Current vs Temperature**

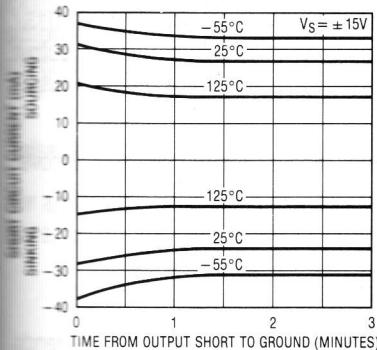


## TYPICAL PERFORMANCE CHARACTERISTICS

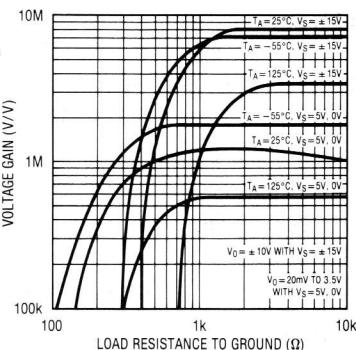
**Input Bias Current vs Common-Mode Voltage****Input Offset Current vs Temperature****Input Bias Current vs Temperature****Output Saturation vs Sink Current vs Temperature****Small Signal Transient Response,  $V_S = \pm 15V$** **Large Signal Transient Response,  $V_S = \pm 15V$** **Small Signal Transient Response,  $V_S = 5V, 0V$** **Large Signal Transient Response,  $V_S = 5V, 0V$** **Large Signal Transient Response,  $V_S = 5V, 0V$** 

## TYPICAL PERFORMANCE CHARACTERISTICS

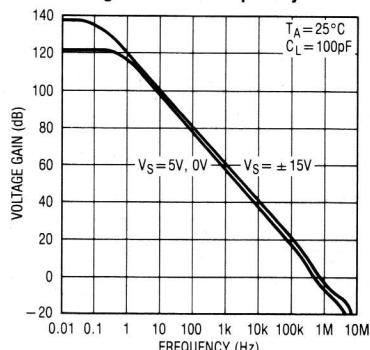
**Output Short Circuit Current vs Time**



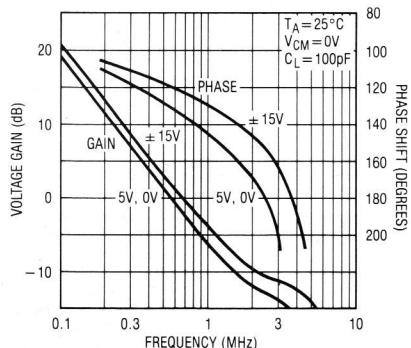
**Voltage Gain vs Load Resistance**



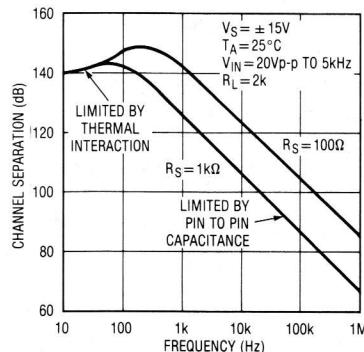
**Voltage Gain vs Frequency**



**Gain, Phase vs Frequency**



**Channel Separation vs Frequency**



## APPLICATIONS INFORMATION

### Single Supply Operation

The LT1013/1014 are fully specified for single supply operation, i.e., when the negative supply is 0V. Input common-mode range includes ground; the output swings within a few millivolts of ground. Single supply operation, however, can create special difficulties, both at the input and at the output. The LT1013/1014 have specific circuitry which addresses these problems.

At the input, the driving signal can fall below 0V— inadvertently or on a transient basis. If the input is more than

a few hundred millivolts below ground, two distinct problems can occur on previous single supply designs, such as the LM124, LM158, OP-20, OP-21, OP-220, OP-221, OP-420:

- When the input is more than a diode drop below ground, unlimited current will flow from the substrate ( $V^-$  terminal) to the input. This can destroy the unit. On the LT1013/1014, the  $400\Omega$  resistors, in series with the input (see schematic diagram), protect the devices even when the input is 5V below ground.

## APPLICATIONS INFORMATION

(b) When the input is more than 400mV below ground (at 25°C), the input stage saturates (transistors Q3 and Q4) and phase reversal occurs at the output. This can cause lock-up in servo systems. Due to a unique phase reversal protection circuitry (Q21, Q22, Q27, Q28), the LT1013/1014's outputs do not reverse, as illustrated below, even when the inputs are at -1.5V.

There is one circumstance, however, under which the phase reversal protection circuitry does not function: when the other op amp on the LT1013, or one specific amplifier of the other three on the LT1014, is driven hard into negative saturation at the output.

- P Phase reversal protection does not work on amplifier: A when D's output is in negative saturation. B's and C's outputs have no effect.
- B when C's output is in negative saturation. A's and D's outputs have no effect.
- C when B's output is in negative saturation. A's and D's outputs have no effect.
- D when A's output is in negative saturation. B's and C's outputs have no effect.

At the output, the aforementioned single supply designs either cannot swing to within 600mV of ground (OP-20) or cannot sink more than a few microamperes while swinging to ground (LM124, LM158). The LT1013/1014's all-NPN output stage maintains its low output resistance and high gain characteristics until the output is saturated.

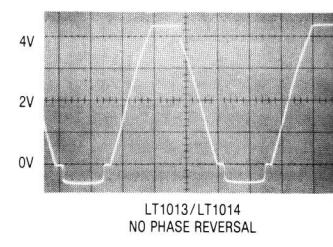
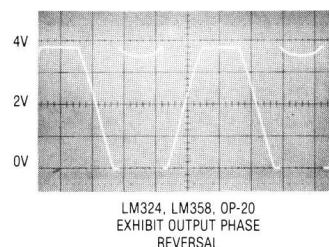
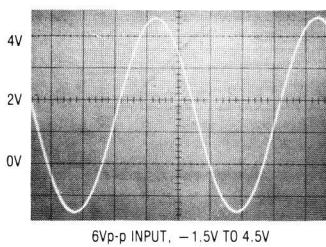
In dual supply operations, the output stage is crossover distortion-free.

### Comparator Applications

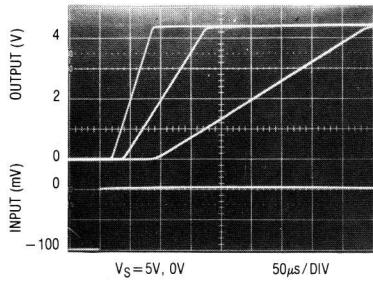
The single supply operation of the LT1013/1014 lends itself to its use as a precision comparator with TTL compatible output:

In systems using both op amps and comparators, the LT1013/1014 can perform multiple duties; for example, on the LT1014, two of devices can be used as op amps and the other two as comparators.

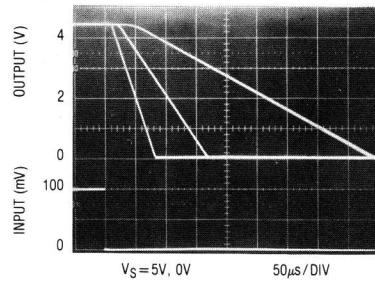
**Voltage Follower with Input Exceeding the Negative Common-Mode Range**



**Comparator Rise Response Time  
10mV, 5mV, 2mV Overdrives**



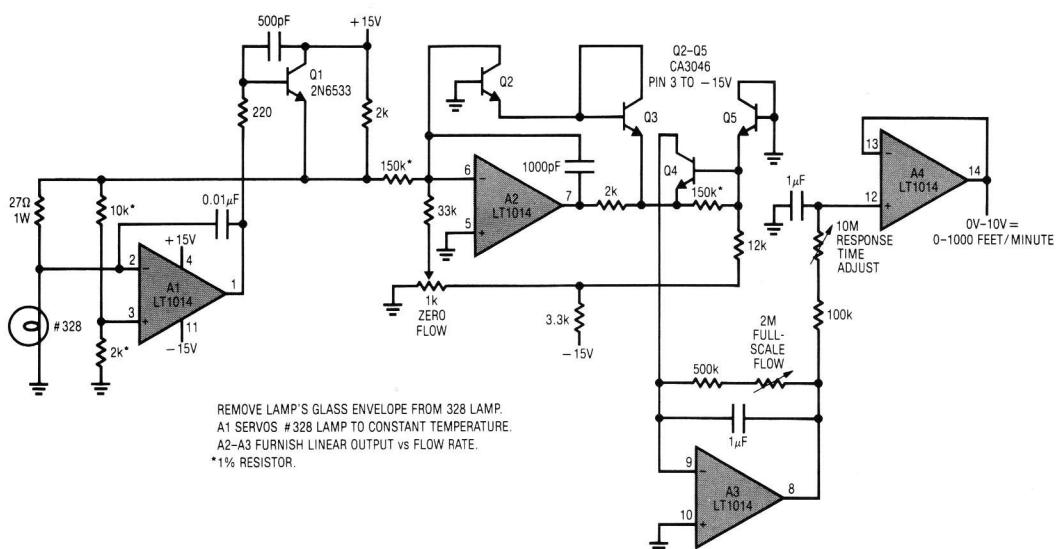
**Comparator Fall Response Time  
to 10mV, 5mV, 2mV Overdrives**



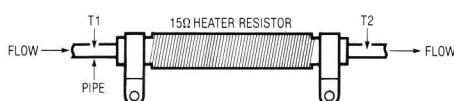
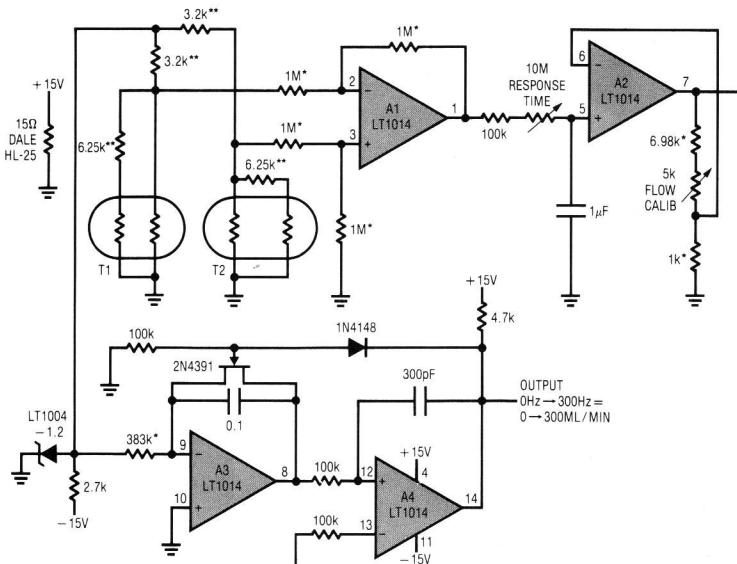


## TYPICAL APPLICATIONS

## Hot Wire Anemometer



## Liquid Flowmeter



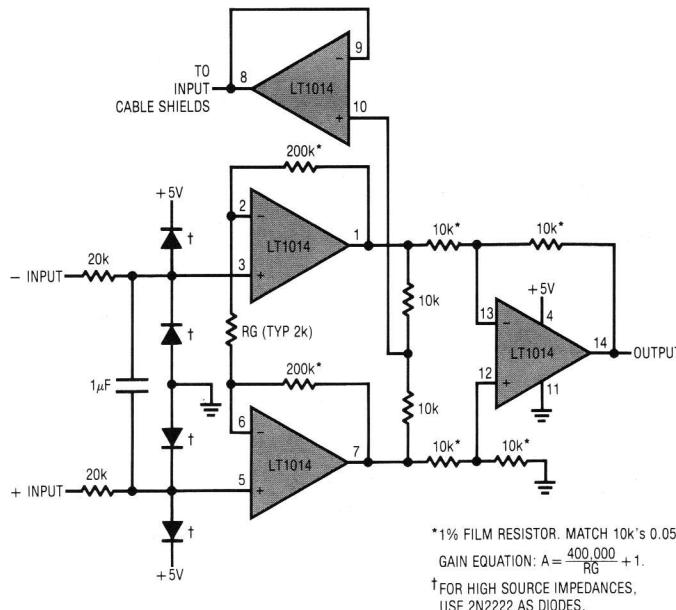
\*1% FILM RESISTOR.

\*\*SUPPLIED WITH YSI THERMISTOR NETWORK.

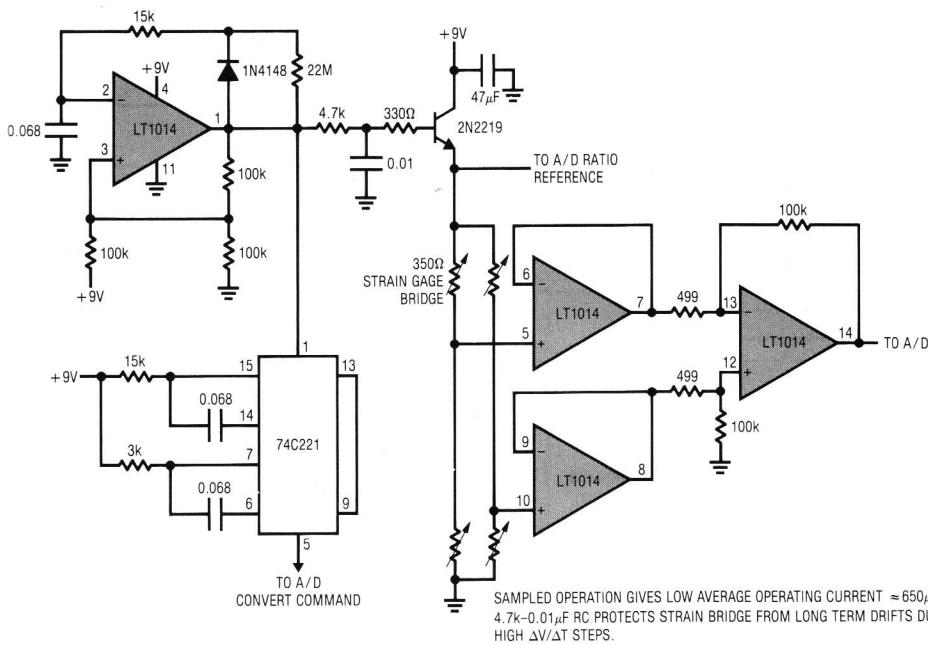
T1, T2 YSI THERMISTOR NETWORK = #44201.  
FLOW IN PIPE IS INVERSELY PROPORTIONAL TO  
RESISTANCE OF T1-T2 TEMPERATURE DIFFERENCE.  
A1-A2 PROVIDE GAIN. A3-A4 PROVIDE LINEARIZED  
FREQUENCY OUTPUT.

## **TYPICAL APPLICATIONS**

## 5V Powered Precision Instrumentation Amplifier

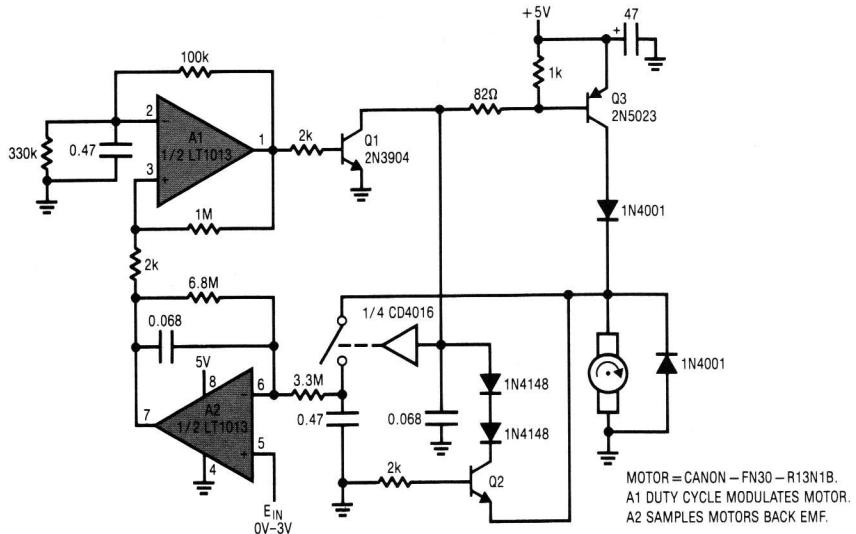


## **9V Battery Powered Strain Gage Signal Conditioner**

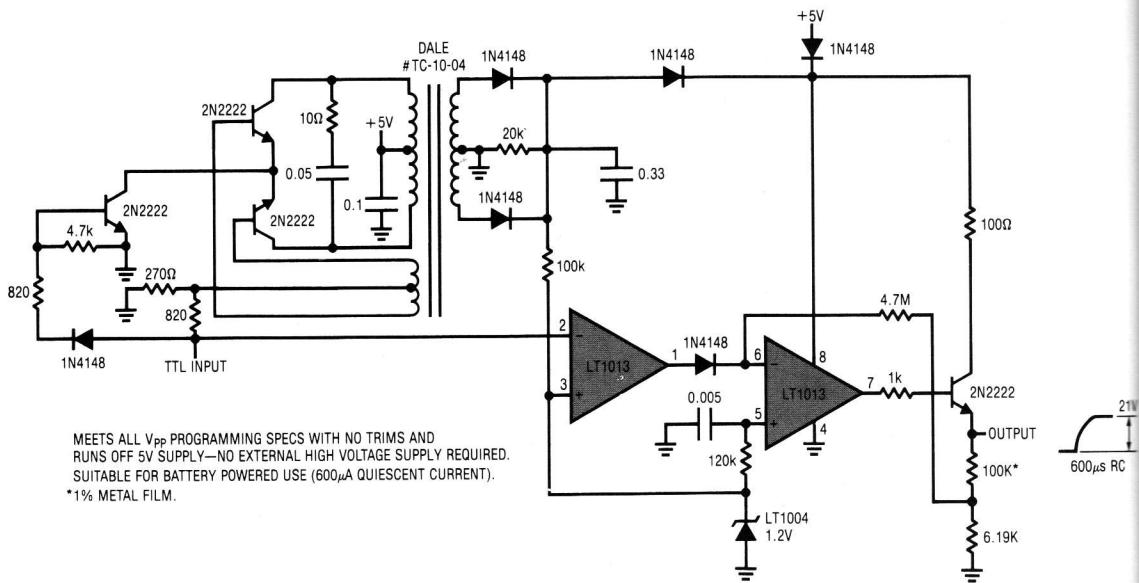


## **TYPICAL APPLICATIONS**

## **5V Powered Motor Speed Controller No Tachometer Required**

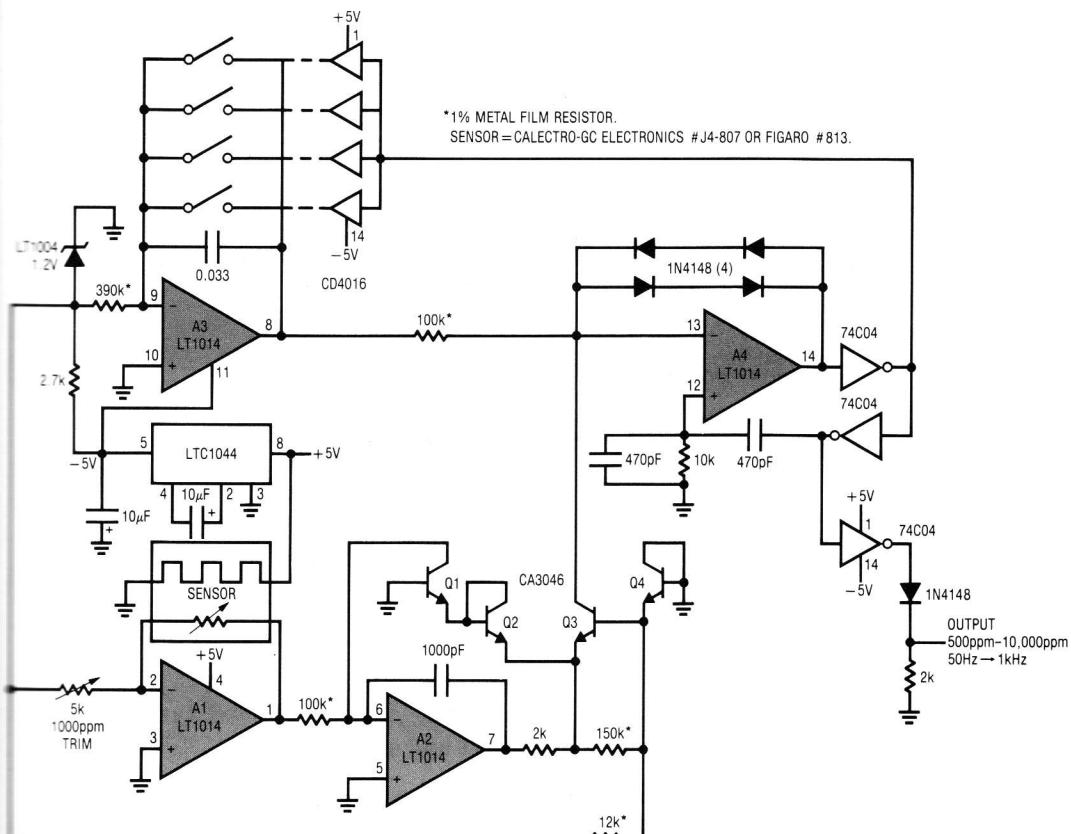


## **5V Powered EEPROM Pulse Generator**

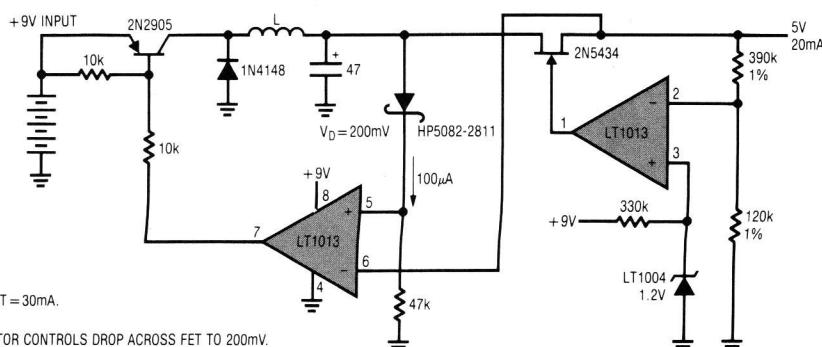


## TYPICAL APPLICATIONS

## Methane Concentration Detector with Linearized Output

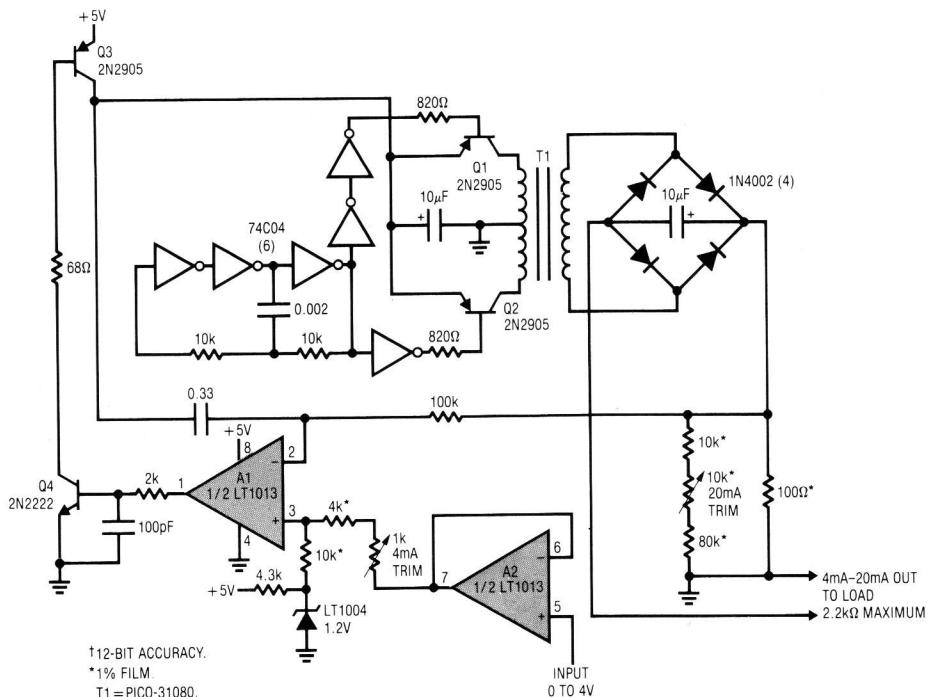


## Low Power 9V to 5V Converter

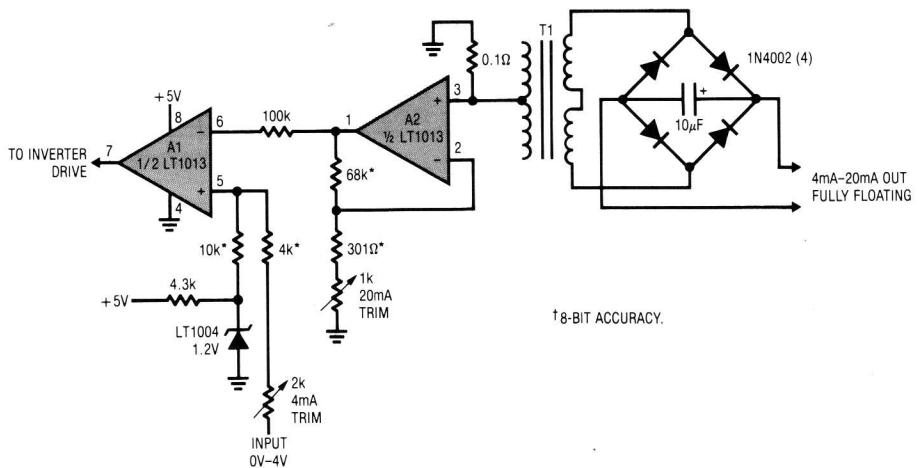


## TYPICAL APPLICATIONS

## 5V Powered 4mA-20mA Current Loop Transmitter †

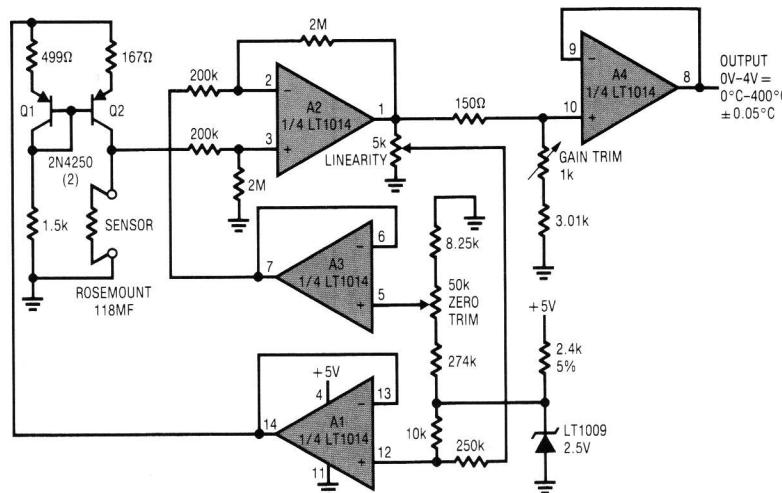


## Fully Floating Modification to 4mA-20mA Current Loop †

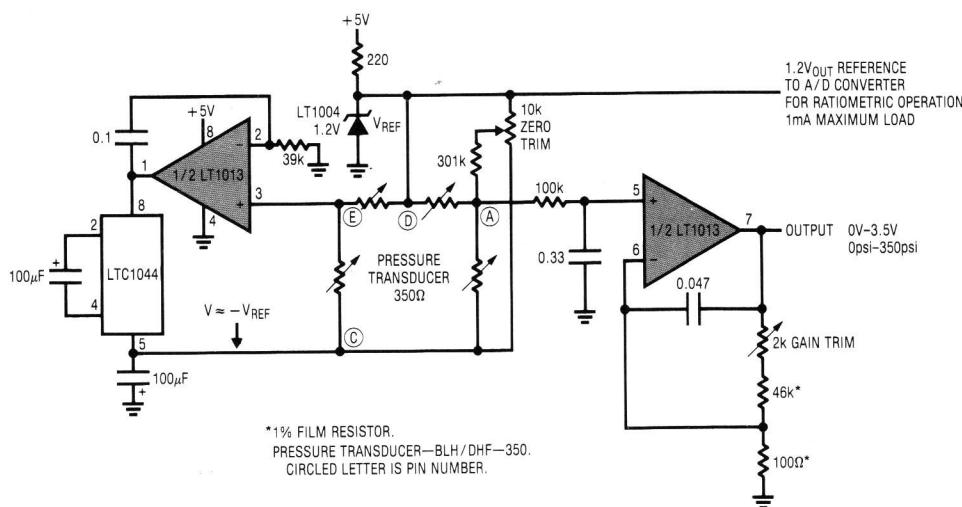


## TYPICAL APPLICATIONS

## 5V Powered, Linearized Platinum RTD Signal Conditioner

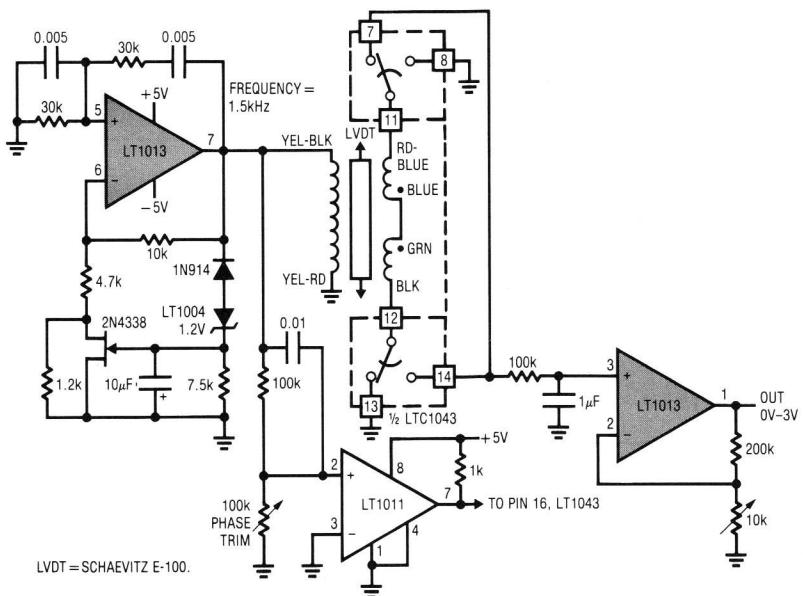


## Strain Gage Bridge Signal Conditioner

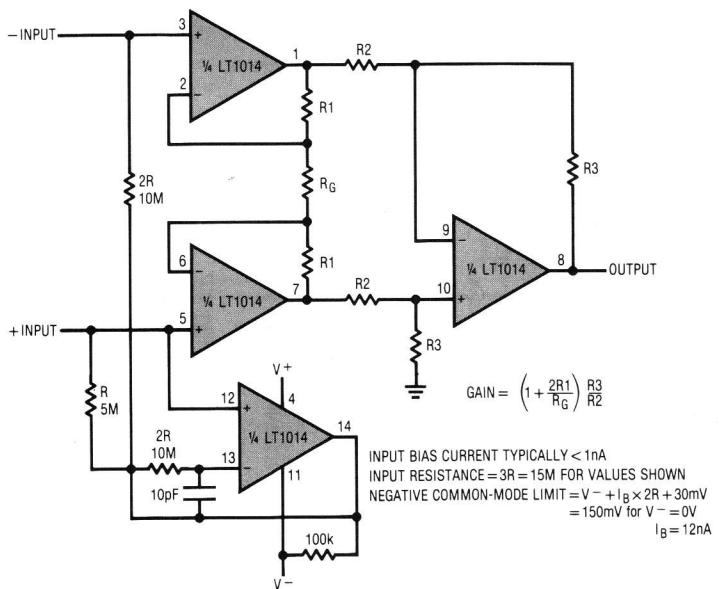


## **TYPICAL APPLICATIONS**

## LVDT Signal Conditioner

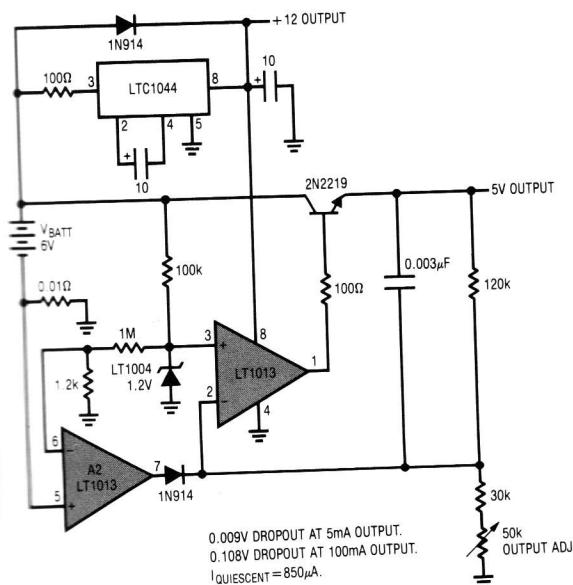


## **Triple Op Amp Instrumentation Amplifier with Bias Current Cancellation**

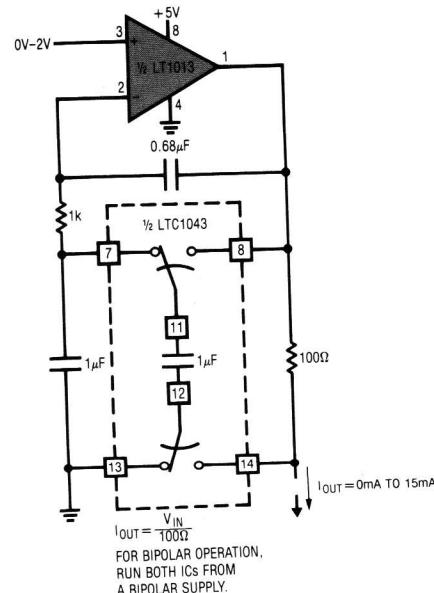
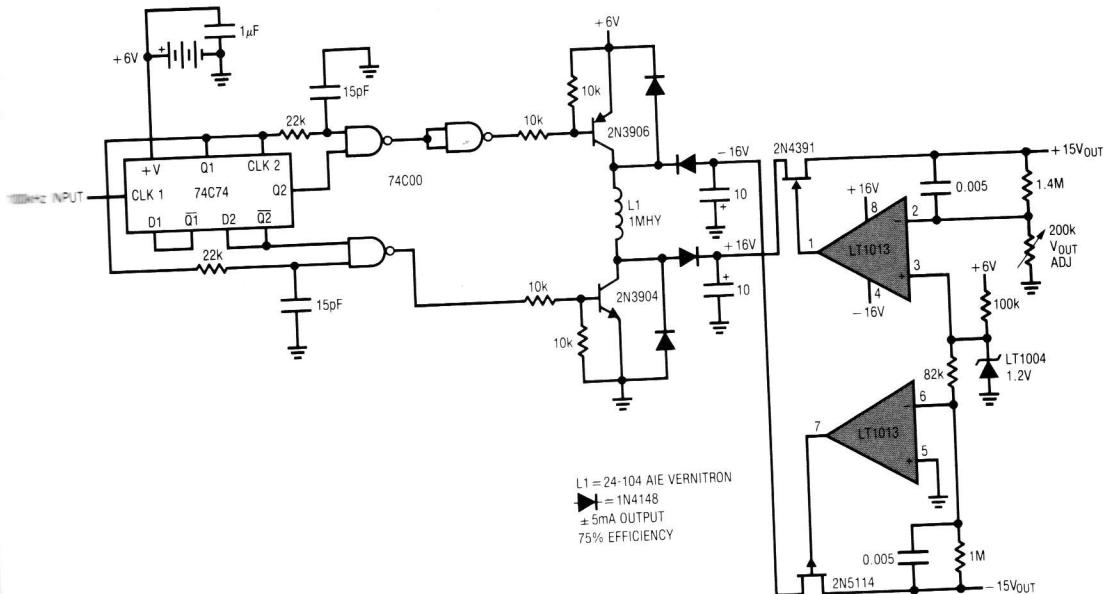


## TYPICAL APPLICATIONS

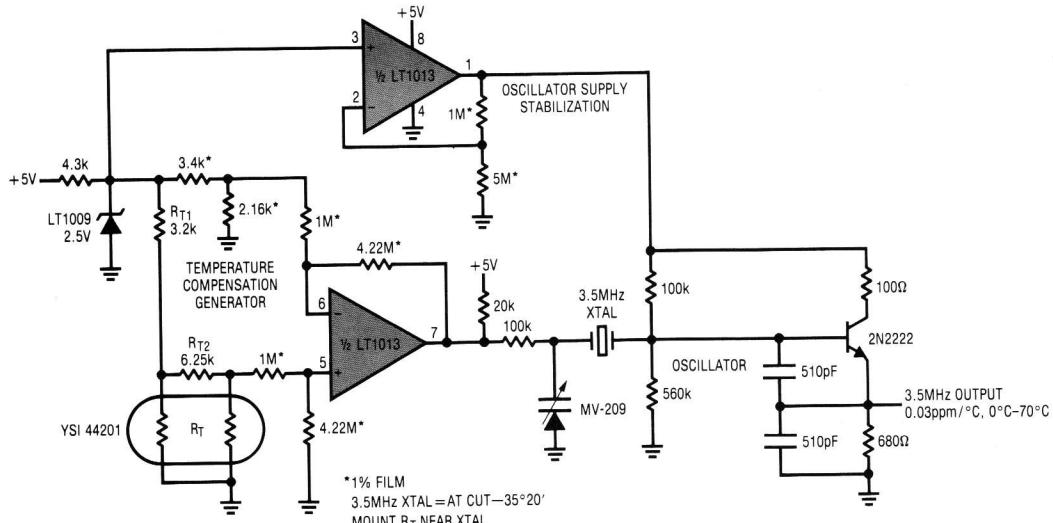
## Low Dropout Regulator for 6V Battery



## Voltage Controlled Current Source with Ground Referred Input and Output

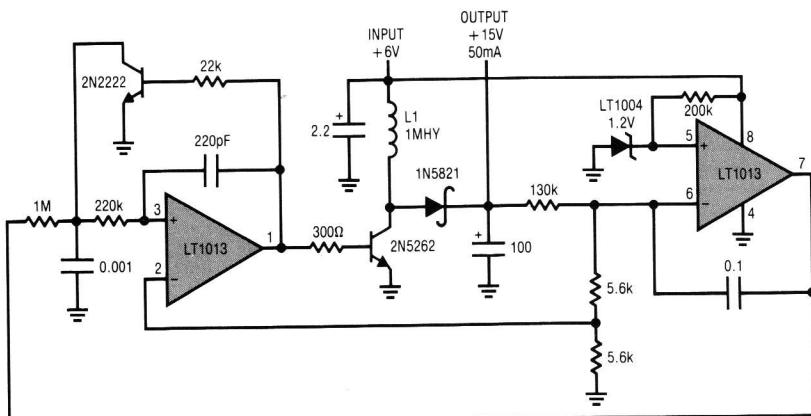
6V to  $\pm 15$ V Regulating Converter

## TYPICAL APPLICATIONS

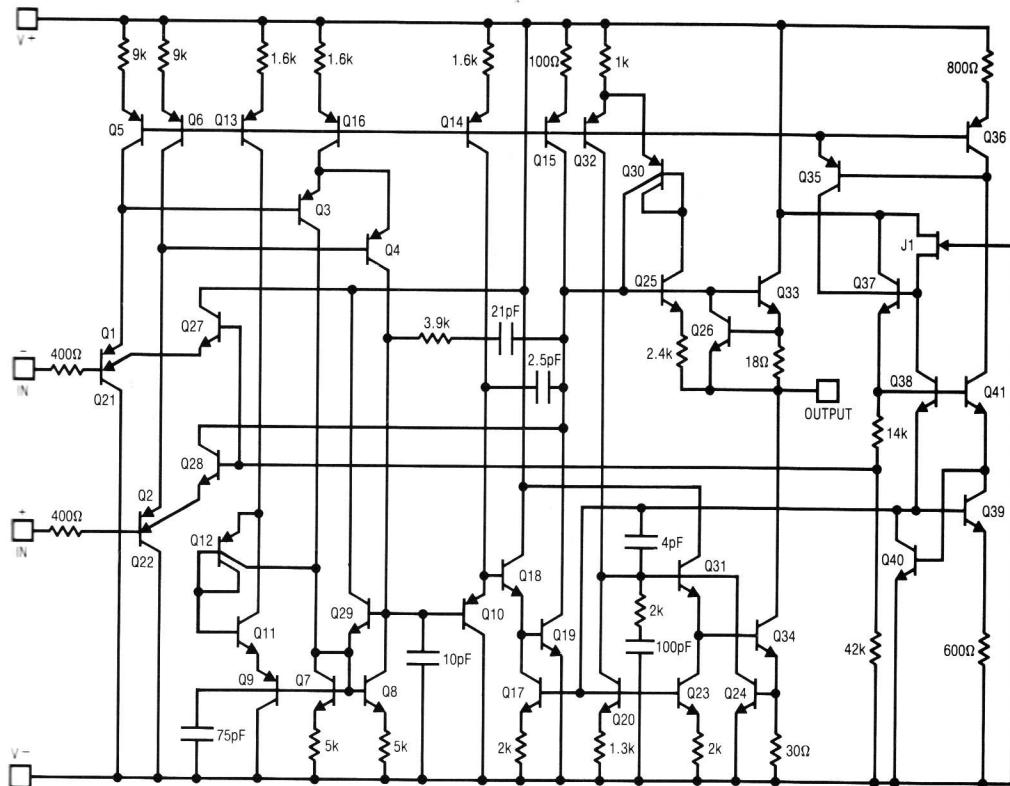
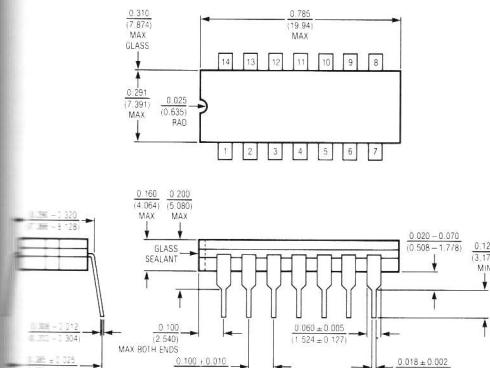
Low Power, 5V Driven, Temperature Compensated Crystal Oscillator (TXCO)<sup>†</sup>

\*1% FILM  
 3.5MHz XTAL = AT CUT -35°20'  
 MOUNT R<sub>T</sub> NEAR XTAL  
 3mA POWER DRAIN  
<sup>†</sup>THERMISTOR-AMPLIFIER-VARACTOR NETWORK GENERATES  
 A TEMPERATURE COEFFICIENT OPPOSITE THE CRYSTAL TO  
 MINIMIZE OVERALL OSCILLATOR DRIFT

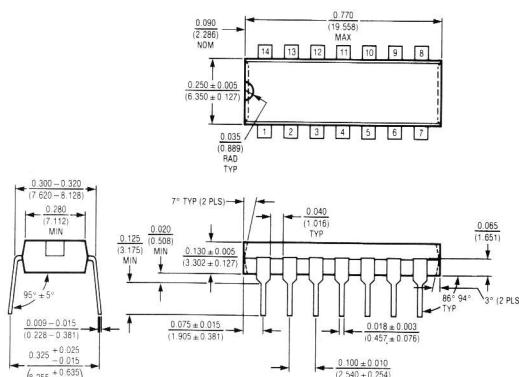
## Step-Up Switching Regulator for 6V Battery



L1 = AIE—VERNITRON 24-104  
 78% EFFICIENCY

**SCHEMATIC DIAGRAM****1/2 LT1013, 1/4 LT1014****PACKAGE DESCRIPTION****J Package 14 Lead Cavity DIP**

$T_{j\max}$	$\Theta_{JA}$
150°C	100°C/W

**N Package 14 Lead Molded DIP**

$T_{j\max}$	$\Theta_{JA}$
100°C	100°C/W