

μA741

FREQUENCY-COMPENSATED OPERATIONAL AMPLIFIER

FAIRCHILD LINEAR INTEGRATED CIRCUIT

GENERAL DESCRIPTION — The μA741 is a high performance monolithic Operational Amplifier constructed using the Fairchild Planar* epitaxial process. It is intended for a wide range of analog applications. High common mode voltage range and absence of latch-up tendencies make the μA741 ideal for use as a voltage follower. The high gain and wide range of operating voltage provides superior performance in integrator, summing amplifier, and general feedback applications. Electrical characteristics of the μA741A and E are identical to MIL-M-38510/10101.

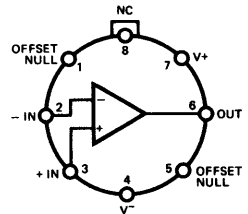
- NO FREQUENCY COMPENSATION REQUIRED
- SHORT CIRCUIT PROTECTION
- OFFSET VOLTAGE NULL CAPABILITY
- LARGE COMMON MODE AND DIFFERENTIAL VOLTAGE RANGES
- LOW POWER CONSUMPTION
- NO LATCH-UP

ABSOLUTE MAXIMUM RATINGS

Supply Voltage	
μA741A, μA741, μA741E	±22 V
μA741C	±18 V
Internal Power Dissipation (Note 1)	
Metal Can	500 mW
Molded and Hermetic DIP	670 mW
Mini DIP	310 mW
Flatpak	570 mW
Differential Input Voltage	±30 V
Input Voltage (Note 2)	±15 V
Storage Temperature Range	
Metal Can, Hermetic DIP, and Flatpak	-65°C to +150°C
Mini DIP, Molded DIP	-55°C to +125°C
Operating Temperature Range	
Military (μA741A, μA741)	-55°C to +125°C
Commercial (μA741E, μA741C)	0°C to +70°C
Lead Temperature (Soldering)	
Metal Can, Hermetic DIPs, and Flatpak (60 s)	300°C
Molded DIPs (10 s)	260°C
Output Short Circuit Duration (Note 3)	Indefinite

CONNECTION DIAGRAMS

**8-LEAD METAL CAN
(TOP VIEW)
PACKAGE OUTLINE 5B**

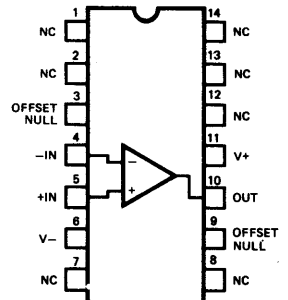


Note: Pin 4 connected to case

ORDER INFORMATION

TYPE	PART NO.
μA741A	μA741AHM
μA741	μA741HM
μA741E	μA741ERC
μA741C	μA741HC

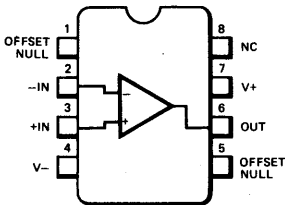
**14-LEAD DIP
(TOP VIEW)
PACKAGE OUTLINE 6A, 9A**



ORDER INFORMATION

TYPE	PART NO.
μA741A	μA741ADM
μA741	μA741DM
μA741E	μA741EDC
μA741C	μA741DC
μA741C	μA741PC

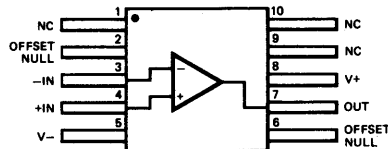
**8-LEAD MINIDIP
(TOP VIEW)
PACKAGE OUTLINES 6T 9T
PACKAGE CODES T R**



ORDER INFORMATION

TYPE	PART NO.
μA741C	μA741TC
μA741C	μA741RC

**10-LEAD FLATPAK
(TOP VIEW)
PACKAGE OUTLINE 3F**



ORDER INFORMATION

TYPE	PART NO.
μA741A	μA741AFM
μA741	μA741FM

FAIRCHILD LINEAR INTEGRATED CIRCUITS • $\mu A741$

$\mu A741A$

ELECTRICAL CHARACTERISTICS ($V_S = \pm 15V$, $T_A = 25^\circ C$ unless otherwise specified)

PARAMETERS (see definitions)		CONDITIONS	MIN	TYP	MAX	UNITS
Input Offset Voltage		$R_S \leq 50\Omega$		0.8	3.0	mV
Average Input Offset Voltage Drift					15	$\mu V/^\circ C$
Input Offset Current				3.0	30	nA
Average Input Offset Current Drift					0.5	$nA/^\circ C$
Input Bias Current				30	80	nA
Power Supply Rejection Ratio		$V_S = +10, -20; V_S = +20, -10V, R_S = 50\Omega$		15	50	$\mu V/V$
Output Short Circuit Current			10	25	35	mA
Power Dissipation		$V_S = \pm 20V$		80	150	mW
Input Impedance		$V_S = \pm 20V$	1.0	6.0		M Ω
Large Signal Voltage Gain		$V_S = \pm 20V, R_L = 2k\Omega, V_{OUT} = \pm 15V$	50			V/mV
Transient Response (Unity Gain)	Rise Time			0.25	0.8	μs
	Overshoot			6.0	20	%
Bandwidth (Note 4)			.437	1.5		MHz
Slew Rate (Unity Gain)		$V_{IN} = \pm 10V$	0.3	0.7		V/ μs
The following specifications apply for $-55^\circ C \leq T_A \leq +125^\circ C$						
Input Offset Voltage					4.0	mV
Input Offset Current					70	nA
Input Bias Current					210	nA
Common Mode Rejection Ratio		$V_S = \pm 20V, V_{IN} = \pm 15V, R_S = 50\Omega$	80	95		dB
Adjustment For Input Offset Voltage		$V_S = \pm 20V$	10			mV
Output Short Circuit Current			10		40	mA
Power Dissipation	$V_S = \pm 20V$	$-55^\circ C$			165	mW
		$+125^\circ C$			135	mW
Input Impedance		$V_S = \pm 20V$	0.5			M Ω
Output Voltage Swing	$V_S = \pm 20V,$	$R_L = 10k\Omega$	± 16			V
		$R_L = 2k\Omega$	± 15			V
Large Signal Voltage Gain		$V_S = \pm 20V, R_L = 2k\Omega, V_{OUT} = \pm 15V$	32			V/mV
		$V_S = \pm 5V, R_L = 2k\Omega, V_{OUT} = \pm 2V$	10			V/mV

NOTES

- Rating applies to ambient temperatures up to $70^\circ C$. Above $70^\circ C$ ambient derate linearly at $6.3mW/^\circ C$ for the metal can, $8.3mW/^\circ C$ for the DIP and $7.1mW/^\circ C$ for the Flatpak.
- For supply voltages less than $\pm 15V$, the absolute maximum input voltage is equal to the supply voltage.
- Short circuit may be to ground or either supply. Rating applies to $+125^\circ C$ case temperature or $75^\circ C$ ambient temperature.
- Calculated value from: $BW(MHz) = \frac{0.35}{\text{Rise Time } (\mu s)}$

FAIRCHILD LINEAR INTEGRATED CIRCUITS • $\mu A741$

$\mu A741$

ELECTRICAL CHARACTERISTICS ($V_S = \pm 15\text{ V}$, $T_A = 25^\circ\text{C}$ unless otherwise specified)

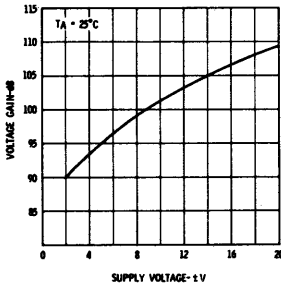
PARAMETERS (see definitions)	CONDITIONS	MIN	TYP	MAX	UNITS
Input Offset Voltage	$R_S < 10\text{ k}\Omega$		1.0	5.0	mV
Input Offset Current			20	200	nA
Input Bias Current			80	500	nA
Input Resistance		0.3	2.0		M Ω
Input Capacitance			1.4		pF
Offset Voltage Adjustment Range			± 15		mV
Large Signal Voltage Gain	$R_L > 2\text{ k}\Omega$, $V_{OUT} = \pm 10\text{ V}$	50,000	200,000		
Output Resistance			75		Ω
Output Short Circuit Current			25		mA
Supply Current			1.7	2.8	mA
Power Consumption			50	85	mW
Transient Response (Unity Gain)	Rise time Overshoot	$V_{IN} = 20\text{ mV}$, $R_L = 2\text{ k}\Omega$, $C_L < 100\text{ pF}$	0.3		μs
			5.0		%
Slew Rate	$R_L > 2\text{ k}\Omega$		0.5		V/ μs

The following specifications apply for $-55^\circ\text{C} < T_A < +125^\circ\text{C}$:

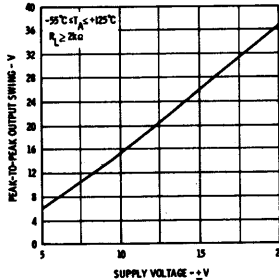
Input Offset Voltage	$R_S < 10\text{ k}\Omega$		1.0	6.0	mV
Input Offset Current	$T_A = +125^\circ\text{C}$		7.0	200	nA
	$T_A = -55^\circ\text{C}$		85	500	nA
Input Bias Current	$T_A = +125^\circ\text{C}$		0.03	0.5	μA
	$T_A = -55^\circ\text{C}$		0.3	1.5	μA
Input Voltage Range		± 12	± 13		V
Common Mode Rejection Ratio	$R_S < 10\text{ k}\Omega$	70	90		dB
Supply Voltage Rejection Ratio	$R_S < 10\text{ k}\Omega$		30	150	$\mu\text{V/V}$
Large Signal Voltage Gain	$R_L > 2\text{ k}\Omega$, $V_{OUT} = \pm 10\text{ V}$	25,000			
Output Voltage Swing	$R_L > 10\text{ k}\Omega$	± 12	± 14		V
	$R_L > 2\text{ k}\Omega$	± 10	± 13		V
Supply Current	$T_A = +125^\circ\text{C}$		1.5	2.5	mA
	$T_A = -55^\circ\text{C}$		2.0	3.3	mA
Power Consumption	$T_A = +125^\circ\text{C}$		45	75	mW
	$T_A = -55^\circ\text{C}$		60	100	mW

TYPICAL PERFORMANCE CURVES FOR $\mu A741A$ AND $\mu A741$

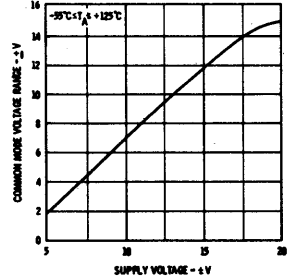
OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF SUPPLY VOLTAGE



OUTPUT VOLTAGE SWING AS A FUNCTION OF SUPPLY VOLTAGE



INPUT COMMON MODE VOLTAGE RANGE AS A FUNCTION OF SUPPLY VOLTAGE

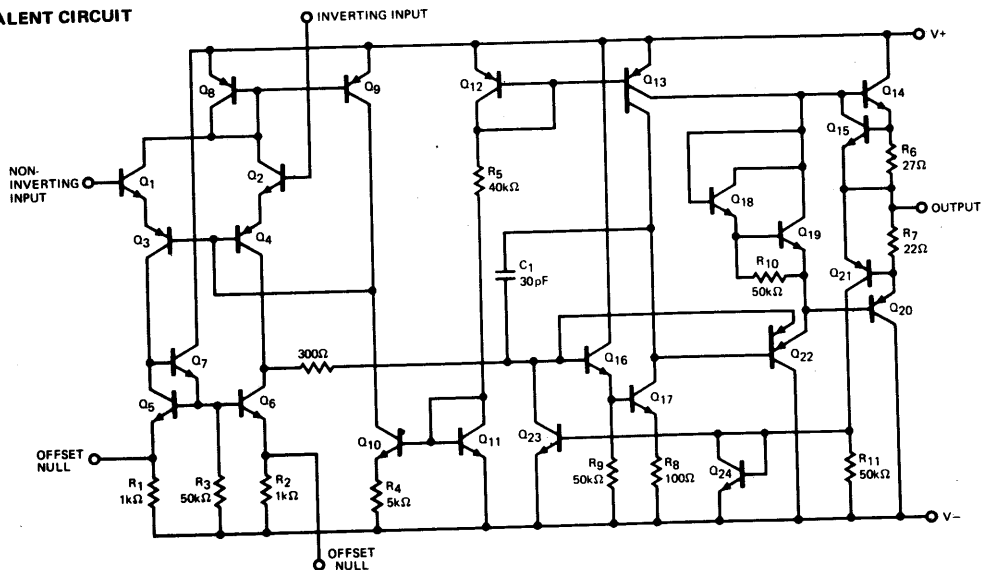


$\mu A741E$

ELECTRICAL CHARACTERISTICS ($V_S = \pm 15V$, $T_A = 25^\circ C$ unless otherwise specified)

PARAMETERS (see definitions)	CONDITIONS	MIN	TYP	MAX	UNITS
Input Offset Voltage	$R_S < 50\Omega$		0.8	3.0	mV
Average Input Offset Voltage Drift				15	$\mu V/^\circ C$
Input Offset Current			3.0	30	nA
Average Input Offset Current Drift				0.5	$nA/^\circ C$
Input Bias Current			30	80	nA
Power Supply Rejection Ratio	$V_S = +10, -20; V_S = +20, -10V, R_S = 50\Omega$		15	50	$\mu V/V$
Output Short Circuit Current		10	25	35	mA
Power Dissipation	$V_S = \pm 20V$		80	150	mW
Input Impedance	$V_S = \pm 20V$	1.0	6.0		$M\Omega$
Large Signal Voltage Gain	$V_S = \pm 20V, R_L = 2k\Omega, V_{OUT} = \pm 15V$	50			V/mV
Transient Response (Unity Gain)	Rise Time		0.25	0.8	μs
	Overshoot		6.0	20	%
Bandwidth (Note 4)		.437	1.5		MHz
Slew Rate (Unity Gain)	$V_{IN} = \pm 10V$	0.3	0.7		V/ μs
The following specifications apply for $0^\circ C < T_A < 70^\circ C$					
Input Offset Voltage				4.0	mV
Input Offset Current				70	nA
Input Bias Current				210	nA
Common Mode Rejection Ratio	$V_S = \pm 20V, V_{IN} = \pm 15V, R_S = 50\Omega$	80	95		dB
Adjustment For Input Offset Voltage	$V_S = \pm 20V$	10			mV
Output Short Circuit Current		10		40	mA
Power Dissipation	$V_S = \pm 20V$			150	mW
Input Impedance	$V_S = \pm 20V$				$M\Omega$
Output Voltage Swing	$V_S = \pm 20V, R_L = 10k\Omega$		± 16		V
	$R_L = 2k\Omega$		± 15		V
Large Signal Voltage Gain	$V_S = \pm 20V, R_L = 2k\Omega, V_{OUT} = \pm 15V$	32			V/mV
	$V_S = \pm 5V, R_L = 2k\Omega, V_{OUT} = \pm 2V$	10			V/mV

EQUIVALENT CIRCUIT



FAIRCHILD LINEAR INTEGRATED CIRCUITS • $\mu A741$

$\mu A741C$

ELECTRICAL CHARACTERISTICS ($V_S = \pm 15\text{ V}$, $T_A = 25^\circ\text{C}$ unless otherwise specified)

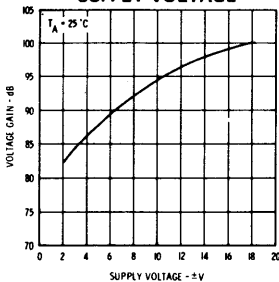
PARAMETERS (see definitions)	CONDITIONS	MIN	TYP	MAX	UNITS
Input Offset Voltage	$R_S \leq 10\text{ k}\Omega$		2.0	6.0	mV
Input Offset Current			20	200	nA
Input Bias Current			80	500	nA
Input Resistance		0.3	2.0		$M\Omega$
Input Capacitance			1.4		pF
Offset Voltage Adjustment Range			± 15		mV
Input Voltage Range		± 12	± 13		V
Common Mode Rejection Ratio	$R_S \leq 10\text{ k}\Omega$	70	90		dB
Supply Voltage Rejection Ratio	$R_S \leq 10\text{ k}\Omega$		30	150	$\mu\text{V/V}$
Large Signal Voltage Gain	$R_L \geq 2\text{ k}\Omega$, $V_{OUT} = \pm 10\text{ V}$	20,000	200,000		
Output Voltage Swing	$R_L \geq 10\text{ k}\Omega$	± 12	± 14		V
	$R_L \geq 2\text{ k}\Omega$	± 10	± 13		V
Output Resistance			75		Ω
Output Short Circuit Current			25		mA
Supply Current			1.7	2.8	mA
Power Consumption			50	85	mW
Transient Response (Unity Gain)	$V_{IN} = 20\text{ mV}$, $R_L = 2\text{ k}\Omega$, $C_L \leq 100\text{ pF}$	Rise time	0.3		μs
		Overshoot	5.0		%
Slew Rate	$R_L \geq 2\text{ k}\Omega$		0.5		$\text{V}/\mu\text{s}$

The following specifications apply for $0^\circ\text{C} \leq T_A \leq +70^\circ\text{C}$:

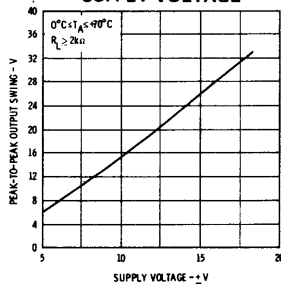
Input Offset Voltage				7.5	mV
Input Offset Current				300	nA
Input Bias Current				800	nA
Large Signal Voltage Gain	$R_L \geq 2\text{ k}\Omega$, $V_{OUT} = \pm 10\text{ V}$	15,000			
Output Voltage Swing	$R_L \geq 2\text{ k}\Omega$	± 10	± 13		V

TYPICAL PERFORMANCE CURVES FOR $\mu A741E$ AND $\mu A741C$

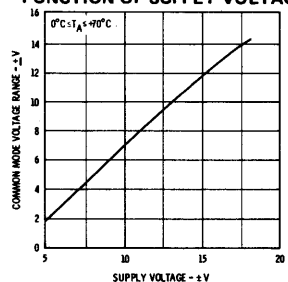
OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF SUPPLY VOLTAGE



OUTPUT VOLTAGE SWING AS A FUNCTION OF SUPPLY VOLTAGE

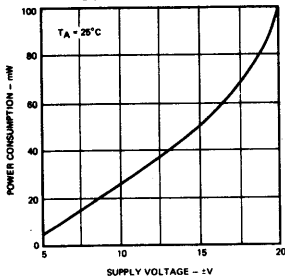


INPUT COMMON MODE VOLTAGE RANGE AS A FUNCTION OF SUPPLY VOLTAGE

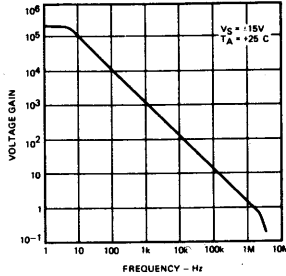


TYPICAL PERFORMANCE CURVES FOR μ A741A, μ A741, μ A741E AND μ A741C

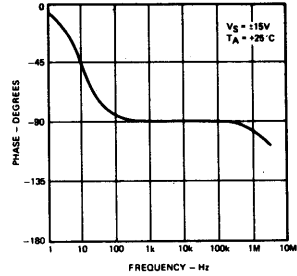
POWER CONSUMPTION AS A FUNCTION OF SUPPLY VOLTAGE



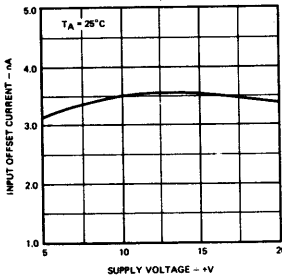
OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF FREQUENCY



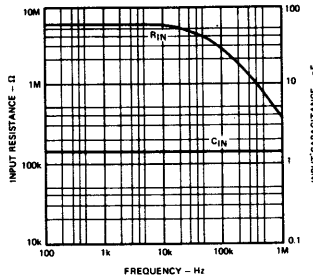
OPEN LOOP PHASE RESPONSE AS A FUNCTION OF FREQUENCY



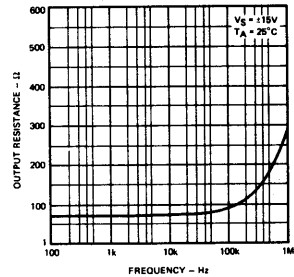
INPUT OFFSET CURRENT AS A FUNCTION OF SUPPLY VOLTAGE



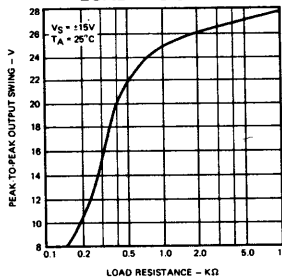
INPUT RESISTANCE AND INPUT CAPACITANCE AS A FUNCTION OF FREQUENCY



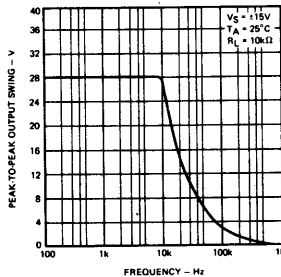
OUTPUT RESISTANCE AS A FUNCTION OF FREQUENCY



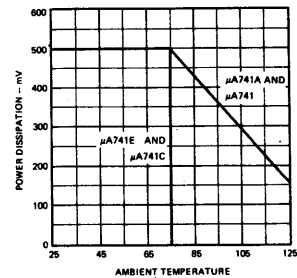
OUTPUT VOLTAGE SWING AS A FUNCTION OF LOAD RESISTANCE



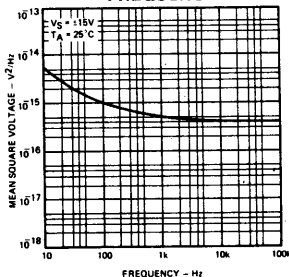
OUTPUT VOLTAGE SWING AS A FUNCTION OF FREQUENCY



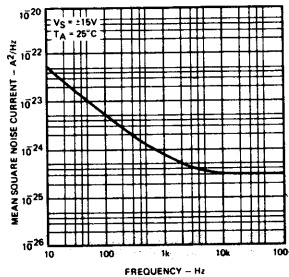
ABSOLUTE MAXIMUM POWER DISSIPATION AS A FUNCTION OF AMBIENT TEMPERATURE



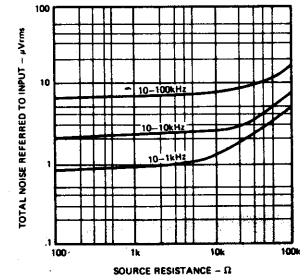
INPUT NOISE VOLTAGE AS A FUNCTION OF FREQUENCY



INPUT NOISE CURRENT AS A FUNCTION OF FREQUENCY

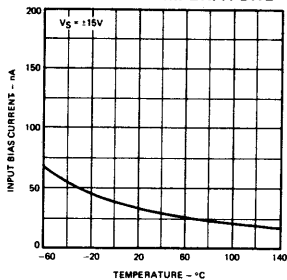


BROADBAND NOISE FOR VARIOUS BANDWIDTHS

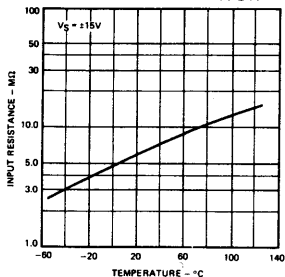


TYPICAL PERFORMANCE CURVES FOR $\mu A741A$ AND $\mu A741$

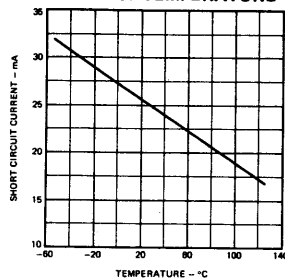
INPUT BIAS CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



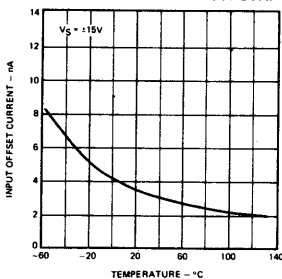
INPUT RESISTANCE AS A FUNCTION OF AMBIENT TEMPERATURE



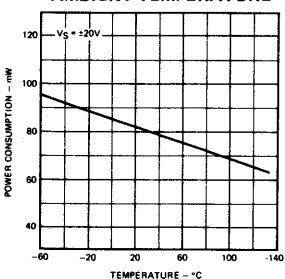
OUTPUT SHORT-CIRCUIT CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



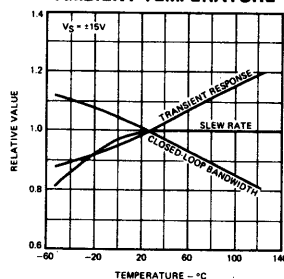
INPUT OFFSET CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



POWER CONSUMPTION AS A FUNCTION OF AMBIENT TEMPERATURE

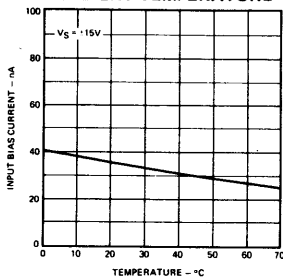


FREQUENCY CHARACTERISTICS AS A FUNCTION OF AMBIENT TEMPERATURE

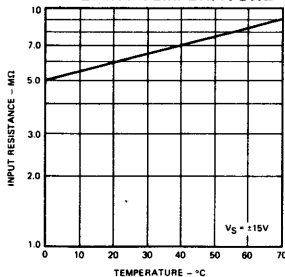


TYPICAL PERFORMANCE CURVES FOR $\mu A741E$ AND $\mu A741C$

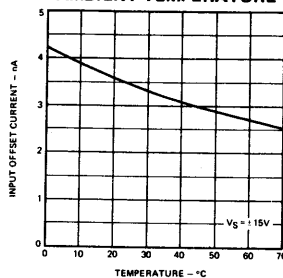
INPUT BIAS CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



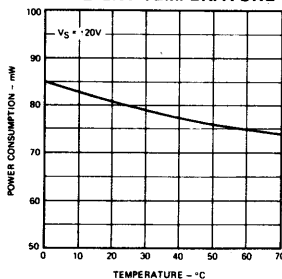
INPUT RESISTANCE AS A FUNCTION OF AMBIENT TEMPERATURE



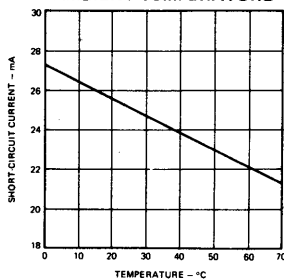
INPUT OFFSET CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



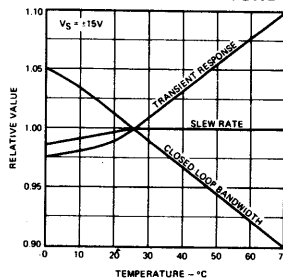
POWER CONSUMPTION AS A FUNCTION OF AMBIENT TEMPERATURE



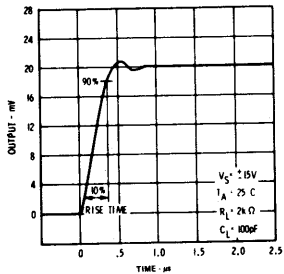
OUTPUT SHORT CIRCUIT CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



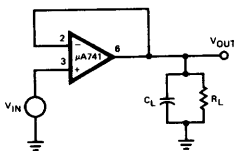
FREQUENCY CHARACTERISTICS AS A FUNCTION OF AMBIENT TEMPERATURE



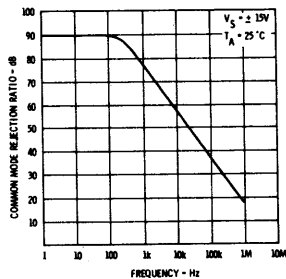
TRANSIENT RESPONSE



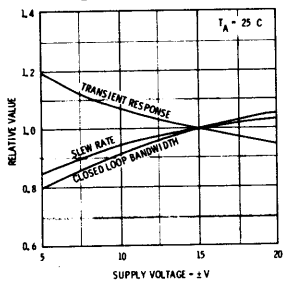
TRANSIENT RESPONSE TEST CIRCUIT



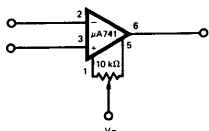
COMMON MODE REJECTION RATIO AS A FUNCTION OF FREQUENCY



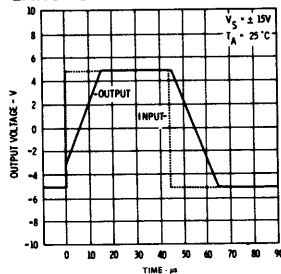
FREQUENCY CHARACTERISTICS AS A FUNCTION OF SUPPLY VOLTAGE



VOLTAGE OFFSET NULL CIRCUIT

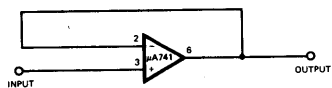


VOLTAGE FOLLOWER LARGE SIGNAL PULSE RESPONSE



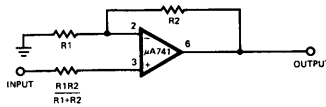
TYPICAL APPLICATIONS

UNITY-GAIN VOLTAGE FOLLOWER



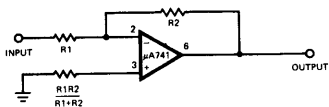
$R_{IN} = 400 M\Omega$
 $C_{IN} = 1 pF$
 $R_{OUT} < < 1 \Omega$
 B.W. = 1 MHz

NON-INVERTING AMPLIFIER



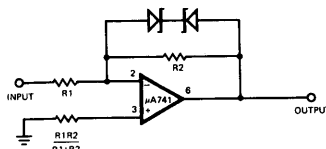
GAIN	R1	R2	BW	R_{IN}
10	1 k Ω	9 k Ω	100 kHz	400 M Ω
100	100 Ω	9.9 k Ω	10 kHz	280 M Ω
1000	100 Ω	99.9 k Ω	1 kHz	80 M Ω

INVERTING AMPLIFIER



GAIN	R1	R2	BW	R_{IN}
1	10 k Ω	10 k Ω	1 MHz	10 k Ω
10	1 k Ω	10 k Ω	100 kHz	1 k Ω
100	1 k Ω	100 k Ω	10 kHz	1 k Ω
1000	100 Ω	100 k Ω	1 kHz	100 Ω

CLIPPING AMPLIFIER



$$\frac{E_{OUT}}{E_{IN}} = \frac{R2}{R1} \text{ if } |E_{OUT}| \leq V_Z + 0.7 V$$

where V_Z = Zener breakdown voltage

μA747

DUAL FREQUENCY-COMPENSATED OPERATIONAL AMPLIFIER

FAIRCHILD LINEAR INTEGRATED CIRCUITS

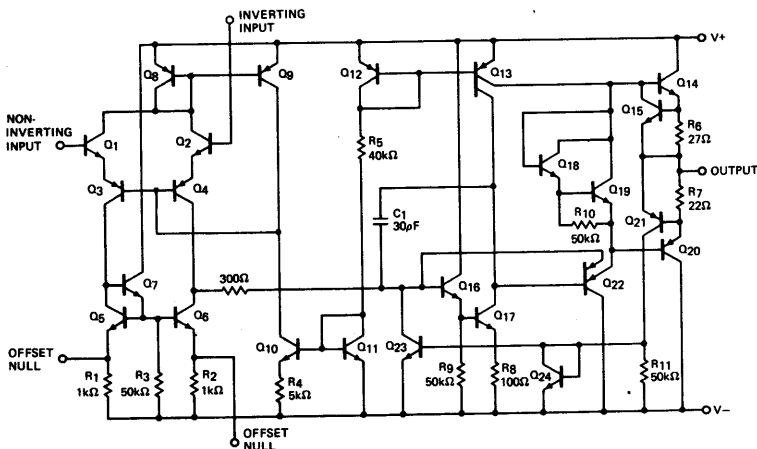
GENERAL DESCRIPTION — The μA747 is a pair of high performance monolithic Operational Amplifiers constructed using the Fairchild Planar* epitaxial process. They are intended for a wide range of analog applications where board space or weight are important. High common mode voltage range and absence of latch-up make the μA747 ideal for use as a voltage follower. The high gain and wide range of operating voltage provides superior performance in integrator, summing amplifier, and general feedback applications. The μA747 is short circuit protected and requires no external components for frequency compensation. The internal 6 dB/octave roll-off insures stability in closed loop applications. For single amplifier performance, see μA741 data sheet. Electrical characteristics are identical to MIL-M-38510/10102.

- NO FREQUENCY COMPENSATION REQUIRED
- SHORT CIRCUIT PROTECTION
- OFFSET VOLTAGE NULL CAPABILITY
- LARGE COMMON MODE AND DIFFERENTIAL VOLTAGE RANGES
- LOW POWER CONSUMPTION
- NO LATCH-UP

ABSOLUTE MAXIMUM RATINGS

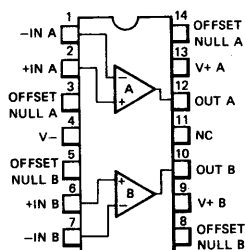
Supply Voltage	
Military (μA747A, μA747, μA747E)	±22 V
Commercial (μA747C)	±18 V
Internal Power Dissipation (Note 1)	
Metal Can	500 mW
DIP	670 mW
Differential Input Voltage	±30 V
Input Voltage (Note 2)	±15 V
Voltage between Offset Null and V-	±0.5 V
Storage Temperature Range	-65°C to +150°C
Operating Temperature Range	
Military (μA747A, μA747)	-55°C to +125°C
Commercial (μA747E, μA747C)	0°C to 70°C
Lead Temperature (Soldering 60s)	300°C
Output Short Circuit Duration (Note 3)	Indefinite

EQUIVALENT CIRCUIT (1/2 μA747)



CONNECTION DIAGRAMS 14-LEAD DIP (TOP VIEW)

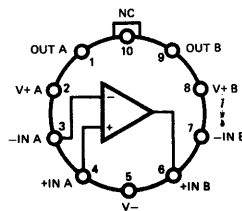
PACKAGE OUTLINE 7A 9A
PACKAGE CODE D P



TYPE	PART NO.
μA747A	μA747ADM
μA747	μA747DM
μA747E	μA747EDC
μA747C	μA747DC
μA747C	μA747PC
μA747-I	μA747-IDM
μA747-IC	μA747-IDC

10-LEAD METAL CAN (TOP VIEW)

PACKAGE OUTLINE 5F



TYPE	PART NO.
μA747A	μA747AHM
μA747	μA747HM
μA747E	μA747EHC
μA747C	μA747HC
μA747-I	μA747-IHM
μA747-IC	μA747-IHC

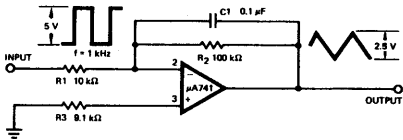
NOTE:

V+ is internally connected to V+B for μA747A, μA747, μA747E, and μA747C. They are not internally connected for μA747-I and μA747-IC.

*Planar is a patented Fairchild process.

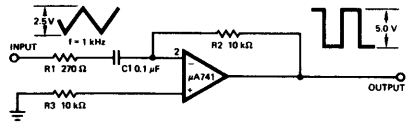
TYPICAL APPLICATIONS (Cont'd)

SIMPLE INTEGRATOR



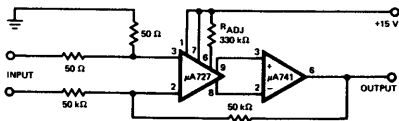
$$E_{OUT} = - \frac{1}{R_1 C_1} \int E_{IN} dt$$

SIMPLE DIFFERENTIATOR



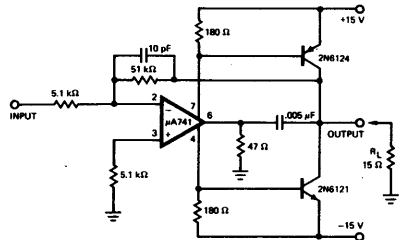
$$E_{OUT} = - R_2 C_1 \frac{dE_{IN}}{dt}$$

LOW DRIFT LOW NOISE AMPLIFIER

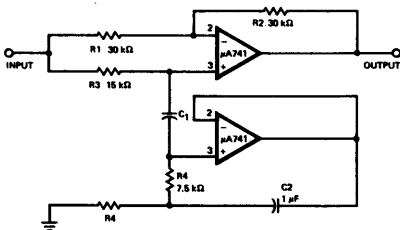


Voltage Gain = 10^3
 Input Offset Voltage Drift = $0.6 \mu V/^{\circ}C$
 Input Offset Current Drift = $2.0 pA/^{\circ}C$

HIGH SLEW RATE POWER AMPLIFIER

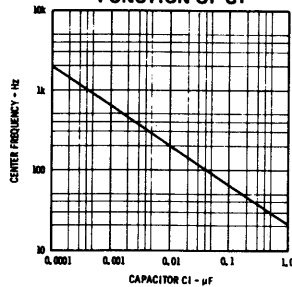


NOTCH FILTER USING THE $\mu A741$ AS A GYRATOR



Trim R3 such that
 $\frac{R1}{R2} = \frac{R3}{2 R4}$

NOTCH FREQUENCY AS A FUNCTION OF C1



FAIRCHILD LINEAR INTEGRATED CIRCUITS • μ A747

μ A747A

ELECTRICAL CHARACTERISTICS $\pm 5\text{ V} < V_S < \pm 20\text{ V}$, $T_A = 25^\circ\text{C}$ unless otherwise specified)

PARAMETERS		CONDITIONS	MIN	TYP	MAX	UNITS
Input Offset Voltage		$R_S < 50\ \Omega$		0.8	3.0	mV
Average Input Offset Voltage Drift					15	$\mu\text{V}/^\circ\text{C}$
Input Offset Current				3.0	30	nA
Average Input Offset Current Drift		$T_A = 25^\circ\text{C}$ to $+125^\circ\text{C}$ $T_A = -55^\circ\text{C}$ to $+25^\circ\text{C}$			0.2 0.5	nA/ $^\circ\text{C}$ nA/ $^\circ\text{C}$
Input Bias Current				30	80	nA
Power Supply Rejection Ratio		$V_S = +10, -20; V_S = +20\text{ V}, -10\text{ V}$ $R_S = 50\ \Omega$	0.2	15	50	$\mu\text{V}/\text{V}$
Common Mode Rejection Ratio		$V_S = \pm 20\text{ V}$, $V_{IN} = \pm 15\text{ V}$ $R_S = 50\ \Omega$	80	95		dB
Adjustment for Input Offset Voltage		$V_S = \pm 20\text{ V}$	10			mV
Output Short Circuit Current			10	25	35	mA
Power Dissipation		$V_S = \pm 20\text{ V}$ per Channel		80	150	mW
Input Impedance		$V_S = \pm 20\text{ V}$	1.0	6		M Ω
Large Signal Voltage Gain		$V_S = \pm 20\text{ V}$, $R_L = 2\text{ k}\Omega$ $V_{OUT} = \pm 15\text{ V}$	50			V/mV
Transient Response (Unity Gain)	Rise Time			0.25	0.8	μs
	Overshoot			6.0	20	%
Bandwidth (Note 4)			0.437	1.5		MHz
Slew Rate (Unity Gain)		$V_{IN} = \pm 10\text{ V}$	0.3	0.7		V/ μs
The following specifications apply for $-55^\circ\text{C} < T_A < +125^\circ\text{C}$						
Input Offset Voltage					4.0	mV
Input Offset Current					70	nA
Input Bias Current					210	nA
Output Short Circuit Current			10		40	mA
Power Dissipation	$V_S = \pm 20\text{ V}$	-55°C			165	mW
		$+125^\circ\text{C}$			135	mW
Input Impedance		$V_S = \pm 20\text{ V}$	0.5			M Ω
Output Voltage Swing	$V_S = \pm 20\text{ V}$, $R_L = 10\text{ k}\Omega$ $R_L = 2\text{ k}\Omega$		± 16			V
			± 15			V
Large Signal Voltage Gain	$V_S = \pm 20\text{ V}$, $R_L = 2\text{ k}\Omega$, $V_{OUT} = \pm 15\text{ V}$ $V_S = \pm 5\text{ V}$, $R_L = 2\text{ k}\Omega$, $V_{OUT} = \pm 2\text{ V}$		32			V/mV
			10			V/mV
Channel Separation		$V_S = \pm 20\text{ V}$	100			dB

- NOTES:**
- Rating applies to ambient temperatures up to 70°C . Above 70°C ambient derate linearly at $6.3\text{ mW}/^\circ\text{C}$ for the Metal Can, $8.3\text{ mW}/^\circ\text{C}$ for the DIP.
 - For supply voltages less than $\pm 15\text{ V}$, the absolute maximum input voltage is equal to the supply voltage.
 - Short circuit may be to ground or either supply. Rating applies to $+125^\circ\text{C}$ case temperature or 75°C ambient temperature.
 - Calculated value from: $\text{BW (MHz)} = \frac{0.35}{\text{RISE TIME } (\mu\text{s})}$

FAIRCHILD LINEAR INTEGRATED CIRCUITS • $\mu A747$

$\mu A747$

ELECTRICAL CHARACTERISTICS – Each Amplifier ($V_S = \pm 15\text{ V}$, $T_A = 25^\circ\text{C}$ unless otherwise specified)

PARAMETERS (see definitions)	CONDITIONS	MIN.	TYP.	MAX.	UNITS
Input Offset Voltage	$R_S < 10\text{ k}\Omega$		1.0	5.0	mV
Input Offset Current			20	200	nA
Input Bias Current			80	500	nA
Input Resistance		0.3	2.0		M Ω
Input Capacitance			1.4		pF
Offset Voltage Adjustment Range			± 15		mV
Large Signal Voltage Gain	$R_L \geq 2\text{ k}\Omega$, $V_{OUT} = \pm 10\text{ V}$	50,000	200,000		V/V
Output Resistance			75		Ω
Output Short-Circuit Current			25		mA
Supply Current			1.7	2.8	mA
Power Consumption			50	85	mW
Transient Response (Unity Gain)	Rise time	$V_{IN} = 20\text{ mV}$, $R_L = 2\text{ k}\Omega$, $C_L < 100\text{ pF}$		0.3	μs
	Overshoot			5.0	%
Slew Rate	$R_L \geq 2\text{ k}\Omega$		0.5		V/ μs
Channel Separation			120		dB

The following specifications apply for $-55^\circ\text{C} < T_A < +125^\circ\text{C}$.

Input Offset Voltage	$R_S < 10\text{ k}\Omega$		1.0	6.0	mV
Input Offset Current	$T_A = +125^\circ\text{C}$		7.0	200	nA
	$T_A = -55^\circ\text{C}$		85	500	nA
Input Bias Current	$T_A = +125^\circ\text{C}$		0.03	0.5	μA
	$T_A = -55^\circ\text{C}$		0.3	1.5	μA
Input Voltage Range		± 12	± 13		V
Common Mode Rejection Ratio	$R_S < 10\text{ k}\Omega$	70	90		dB
Supply Voltage Rejection Ratio	$R_S < 10\text{ k}\Omega$		30	150	$\mu\text{V/V}$
Large Signal Voltage Gain	$R_L \geq 2\text{ k}\Omega$, $V_{OUT} = \pm 10\text{ V}$	25,000			V/V
Output Voltage Swing	$R_L \geq 10\text{ k}\Omega$	± 12	± 14		V
	$R_L \geq 2\text{ k}\Omega$	± 10	± 13		V
Supply Current	$T_A = +125^\circ\text{C}$		1.5	2.5	mA
	$T_A = -55^\circ\text{C}$		2.0	3.3	mA
Power Consumption	$T_A = +125^\circ\text{C}$		45	75	mW
	$T_A = -55^\circ\text{C}$		60	100	mW

FAIRCHILD LINEAR INTEGRATED CIRCUITS • $\mu A747$

$\mu A747C$

ELECTRICAL CHARACTERISTICS — Each Amplifier ($V_S = \pm 15\text{ V}$, $T_A = 25^\circ\text{C}$ unless otherwise specified)

PARAMETERS (see definitions)	CONDITIONS	MIN	TYP	MAX	UNITS
Input Offset Voltage	$R_S < 10\text{ k}\Omega$		1.0	6.0	mV
Input Offset Current			20	200	nA
Input Bias Current			80	500	nA
Input Resistance		0.3	2.0		M Ω
Input Capacitance			1.4		pF
Offset Voltage Adjustment Range			± 15		mV
Large Signal Voltage Gain	$R_L > 2\text{ k}\Omega$, $V_{OUT} = \pm 10\text{ V}$	25,000	200,000		V/V
Output Resistance			75		Ω
Output Short-Circuit Current			25		mA
Supply Current			1.7	2.8	mA
Power Consumption			50	85	mW
Transient Response (Unity Gain)	Rise time	$V_{IN} = 20\text{ mV}$, $R_L = 2\text{ k}\Omega$, $C_L < 100\text{ pF}$	0.3		μs
	Overshoot		5.0		%
Slew Rate	$R_L > 2\text{ k}\Omega$		0.5		V/ μs
Channel Separation			120		dB

The following specifications apply for $0^\circ\text{C} < T_A < +70^\circ\text{C}$.

Input Offset Voltage	$R_S < 10\text{ k}\Omega$		1.0	7.5	mV
Input Offset Current			7.0	300	nA
Input Bias Current			0.03	0.8	μA
Input Voltage Range		± 12	± 13		V
Common Mode Rejection Ratio	$R_S < 10\text{ k}\Omega$	70	90		dB
Supply Voltage Rejection Ratio	$R_S < 10\text{ k}\Omega$		30	150	$\mu\text{V/V}$
Large Signal Voltage Gain	$R_L > 2\text{ k}\Omega$, $V_{OUT} = \pm 10\text{ V}$	15,000			V/V
	$R_L > 10\text{ k}\Omega$	± 12	± 14		V
Output Voltage Swing	$R_L > 2\text{ k}\Omega$	± 10	± 13		V
Supply Current			2.0	3.3	mA
Power Consumption			60	100	mW

$\mu A747E$

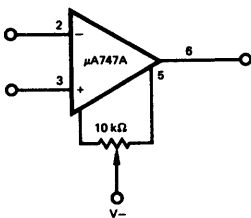
ELECTRICAL CHARACTERISTICS ($\pm 5\text{ V} < V_S < \pm 20\text{ V}$, $T_A = 25^\circ\text{C}$ unless otherwise specified)

PARAMETERS (see definitions)	CONDITIONS	MIN	TYP	MAX	UNITS
Input Offset Voltage	$R_S < 50\ \Omega$		0.8	3.0	mV
Average Input Offset Voltage Drift				15	$\mu\text{V}/^\circ\text{C}$
Input Offset Current			3	30	nA
Average Input Offset Current Drift	$T_A = 25^\circ\text{C}$ to 70°C $T_A = 0^\circ\text{C}$ to 25°C			0.2 0.5	nA/ $^\circ\text{C}$ nA/ $^\circ\text{C}$
Input Bias Current			30	80	nA
Power Supply Rejection Ratio	$V_S = +10, -20; V_S = +20\text{ V}, -10\text{ V}$ $R_S = 50\ \Omega$		15	50	$\mu\text{V}/\text{V}$
Common Mode Rejection Ratio	$V_S = \pm 20\text{ V}, V_{IN} = \pm 15\text{ V}$ $R_S = 50\ \Omega$	80	95		dB
Adjustment for Input Offset Voltage	$V_S = \pm 20\text{ V}$	10			mV
Output Short Circuit Current		10	25	35	mA
Power Dissipation	$V_S = \pm 20\text{ V}$		80	150	mW
Input Impedance	$V_S = \pm 20\text{ V}$	1.0	6		M Ω
Large Signal Voltage Gain	$V_S = \pm 20\text{ V}, R_L = 2\text{ k}\Omega, V_{OUT} = \pm 15\text{ V}$	50			V/mV
Transient Response (Unity Gain)	Rise Time		0.25	0.8	μs
	Overshoot		6	20	%
Bandwidth (Note 4)		0.437	1.5		MHz
Slew Rate (Unity Gain)	$V_{IN} = \pm 10\text{ V}$	0.3	0.7		V/ μs

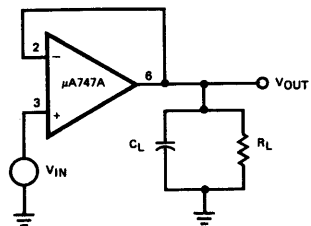
The following specifications apply for $0^\circ\text{C} < T_A < 70^\circ\text{C}$

Input Offset Voltage				4.0	mV
Input Offset Current				70	nA
Input Bias Current				210	nA
Output Short Circuit Current		10		40	mA
Power Dissipation	$V_S = \pm 20\text{ V}$			165	mW
Input Impedance	$V_S = \pm 20\text{ V}$				M Ω
Output Voltage Swing	$V_S = \pm 20\text{ V}, R_L = 10\text{ k}\Omega$ $R_L = 2\text{ k}\Omega$	± 16			V
		± 15			V
Large Signal Voltage Gain	$V_S = \pm 20\text{ V}, R_L = 2\text{ k}\Omega, V_{OUT} = \pm 15\text{ V}$ $V_S = \pm 5\text{ V}, R_L = 2\text{ k}\Omega, V_{OUT} = \pm 2\text{ V}$	32			V/mV
		10			V/mV
Channel Separation	$V_S = \pm 20\text{ V}$	100			dB

VOLTAGE OFFSET NULL CIRCUIT

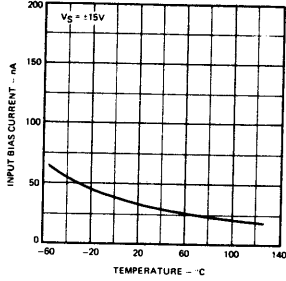


TRANSIENT RESPONSE TEST CIRCUIT

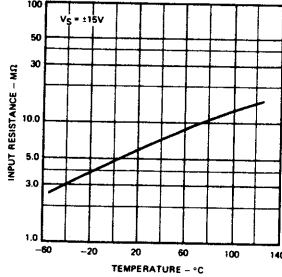


TYPICAL PERFORMANCE CURVES FOR $\mu A747A$ AND $\mu A747$

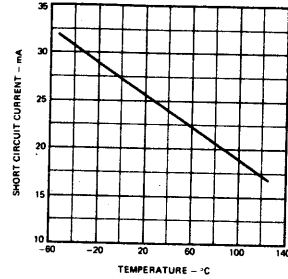
INPUT BIAS CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



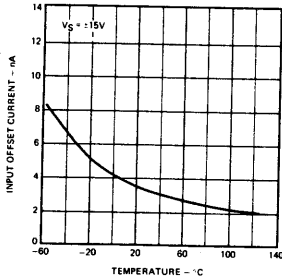
INPUT RESISTANCE AS A FUNCTION OF AMBIENT TEMPERATURE



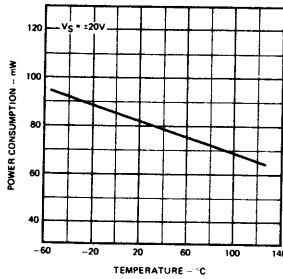
OUTPUT SHORT-CIRCUIT CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



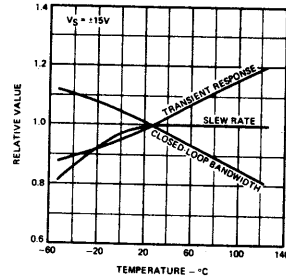
INPUT OFFSET CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



POWER CONSUMPTION AS A FUNCTION OF AMBIENT TEMPERATURE

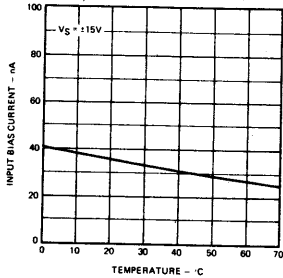


FREQUENCY CHARACTERISTICS AS A FUNCTION OF AMBIENT TEMPERATURE

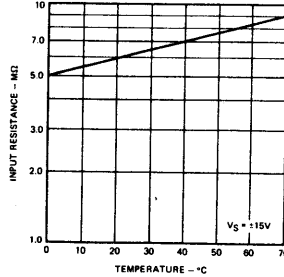


TYPICAL PERFORMANCE CURVES FOR $\mu A747E$ AND $\mu A747C$

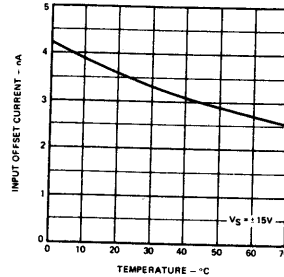
INPUT BIAS CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



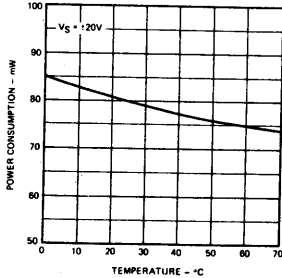
INPUT RESISTANCE AS A FUNCTION OF AMBIENT TEMPERATURE



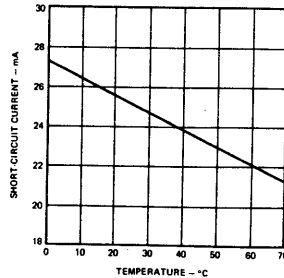
INPUT OFFSET CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



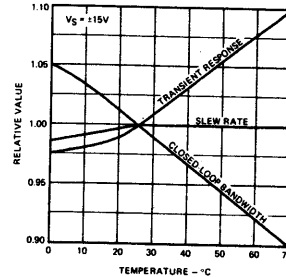
POWER CONSUMPTION AS A FUNCTION OF AMBIENT TEMPERATURE



OUTPUT SHORT CIRCUIT CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE

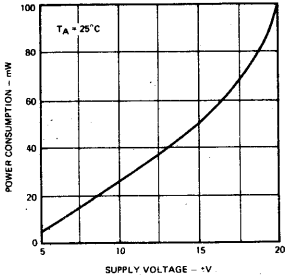


FREQUENCY CHARACTERISTICS AS A FUNCTION OF AMBIENT TEMPERATURE

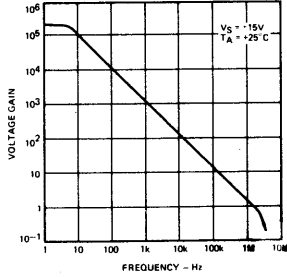


TYPICAL PERFORMANCE CURVES FOR $\mu A747A$, $\mu A747C$, $\mu A747$ AND $\mu A747E$

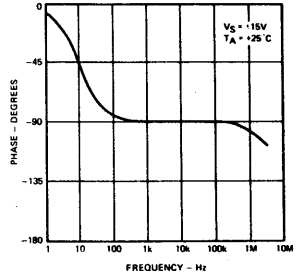
POWER CONSUMPTION AS A FUNCTION OF SUPPLY VOLTAGE



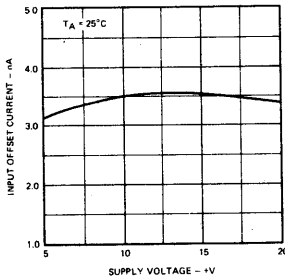
OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF FREQUENCY



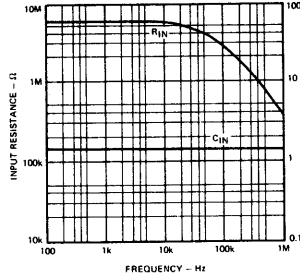
OPEN LOOP PHASE RESPONSE AS A FUNCTION OF FREQUENCY



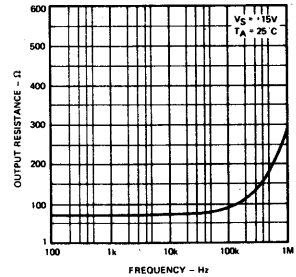
INPUT OFFSET CURRENT AS A FUNCTION OF SUPPLY VOLTAGE



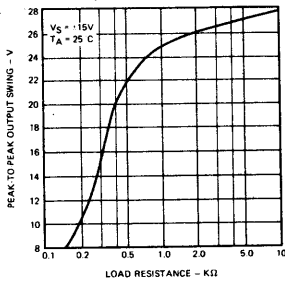
INPUT RESISTANCE AND INPUT CAPACITANCE AS A FUNCTION OF FREQUENCY



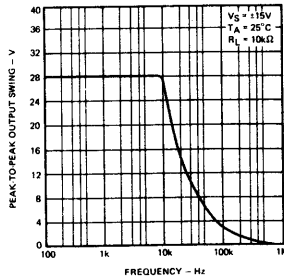
OUTPUT RESISTANCE AS A FUNCTION OF FREQUENCY



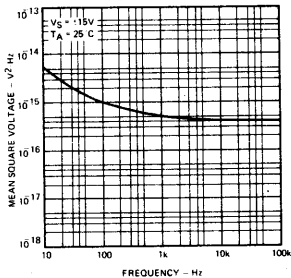
OUTPUT VOLTAGE SWING AS A FUNCTION OF LOAD RESISTANCE



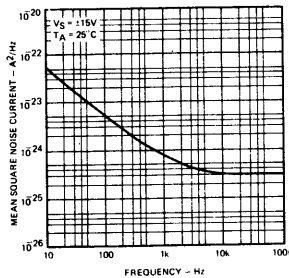
OUTPUT VOLTAGE SWING AS A FUNCTION OF FREQUENCY



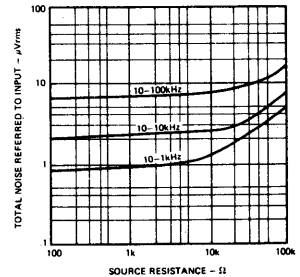
INPUT NOISE VOLTAGE AS A FUNCTION OF FREQUENCY



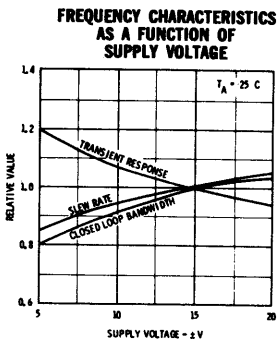
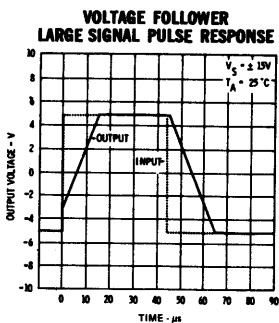
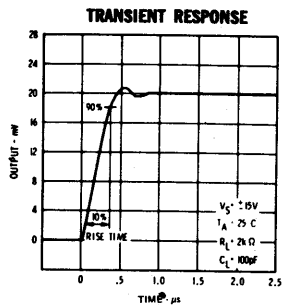
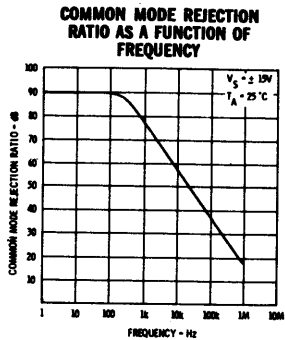
INPUT NOISE CURRENT AS A FUNCTION OF FREQUENCY



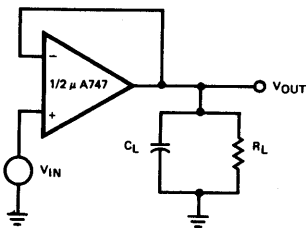
BROADBAND NOISE FOR VARIOUS BANDWIDTHS



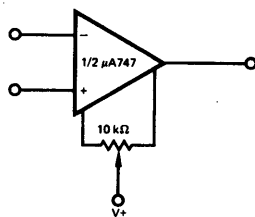
TYPICAL PERFORMANCE CURVES (Each Amplifier) FOR μ A747 AND μ A747C



TRANSIENT RESPONSE TEST CIRCUIT

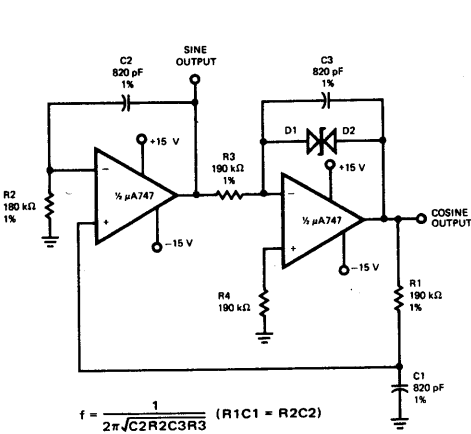


VOLTAGE OFFSET NULL CIRCUIT

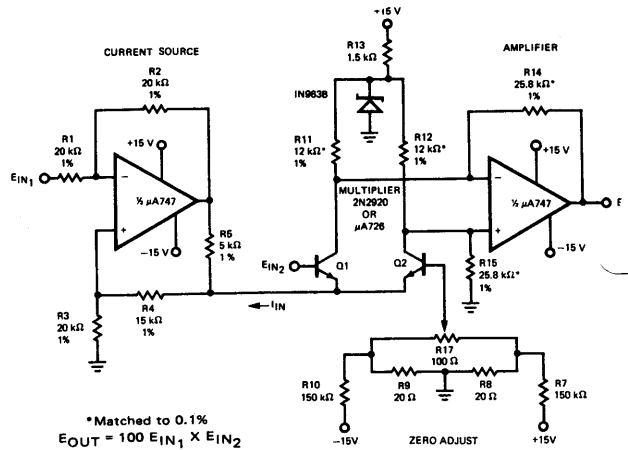


TYPICAL APPLICATIONS

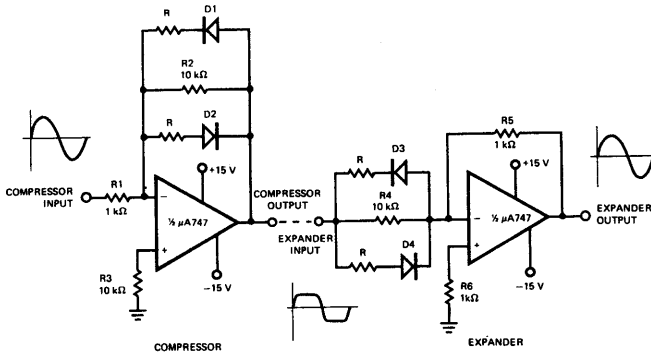
QUADRATURE OSCILLATOR



ANALOG MULTIPLIER

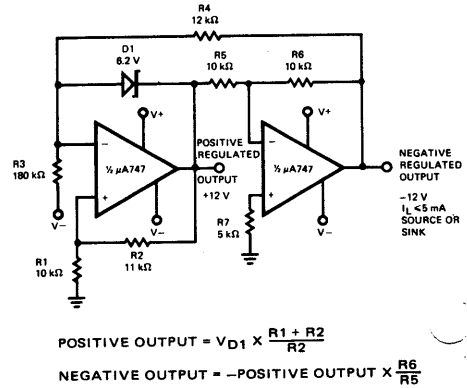


COMPRESSOR/EXPANDER AMPLIFIERS

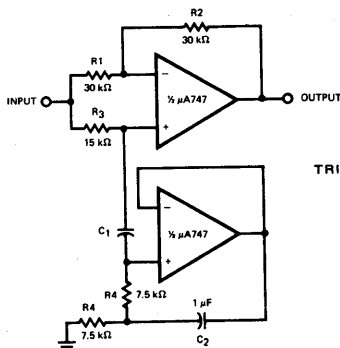


MAXIMUM COMPRESSION EXPANSION RATIO = R_1/R_2 ($10 \text{ k}\Omega > R > 0$)
 NOTE: DIODES D1 THROUGH D4 ARE MATCHED FD668 OR EQUIVALENT

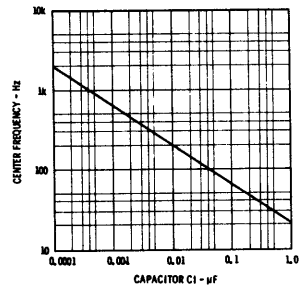
TRACKING POSITIVE AND NEGATIVE VOLTAGE REFERENCES



NOTCH FILTER USING THE $\mu A747$ AS A GYRATOR

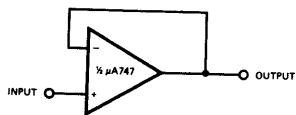


NOTCH FREQUENCY AS A FUNCTION OF C1



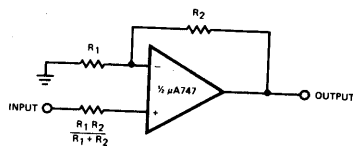
TYPICAL APPLICATIONS

UNITY-GAIN VOLTAGE FOLLOWER



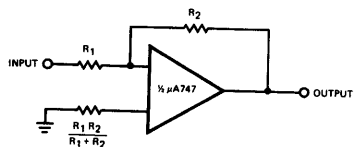
$R_{IN} = 400 \text{ M}\Omega$
 $C_{IN} = 1 \text{ pF}$
 $R_{OUT} < < 1 \Omega$
 $BW = 1 \text{ MHz}$

NON-INVERTING AMPLIFIER



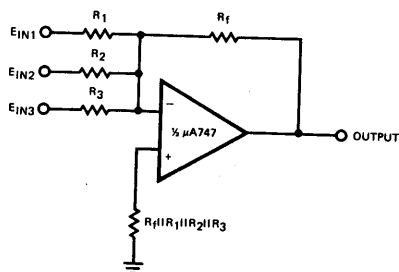
GAIN	R_1	R_2	B.W.	R_{IN}
10	1 k Ω	9 k Ω	100 kHz	400 M Ω
100	100 Ω	9.9 k Ω	10 kHz	280 M Ω
1000	100 Ω	99.9 k Ω	1 kHz	80 M Ω

INVERTING AMPLIFIER



GAIN	R_1	R_2	BW	R_{IN}
1	10 k Ω	10 k Ω	1 MHz	10 k Ω
10	1 k Ω	10 k Ω	100 kHz	1 k Ω
100	1 k Ω	100 k Ω	10 kHz	1 k Ω
1000	100 Ω	100 k Ω	1 kHz	100 Ω

WEIGHTED AVERAGING AMPLIFIER



$$-E_{OUT} = E_{IN1} \left(\frac{R_f}{R_1} \right) + E_{IN2} \left(\frac{R_f}{R_2} \right) + E_{IN3} \left(\frac{R_f}{R_3} \right)$$

μA748

OPERATIONAL AMPLIFIER

FAIRCHILD LINEAR INTEGRATED CIRCUITS

1A748

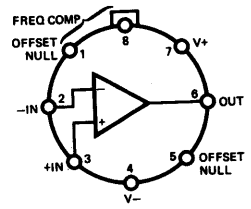
GENERAL DESCRIPTION — The μA748 is a High Performance Monolithic Operational Amplifier constructed using the Fairchild Planar* epitaxial process. It is intended for a high wide range of analog applications where tailoring of frequency characteristics is desirable. High common mode voltage range and absence of latch-up make the μA748 ideal for use as a voltage follower. The high gain and wide range of operating voltages provide superior performance in integrator, summing amplifier, and general feedback applications. The μA748 is short-circuit protected and has the same pin configuration as the popular μA741 operational amplifier. Unity gain frequency compensation is achieved by means of a single 30 pF capacitor. For superior performance, see μA777 data sheet.

- **SHORT-CIRCUIT PROTECTION**
- **OFFSET VOLTAGE NULL CAPABILITY**
- **LARGE COMMON-MODE AND DIFFERENTIAL VOLTAGE RANGES**
- **LOW POWER CONSUMPTION**
- **NO LATCH UP**

ABSOLUTE MAXIMUM RATINGS

Supply Voltage	±22 V
Internal Power Dissipation (Note 1)	
Metal Can	500 mW
DIP	670 mW
Mini DIP	310 mW
Flatpak	570 mW
Differential Input Voltage	±30 V
Input Voltage (Note 2)	±15 V
Storage Temperature Range	
Metal Can, DIP, and Flatpak	-65°C to +150°C
Mini DIP	-55°C to +125°C
Operating Temperature Range	
Military (μA748)	-55°C to +125°C
Commercial (μA748C)	0°C to +70°C
Lead Temperature (Soldering, 60 Seconds)	300°C
Metal Can, Flatpak, and Hermetic DIPs	260°C
Molded Mini DIP	Indefinite
Output Short Circuit Duration (Note 3)	Indefinite

CONNECTION DIAGRAMS
8-LEAD METAL CAN
 (TOP VIEW)
PACKAGE OUTLINE 5S
PACKAGE CODE H

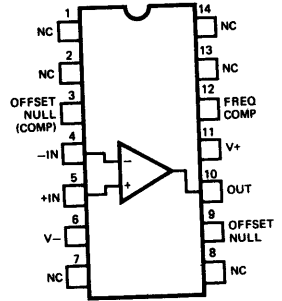


NOTE: Pin 4 connected to case

ORDER INFORMATION

TYPE	PART NO.
μA748	μA748HM
μA748A	μA748AHM
μA748C	μA748HC

14-LEAD DIP
 (TOP VIEW)
PACKAGE OUTLINE 6A
PACKAGE CODE D

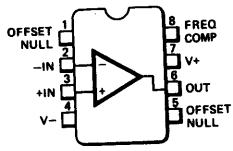


ORDER INFORMATION

TYPE	PART NO.
μA748	μA748DM
μA748A	μA748ADM
μA748C	μA748DC

CONNECTION DIAGRAMS

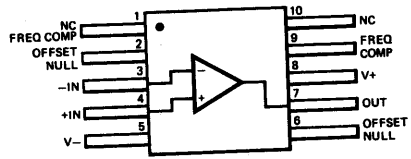
8-LEAD MINI DIP
 (TOP VIEW)
PACKAGE OUTLINE 9T
PACKAGE CODE T



ORDER INFORMATION

TYPE	PART NO.
μA748C	μA748TC

10-LEAD FLATPAK†
 (TOP VIEW)
PACKAGE OUTLINE 3F
PACKAGE CODE F



† Available on special request

ORDER INFORMATION

TYPE	PART NO.
μA748	μA748FM
μA748A	μA748AFM

*Planar is a patented Fairchild process.

FAIRCHILD LINEAR INTEGRATED CIRCUITS • μ A748

μ A748A

ELECTRICAL CHARACTERISTICS ($V_S = \pm 15$ V, $T_A = 25^\circ$ C, $C_C = 30$ pF unless otherwise specified)

PARAMETERS	CONDITIONS	MIN	TYP	MAX	UNITS
Input Offset Voltage	$R_S < 50$ k Ω		0.5	2.0	mV
Input Offset Current			2.0	10	nA
Input Bias Current			20	75	nA
Input Resistance		2.0	10.0		M Ω
Input Capacitance			3.0		pF
Offset Voltage Adjustment Range			± 25		mV
Large Signal Voltage Gain	$R_L \geq 2$ k Ω , $V_{OUT} = \pm 10$ V	50,000	250,000		V/V
Output Resistance			100		Ω
Output Short Circuit Current			± 25		mA
Supply Current			1.9	2.8	mA
Power Consumption			60	85	mW
Transient Response (Voltage Follower, Gain of 1)	$V_{IN} = 20$ mV, $C_C = 30$ pF, $R_L = 2$ k Ω , $C_L < 100$ pF		Rise Time		
			Overshoot	0.3	
Slew Rate (Voltage Follower, Gain of 1)	$R_L \geq 2$ k Ω		5.0		%
			0.5		V/ μ s
Transient Response (Voltage Follower, Gain of 10)	$V_{IN} = 20$ mV, $C_C = 3.5$ pF, $R_L = 2$ k Ω , $C_L < 100$ pF		Rise Time		
			Overshoot	0.2	
Slew Rate (Voltage Follower, Gain of 10)	$R_L \geq 2$ k Ω , $C_C = 3.5$ pF		5.0		%
			5.5		V/ μ s
The following specifications apply for -55° C $< T_A < +125^\circ$ C:					
Input Offset Voltage	$R_S < 50$ k Ω		0.5	3.0	mV
Average Input Offset Voltage Drift	$R_S < 50$ k Ω		2.5	15	μ V/ $^\circ$ C
Input Offset Current				25	nA
Average Input Offset Current Drift	25° C $< T_A < +125^\circ$ C		2.5	30	pA/ $^\circ$ C
	-55° C $< T_A < 25^\circ$ C		6.5	150	pA/ $^\circ$ C
Input Bias Current				100	nA
Input Voltage Range		± 12	± 13		V
Common Mode Rejection Ratio	$R_S < 50$ k Ω	80	95		dB
Supply Voltage Rejection Ratio	$R_S < 50$ k Ω		13	100	μ V/V
Large Signal Voltage Gain	$R_L \geq 2$ k Ω , $V_{OUT} = \pm 10$ V	25,000			V/V
Output Voltage Swing	$R_L \geq 10$ k Ω	± 12	± 14		V
	$R_L \geq 2$ k Ω	± 10	± 13		V
Supply Current	$T_A = +125^\circ$ C		1.5	2.5	mA
	$T_A = -55^\circ$ C		2.0	3.3	mA
Power Consumption	$T_A = +125^\circ$ C		40	75	mW
	$T_A = -55^\circ$ C		60	100	mW

NOTES

1. Rating applies to ambient temperatures up to 70° C. Above 70° C ambient derate linearly at 6.3 mW/ $^\circ$ C for metal can, 8.3 mW/ $^\circ$ C for the DIP 5.6 mW/ $^\circ$ C for the mini DIP and 7.1 mW/ $^\circ$ C for the flatpak.
2. For supply voltages less than ± 15 V, the absolute maximum input voltage is equal to the supply voltage.
3. Short circuit may be to ground or either supply. Rating applies to $+125^\circ$ C case temperature or $+75^\circ$ C ambient temperature.

$\mu A748$

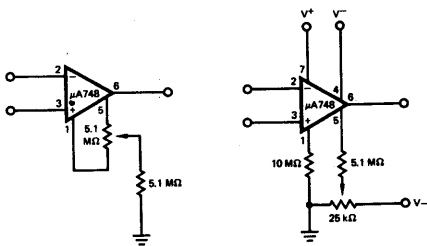
ELECTRICAL CHARACTERISTICS ($V_S = \pm 15$ V, $T_A = 25^\circ$ C, $C_C = 30$ pF unless otherwise specified)

PARAMETERS (see definitions)	CONDITIONS	MIN	TYP	MAX	UNITS
Input Offset Voltage	$R_S < 10$ k Ω		1.0	5.0	mV
Input Offset Current			20	200	nA
Input Bias Current			80	500	nA
Input Resistance		0.3	2.0		M Ω
Input Capacitance			2.0		pF
Offset Voltage Adjustment Range			± 15		mV
Large Signal Voltage Gain	$R_L > 2$ k Ω , $V_{OUT} = \pm 10$ V	50,000	150,000		V/V
Output Resistance			75		Ω
Output Short-Circuit Current			25		mA
Supply Current			1.9	2.8	mA
Power Consumption			60	85	mW
Transient Response (Voltage Follower, Gain of 1)	$V_{IN} = 20$ mV, $C_C = 30$ pF, $R_L = 2$ k Ω , $C_L < 100$ pF	Rise Time	0.3		μ s
		Overshoot	5.0		%
Slew Rate (Voltage Follower, Gain of 1)	$R_L > 2$ k Ω		0.5		V/ μ s
Transient Response (Voltage Follower, Gain of 10)	$V_{IN} = 20$ mV, $C_C = 3.5$ pF, $R_L = 2$ k Ω , $C_L < 100$ pF	Rise Time	0.2		μ s
		Overshoot	5.0		%
Slew Rate (Voltage Follower, Gain of 10)	$R_L > 2$ k Ω , $C_C = 3.5$ pF		5.5		V/ μ s

The following specifications apply for -55° C $< T_A < +125^\circ$ C:

Input Offset Voltage	$R_S < 10$ k Ω		1.0	6.0	mV
Input Offset Current	$T_A = +125^\circ$ C		10	200	nA
	$T_A = -55^\circ$ C		50	500	nA
Input Bias Current	$T_A = +125^\circ$ C		0.03	0.5	μ A
	$T_A = -55^\circ$ C		0.3	1.5	μ A
Input Voltage Range		± 12	± 13		V
Common Mode Rejection Ratio	$R_S < 10$ k Ω	70	90		dB
Supply Voltage Rejection Ratio	$R_S < 10$ k Ω		30	150	μ V/V
Large Signal Voltage Gain	$R_L > 2$ k Ω , $V_{OUT} = \pm 10$ V	25,000			V/V
Output Voltage Swing	$R_L > 10$ k Ω	± 12	± 14		V
	$R_L > 2$ k Ω	± 10	± 13		V
Supply Current	$T_A = +125^\circ$ C		1.5	2.5	mA
	$T_A = -55^\circ$ C		2.0	3.3	mA
Power Consumption	$T_A = +125^\circ$ C		45	75	mW
	$T_A = -55^\circ$ C		60	100	mW

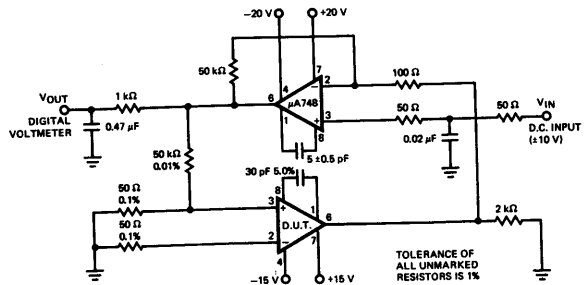
VOLTAGE OFFSET
NULL CIRCUIT



SUGGESTED

ALTERNATE

GAIN TEST CIRCUIT



$$A_{VO} = \frac{V_{IN} \times 10^3}{V_{OUT}} = \frac{10 \times 10^3}{V_{OUT}} \text{ FOR } V_{IN} \text{ SPECIFIED}$$

$\mu A748C$

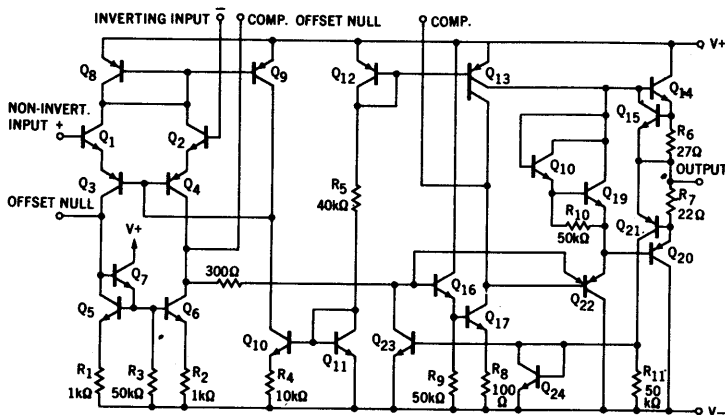
ELECTRICAL CHARACTERISTICS ($V_S = \pm 15\text{ V}$, $T_A = 25^\circ\text{C}$, $C_C = 30\text{ pF}$ unless otherwise specified)

PARAMETERS		CONDITIONS	MIN	TYP	MAX	UNITS
Input Offset Voltage		$R_S < 10\text{ k}\Omega$		2.0	6.0	mV
Input Offset Current				20	200	nA
Input Bias Current				80	500	nA
Input Resistance			0.3	2.0		M Ω
Input Capacitance				2.0		pF
Offset Voltage Adjustment Range				± 15		mV
Large Signal Voltage Gain		$R_L > 2\text{ k}\Omega$, $V_{OUT} = \pm 10\text{ V}$	20,000	150,000		V/V
Output Resistance				75		Ω
Output Short-Circuit Current				25		mA
Supply Current				1.9	2.8	mA
Power Consumption				60	85	mW
Transient Response (Voltage Follower, Gain of 1)	Rise Time	$V_{IN} = 20\text{ mV}$, $C_C = 30\text{ pF}$, $R_L = 2\text{ k}\Omega$, $C_L < 100\text{ pF}$		0.3		μs
	Overshoot			5.0		%
Slew Rate (Voltage Follower, Gain of 1)		$R_L > 2\text{ k}\Omega$		0.5		V/ μs
Transient Response (Voltage Follower, Gain of 10)	Rise Time	$V_{IN} = 20\text{ mV}$, $C_C = 3.5\text{ pF}$, $R_L = 2\text{ k}\Omega$, $C_L < 100\text{ pF}$		0.2		μs
	Overshoot			5.0		%
Slew Rate (Voltage Follower, Gain of 10)		$R_L > 2\text{ k}\Omega$, $C_C = 3.5\text{ pF}$		5.5		V/ μs

The following specifications apply for $0^\circ\text{C} < T_A < +70^\circ\text{C}$:

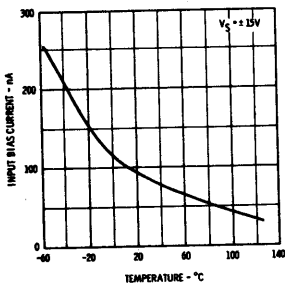
Input Offset Voltage	$R_S < 10\text{ k}\Omega$			7.5	mV
Input Offset Current				300	nA
Input Bias Current				800	nA
Input Voltage Range		± 12	± 13		V
Common Mode Rejection Ratio	$R_S < 10\text{ k}\Omega$	70	90		dB
Supply Voltage Rejection Ratio	$R_S < 10\text{ k}\Omega$		30	150	$\mu\text{V/V}$
Large Signal Voltage Gain	$R_L > 2\text{ k}\Omega$, $V_{OUT} = \pm 10\text{ V}$	15,000			V/V
Output Voltage Swing	$R_L > 10\text{ k}\Omega$	± 12	± 14		V
	$R_L > 2\text{ k}\Omega$	± 10	± 13		V
Power Consumption			60	100	mW

EQUIVALENT CIRCUIT

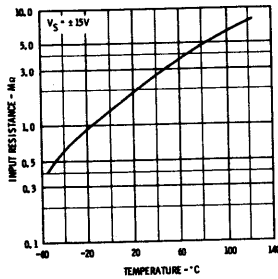


TYPICAL PERFORMANCE CURVES FOR $\mu A748$

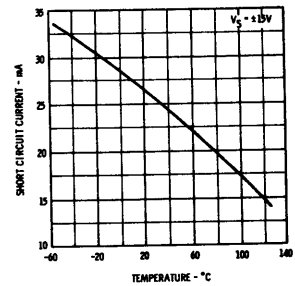
INPUT BIAS CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



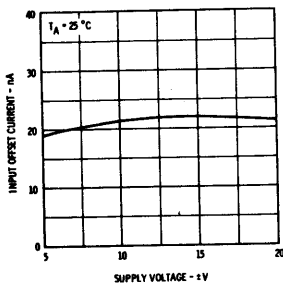
INPUT RESISTANCE AS A FUNCTION OF AMBIENT TEMPERATURE



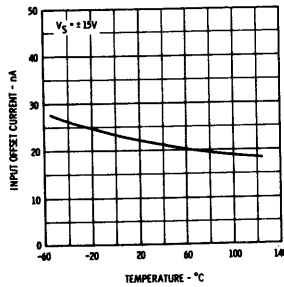
OUTPUT SHORT-CIRCUIT CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



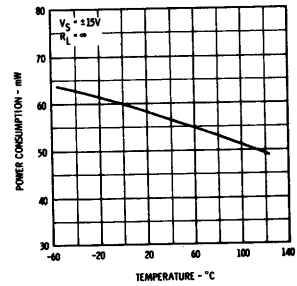
INPUT OFFSET CURRENT AS A FUNCTION OF SUPPLY VOLTAGE



INPUT OFFSET CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE

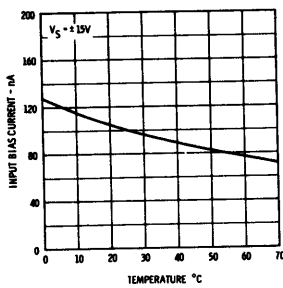


POWER CONSUMPTION AS A FUNCTION OF AMBIENT TEMPERATURE

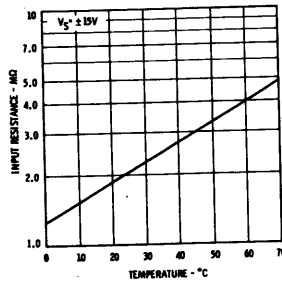


TYPICAL PERFORMANCE CURVES FOR $\mu A748C$

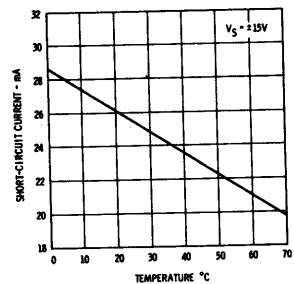
INPUT BIAS CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



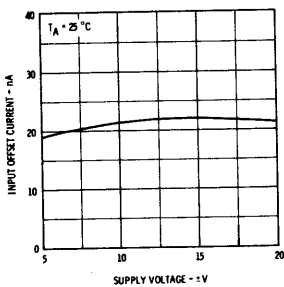
INPUT RESISTANCE AS A FUNCTION OF AMBIENT TEMPERATURE



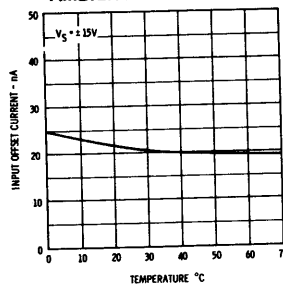
OUTPUT SHORT-CIRCUIT CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



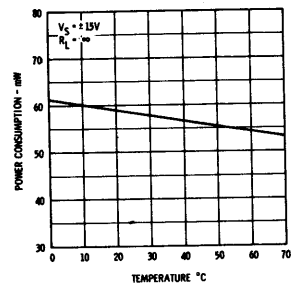
INPUT OFFSET CURRENT AS A FUNCTION OF SUPPLY VOLTAGE



INPUT OFFSET CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE

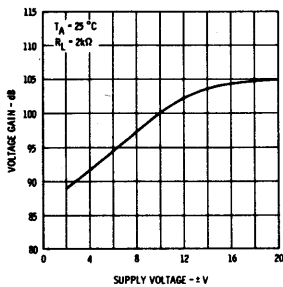


POWER CONSUMPTION AS A FUNCTION OF AMBIENT TEMPERATURE

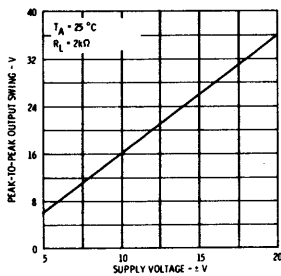


TYPICAL PERFORMANCE CURVES FOR $\mu A748$ AND $\mu A748C$

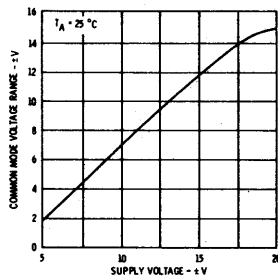
OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF SUPPLY VOLTAGE



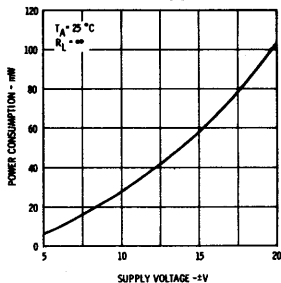
OUTPUT VOLTAGE SWING AS A FUNCTION OF SUPPLY VOLTAGE



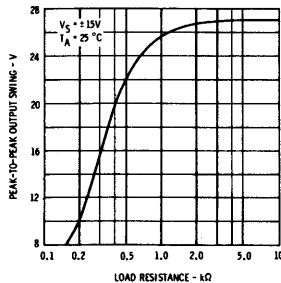
INPUT COMMON MODE VOLTAGE RANGE AS A FUNCTION OF SUPPLY VOLTAGE



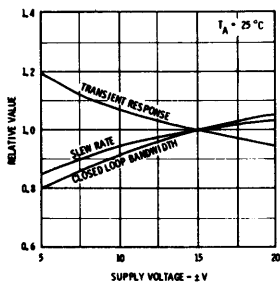
POWER CONSUMPTION AS A FUNCTION OF SUPPLY VOLTAGE



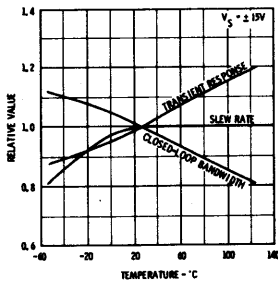
OUTPUT VOLTAGE SWING AS A FUNCTION OF LOAD RESISTANCE



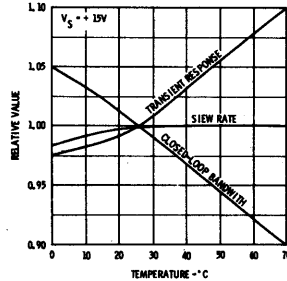
FREQUENCY CHARACTERISTICS AS A FUNCTION OF SUPPLY VOLTAGE



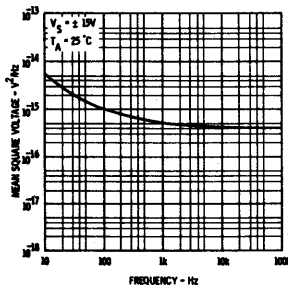
$\mu A748$ FREQUENCY CHARACTERISTICS AS A FUNCTION OF AMBIENT TEMPERATURE



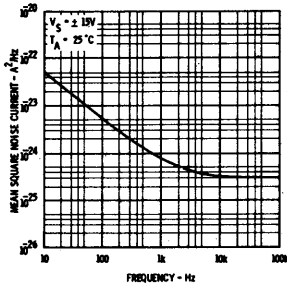
748C FREQUENCY CHARACTERISTICS AS A FUNCTION OF AMBIENT TEMPERATURE



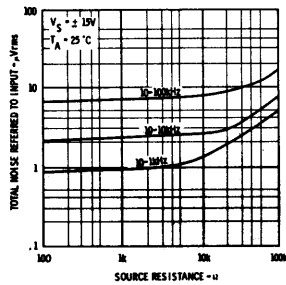
INPUT NOISE VOLTAGE AS A FUNCTION OF FREQUENCY



INPUT NOISE CURRENT AS A FUNCTION OF FREQUENCY



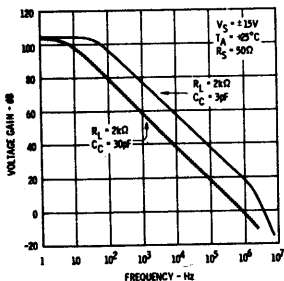
BROAD BAND NOISE FOR VARIOUS BANDWIDTHS



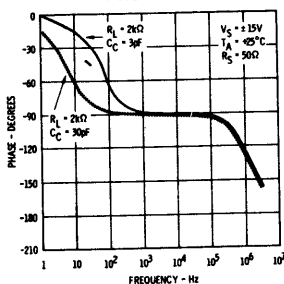
→ 749

TYPICAL PERFORMANCE CURVES FOR $\mu A748$ AND $\mu A748C$

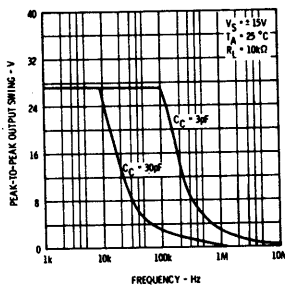
OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF FREQUENCY



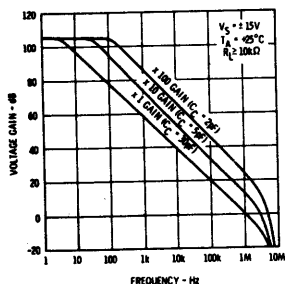
OPEN LOOP PHASE RESPONSE AS A FUNCTION OF FREQUENCY



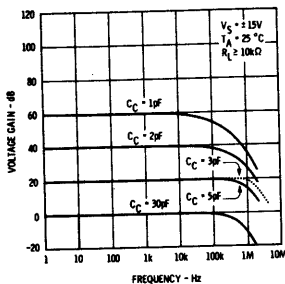
OUTPUT VOLTAGE SWING AS A FUNCTION OF FREQUENCY



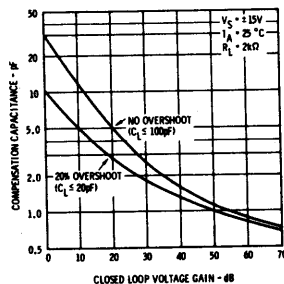
OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF FREQUENCY FOR VARIOUS GAIN/COMPENSATION OPTIONS



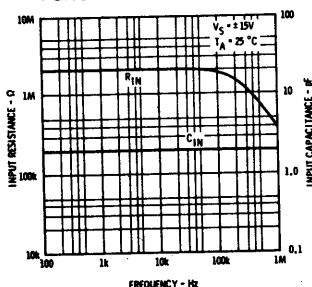
FREQUENCY RESPONSE FOR VARIOUS CLOSED LOOP GAINS



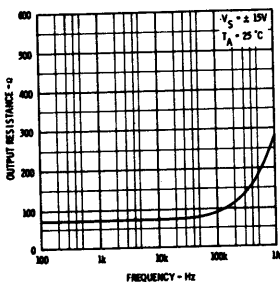
COMPENSATION CAPACITANCE AS A FUNCTION OF CLOSED LOOP VOLTAGE GAIN



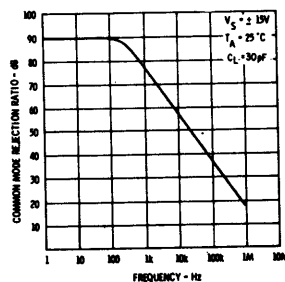
INPUT RESISTANCE AND INPUT CAPACITANCE AS A FUNCTION OF FREQUENCY



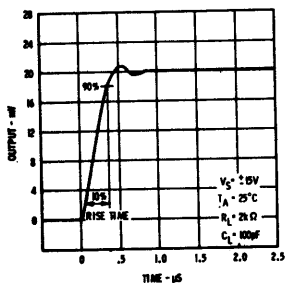
OUTPUT RESISTANCE AS A FUNCTION OF FREQUENCY



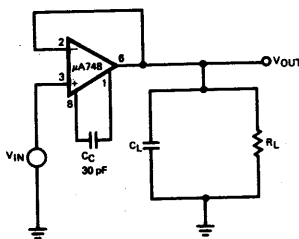
COMMON MODE REJECTION RATIO AS A FUNCTION OF FREQUENCY



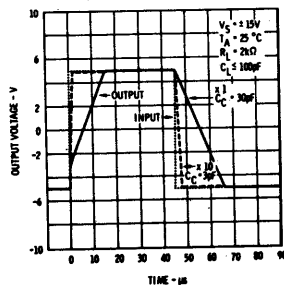
VOLTAGE FOLLOWER TRANSIENT RESPONSE (GAIN OF 1)



TRANSIENT RESPONSE TEST CIRCUIT

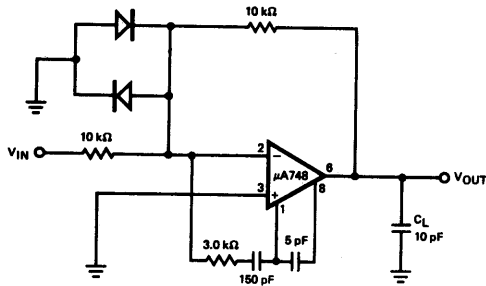


VOLTAGE FOLLOWER LARGE-SIGNAL PULSE RESPONSE

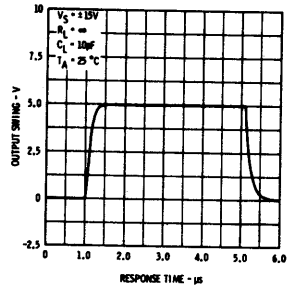


TYPICAL PERFORMANCE CURVES FOR $\mu A748$ AND $\mu A748C$

FEED FORWARD COMPENSATION

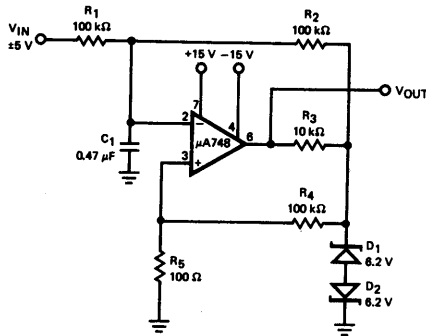


LARGE SIGNAL FEED FORWARD TRANSIENT RESPONSE



TYPICAL APPLICATIONS

PULSE WIDTH MODULATOR



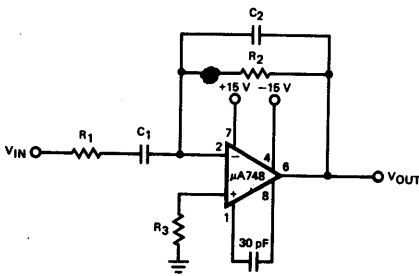
$$f_c = \frac{1}{2\pi R_2 C_1}$$

$$f_n = \frac{1}{2\pi R_1 C_1}$$

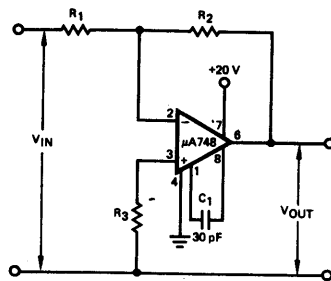
$$= \frac{1}{2\pi R_2 C_2}$$

$$f_c < f_n < f_{\text{unity gain}}$$

PRACTICAL DIFFERENTIATOR



CIRCUIT FOR OPERATING THE $\mu A748$ WITHOUT A NEGATIVE SUPPLY



NOTES

1. Rating applies to ambient temperature up to 70°C. Above 70°C ambient derate linearly at 6.3 mW/°C for the metal can, 8.3 mW/°C for the DIP, 5.6 mW/°C for the mini DIP and 7.1 mW/°C for the flatpak.
2. For supply voltages less than ±15 V, the absolute maximum input voltage is equal to the supply voltage.
3. Short circuit may be to ground or either supply. Rating applies to +125°C case temperature or +75°C case temperature or +75°C ambient temperature.